iSBC 80/30 SINGLE BOARD COMPUTER HARDWARE **REFERENCE MANUAL**

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This manual provides general information, installation, programming information, principles of operation, and service information for the Intel iSBC 80/30 Single Board Computer. Additional information is available in the following documents:

- Intel MCS-85 User's Manual, Order No. 98-366
- Intel UPI-41 User's Manual, Order No. 9800504
- Intel 8080/8085 Assembly Language Programming Manual, Order No. 9800301
- Intel 8255A Programmable Peripheral Interface, Application Note AP-15
- Intel 8251 Universal Synchronous/Asynchronous Receiver/Transmitter, Application Note AP-16
- Intel MULTIBUS Interfacing, Application Note AP-28
- Intel 8259 Programmable Interrupt Controller, Application Note AP-31



CONTENTS

CHAPTER 1 GENERAL INFORMATION

Introduction	1-1
Description	1-1
System Software Development	1-3
Equipment Supplied	1-4
Equipment Required	1-4
Specifications	1-4

PAGE

CHAPTER 2 PREPARATION FOR USE

l
ľ
l
1
1
1
1
3
3
4
4
4
1
4
4
7
7
8
2
2
2
5
5
3
4
4
7
7
9

CHAPTER 3 PROGRAMMING INFORMATION

Introduction	3-1
Failsafe Timer	3-1
Memory Addressing	3-1
I/O Addressing	3-2
System Initialization	3-2
8251A USART Programming	3-3
Mode Instruction Format	3-3
Sync Characters	3-3
Command Instruction Format	3-3
Reset	3-4
Addressing	3-4
Initialization	3-4

	PAGE
Operation	3-5
Data Input/Output	3-5
Status Read	3-6
8253 PIT Programming	3-6
Mode Control Word and Count	3-7
Addressing	
Initialization	3-10
Operation	3-11
Counter Read	3-11
Clock Frequency/Divide Ratio Selection	3-11
Rate Generator/Interval Timer	3-12
Interrupt Timer	3-12
8255A PPI Programming	3-13
Control Word Format	
Addressing	3-14
Initialization	3-14
Operation	3-14
Read Operation	3-14
Write Operation	3-14
8259A PIC Programming	3-14
Interrupt Priority Modes	3-14
Fully Nested Mode	
Auto-Rotating Mode	
Specific Rotating Mode	
Interrupt Mask	3-15
Status Read	3-15
Initialization Command Words	
Operation Command Words	3-16
Addressing	3-16
Initialization	3-16
Operation	
8041/8741A UPI Programming	3-17
Interrupt Handling	
TRAP Input	
RST 7.5, 6.5, 5.5 Inputs	3-23
INTR Input	
▲	

CHAPTER 4 PRINCIPLES OF OPERATION

Introduction	-1
Functional Description 4	-1
Clock Circuits 4	-1
Central Processor Unit 4	-1
Interval Timer 4	-1
Serial I/O	-1
Parallel I/O 4	-1
Universal Peripheral Interface	1-2
Interrupt Control 4	1-2
ROM/EPROM Configuration 4	1-2
RAM Configuration 4	1-2
Bus Interface	1-2
Dual-Port Control 4	1-2
Circuit Analysis 4	1-2

CONTENTS (Continued)



Initialization
Clock Circuits 4-3
8085A CPU Timing 4-3
Instruction Timing 4-3
Opcode Fetch Timing 4-4
Memory Read Timing 4-6
I/O Read Timing 4-6
Memory Write Timing 4-6
I/O Write Timing 4-7
Interrupt Acknowledge Timing 4-7
Address Bus 4-9
Bus Time Out 4-9
Data Bus
Read/Write Signal Generation 4-9
I/O Control Signals 4-9
Memory Control Signals 4-9
Dual Port Control Logic 4-9
Bus Access Timing
CPU Access Timing4-12
Multibus Interface
I/O Operation 4-13

PAGE

	PAGE
On-Board I/O Operation	4-13
System I/O Operation	4-13
ROM/EPROM Operation	4-13
RAM Operation	4-14
RAM Controller	4-14
On-Board Read/Write Operation	4-14
Bus Read/Write Operation	4-14
Interrupt Operation	4-15

CHAPTER 5

SERVICE	INFORMATION
T	

Introduction	2-1
Replaceable Parts	5-1
Service Diagrams	5-1
Service and Repair Assistance	5-1

APPENDIX A 8085 INSTRUCTION SET

APPENDIX B TELETYPEWRITER MODIFICATIONS



TABLES

TABLE	TITLE PAG	E
1-1.	Specifications 1-	4
2-1.	User-Furnished and Installed Components 2-	2
2-2.	User-Furnished Connector Details 2-	3
2-3.	Line Driver and I/O Terminator Locations 2-	4
2-4.	Jumper Selectable Options 2-	-5
2-5.	ROM/EPROM Configuration Jumpers 2-	.7
2-6.	Jumper Configuration for On-Board	
	8085A CPU Access of On-Board RAM 2	-7
2-7.	65K Page System Memory Selection 2-	-8
2-8.	8K/16K Block Selection Within 65K Page 2-	-8
2-9.	Priority Interrupt Jumper Matrix2-1	1
2-10.	Connector J3 Pin Assignments Vs.	
	Jumper Configuration	2
2-11.	8255A Port Configuration Jumpers	3
2-12.	Multibus Connector P1 Pin Assignments2-1	6
2-13.	Multibus Signal Functions	7
2-14.	iSBC 80/30 DC Characteristics	8

TABLE	TITLE PA	GE
2-15.	iSBC 80/30 AC Characteristics	
	(Master Mode)	2-21
2-16.	iSBC 80/30 AC Characteristics (Slave Mode)	2-21
2-17.	Auxiliary Connector P2 Pin Assignments	2-25
2-18.	Auxiliary Signal (Connector P2)	
	DC Characteristics	2-26
2-19.	Connector J1 Pin Assignments	2-27
2-20.	Connector J2 Pin Assignments	2-27
2-21.	Parallel I/O Signal (Connector J1/J2)	
	DC Characteristics	2-28
2-22.	Connector J3 Vs. RS232C Pin Correspondence	2-29
3-1.	On-Board Memory Address (for CPU Access)	3-1
3-2.	I/O Address Assignments	3-2
3-3.	Typical USART Mode or Command	
	Instruction Subroutine	3-5
3-4.	Typical USART Data Character	
	Read Subroutine	3-6

v

TABLES (Continued)



TABLE	TITLE P.	AGE	TABLE	TITLE PAGE
3-5.	Typical USART Data Character		3-18.	Typical PIC Initialization Subroutine
	Write Subroutine	3-6	3-19.	PIC Operation Procedures
3-6.	Typical USART Status Read Subroutine	. 3-7	3-20.	Typical PIC Interrupt Request Register
3-7.	PIT Counter Operation Vs. Gate Inputs	.3-10		Read Subroutine
3-8.	Typical PIT Control Word Subroutine	.3-10	3-21.	Typical PIC In-Service Register
3-9.	Typical PIT Count Value Load Subroutine	.3-11		Read Subroutine
3-10.	Typical PIT Counter Read Subroutine	.3-11	3-22.	Typical PIC Set Mask Register Subroutine 3-20
3-11.	PIT Count Value Vs. Rate Multiplier		3-23.	Typical PIC Mask Register Read Subroutine3-21
	for Each Baud Rate	3-12	3-24.	Typical PIC End-of-Interrupt
3-12.	PIT Rate Generator Frequencies and			Command Subroutine
	Timer Intervals	3-13	3-25.	Typical UPI Data Byte Read Subroutine3-22
3-13.	PIT Time Intervals Vs. Timer Counts	.3-13	3-26.	Typical UPI Data Byte Write Subroutine3-22
3-14.	Typical PPI Initialization Subroutine	.3-14	3-27.	8085A CPU Restart Interrupt Vectors
3-15.	Typical PPI Port Read Subroutine		4-1.	CPU Status and Control Lines 4-4
3-16.	Typical PPI Port Write Subroutine	3-15	5-1.	Replaceable Parts 5-1
3-17.	PIC Device Address Insertion	.3-16	5-2.	List of Manufacturers' Codes 5-3

FIGURE



ILLUSTRATIONS

PAGE

TITLE

FIGURE	TITLE	PAGE
1-1.	iSBC 80/30 Single Board Computer	1-1
2-1.	Jumper Configuration for Multibus Access of	f
	On-Board RAM (16-Bit Address System)	2-9
2-2.	Jumper Configuration for Multibus Access of	f
	On-Board RAM (20-Bit Address System)	2-10
2-3.	Bus Exchange Timing (Master Mode)	2-22
2-4.	Bus Exchange Timing (Slave Mode)	2-23
2-5.	Serial Priority Resolution Scheme	2-24
2-6.	Parallel Priority Resolution Scheme	2-25
3-1.	USART Synchronous Mode Instruction	
	Word Format	3-3
3-2.	USART Synchronous Mode Transmission	
	Format	3-3
3-3.	USART Asynchronous Mode Instruction	
	Word Format	3-3
3-4.	USART Asynchronous Mode Transmission	
	Format	3-4
3-5.	USART Command Instruction Word Format	3-4
3-6.	Typical USART Initialization and	
	I/O Data Sequence	3-5
3-7.	USART Status Read Format	3-7
3-8.	PIT Mode Control Word Format	3-8
3-9.	PIT Programming Sequence Examples	3-9
3-10.	PIT Counter Register Latch Control	
	Word Format	
3-11.	PPI Control Word Format	3-13

3-12.	PPI Port C Bit Set/Reset Control
	Word Format
3-13.	PIC Device Interrupt Addresses
3-14.	PIC Initialization Command Word Formats3-17
3-15.	PIC Operation Control Word Formats
3-16.	UPI Data Bus Buffer and
	Status Registers
4-1.	iSBC 80/30 Input/Output and
	Interrupt Block Diagram
4-2.	iSBC 80/30 ROM/EPROM and Dual Port
	RAM Block Diagram4-19
4-3.	Typical CPU Instruction Cycle 4-4
4-4.	Opcode Fetch Machine Cycle 4-5
4-5.	Opcode Fetch Machine Cycle (With Wait) 4-5
4-6.	Memory Read (or I/O Read) Machine Cycles . 4-6
4-7.	Memory Write (or I/O Write) Machine Cycles 4-7
4-8.	Interrupt Acknowledge Machine Cycles 4-8
4-9.	Dual Port Control Bus Access Timing
	With CPU Lockout
4-10.	Dual Port Control CPU Access Timing
	With Bus Lockout4-11
5-1.	iSBC 80/30 Parts Locations Diagram 5-5
5-2.	iSBC 80/30 Schematic Diagram 5-7
5-3.	iSBC 604 Schematic Diagram5-25
5-4.	iSBC 614 Schematic Diagram5-27



CHAPTER 1 GENERAL INFORMATION

1-1. INTRODUCTION

The iSBC 80/30 Single Board Computer, which is a member of Intel's complete line of iSBC 80 computer products, is a computer system on a single printedcircuit assembly. The iSBC 80/30 includes a central processor unit (CPU), 16K bytes of dynamic random access memory (RAM), one serial and three parallel I/O ports, a programmable timer, priority interrupt logic, and Multibus control logic. Also included is dual port control logic to allow the iSBC 80/30 to act as a slave RAM device to other bus masters in the system. Provision is made for user-installation of masked or programmable read only memory (ROM or EPROM) and an Intel 8041 or 8741A Universal Peripheral Interface.

1-2. DESCRIPTION

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The iSBC 80/30 Single Board Computer (figure 1-1) is controlled by an Intel 8085A Microprocessor (CPU), which includes six 8-bit general-purpose registers and an accumulator. The six general-purpose registers may be addressed individually or in pairs, which allows both single precision and double precision operations. The CPU has a 16-bit program counter which allows direct addressing of up to 65K of memory. An external stack, located within any portion of read/write memory, may be used as a last-in/first-out storage area for the contents of the program counter, flags, accumulator, and all six general-purpose registers. A 16-bit stack pointer controls the addressing of this external stack, which allows subroutine nesting that is bounded only by the 65K address limitation.

The iSBC 80/30 has an internal bus for all on-board memory and I/O operations and accesses the system bus (Multibus) for all external memory and I/O operations. Hence, local (on-board) operations do not involve the Multibus and allow true parallel processing when several bus masters (e.g., DMA devices and other single board computers) are used in a multimaster scheme.

The 16K of dynamic RAM is implemented with eight Intel 2717 chips and an Intel 8202 Dynamic RAM Controller. Dual port control logic is included to interface this 16K RAM with the Multibus so that the iSBC 80/30 can function as a slave RAM device when not in control of the Multibus. The CPU has priority when accessing on-board RAM. After the CPU completes its read or write operation, the controlling bus master is allowed to access RAM and complete its operation. Where both the CPU and the controlling bus master have the need to write or read





iSBC 80/30

several words to or from on-board RAM, their operations are interleaved. The slave RAM decode logic allows extended Multibus addressing so that bus masters having a 20-bit address capability can partition the iSBC 80/30 RAM into any 8K or 16K segment in a 1-megabyte address space. The CPU, however, has only 16 address lines and memory must therefore reside in the 0-65K byte address space. There is no conflict in assigning RAM addresses for CPU access and slave access since separate decoding logic is used.

Jumpers are included to allow the user to reserve 8K bytes of on-board RAM for use by the 8085A CPU only. This reserved RAM address space is not accessible via the Multibus and does not occupy any system address space.

Two IC sockets are included to accommodate up to 8K of user-installed ROM or EPROM. Configuration jumpers allow ROM or EPROM to be installed in 1K, 2K, or 4K increments. All on-board ROM/EPROM operations are performed at maximum processor speed.

The iSBC 80/30 includes 24 programmable parallel I/O lines implemented by means of an Intel 8255A Programmable Peripheral Interface (PPI). The system software is used to configure the I/O lines in any combination of unidirectional input/output and bidirectional ports. The I/O interface may be customized to meet specific peripheral requirements and, in order to take full advantage of the large number of possible I/O configurations, IC sockets are provided for interchangeable I/O line drivers and terminators. Hence, the flexibility of the parallel I/O interface is further enhanced by the capability of selecting the appropriate combination of optional line drivers and terminators to provide the required sink current, polarity, and drive/termination characteristics for each application. The 24 programmable I/O lines and signal ground lines are brought out to a 50-pin edge connector (J1) that mates with flat, woven, or round cable.

Sockets are provided for a user-supplied Intel 8041/8741A Universal Programmable Interface (UPI) and associated line drivers and terminators. The 8041/8741A is a singlechip microcomputer which contains a CPU, 1K bytes of ROM (8041) or EPROM (8741), 64 bytes of RAM, 16 programmable I/O lines, and an 8-bit timer. Special interface registers are included in the chip which enable the UPI to function as a slave processor to the 8085A CPU. The UPI allows the user to specify algorithms for controlling user peripherals directly in the chip, thereby relieving the 8085A CPU for other system functions. An RS232C driver and an RS232C receiver are included so that the UPI may optionally be used to handle a simple serial I/O interface. In addition to providing the capability of user-supplied algorithms for the 8041/8741A the iSBC 80/30 supports all the preprogrammed 8041/8741A devices such as the Intel 8278 Keyboard Encoder, 8294 Data Encryption Controller, and 8295 Matrix Printer Driver.

The RS232C compatible serial I/O port is controlled and interfaced by an Intel 8251A USART (Universal Synchronous/Asynchronous Receiver/Transmitter) chip. The USART is individually programmable for operation in most synchronous or asynchronous serial data transmission formats (including IBM Bi-Sync).

In the synchronous mode the following are programmable:

- a. Character length,
- b. Sync character (or characters), and
- c. Parity.
- In the asynchronous mode the following are programmable:
- a. Character length,
- b. Baud rate factor (clock divide ratios of 1, 16, or 64).
- c. Stop bits, and
- d. Parity.

In both the synchronous and asynchronous modes, the serial I/O port features half- or full-duplex, double-buffered transmit and receive capability. In addition, USART error detection circuits can check for parity, overrun, and framing errors. The USART transmit and receive clock rates are supplied by a programmable baud rate/time generator. These clocks may optionally be supplied from an external source. The RS232C command lines, serial data lines, and signal ground lines are brought out to a 50-pin edge connector (J3) that mates with flat or round cable.

Three independent, fully programmable 16-bit interval timer/event counters are provided by an Intel 8253 Programmable Interval Timer (PIT). Each counter is capable of operating in either BCD or binary modes; two of these counters are available to the systems designer to generate accurate time intervals under software control. Routing for the outputs and gate/trigger inputs of two of these counters is jumper-selectable; the outputs of these two counters may be independently routed to the 8259A Programmable Interrupt Controller (PIC), the I/O line drivers associated with the 8255A Programmable Peripheral Interface (PPI), the 8041/8741A Universal Programmable Interface, or used as inputs to the 8255A PPI and 8041/8741A (UPI). The gate/trigger inputs of the two counters may be routed to I/O terminators associated with the 8255A PPI or as output connections from the 8255A PPI. The third counter is used as a programmable baud rate generator for the serial I/O port. In utilizing the iSBC 80/30, the systems designer simply configures, via software, each counter independently to meet system requirements. Whenever a given time delay or count is needed, software commands to the 8253 PIT select the desired function. The contents of each counter may be read at any time during system operation with simple operations for event counting applications, and special commands are included so that the contents of each counter can be read "on the fly."

The iSBC 80/30 provides vectoring for 12 interrupt levels, four of which are handled directly by the interrupt processing capability of the 8085A CPU. These four levels (TRAP, RST 7.5, RST 6.5, and RST 5.5) represent (in decreasing order of priority) the four highest priority interrupts of the iSBC 80/30. These four interrupts generate the following unique memory address: TRAP (24H), RST 7.5 (3CH), RST 6.5 (34H), and RST 5.5 (2CH). An 8085A JUMP instruction at each of these addresses then provides linkage to interrupt service routines located independently anywhere in the lower 65K bytes of memory. All interrupt inputs with the exception of TRAP may be masked via software. The TRAP interrupt should be used for conditions such as power-down sequences which require immediate attention by the 8085A CPU.

An Intel 8259A Programmable Interrupt Controller (PIC) provides vectoring for the next eight interrupt levels. The PIC treats each true input signal condition as an interrupt request. After resolving the interrupt priority, the PIC issues a single interrupt request to the CPU. Interrupt priorities are independently programmable under software control. Similarly, an interrupt can be masked under software control. The programmable interrupt priority modes are:

- a. Fully Nested Priority. Each interrupt request has a fixed priority: input 0 is highest, input 7 is lowest.
- b. Auto-Rotating Priority. Each interrupt request has equal priority. Each level, after receiving service, becomes the lowest priority level until the next interrupt occurs.
- c. Specific Priority. Software assigns lowest priority. Priority of all other levels is in numerical sequence based on lowest priority.

The PIC, which can be programmed to respond to edgesensitive or level-sensitive inputs, generates a unique memory address for each interrupt level. These addresses are equally spaced at intervals of 4 to 8 (software selectable) bytes. This 32- or 64-byte block may be located to begin at any 32- or 64-byte boundary in the 65,536 byte memory space. A single 8085A JUMP instruction at each of these addresses then provides linkage to locate each interrupt service routine independently anywhere in memory.

Interrupt requests may originate from 18 sources. Two jumper-selectable interrupt requests can be automatically generated by the Programmable Peripheral Interface (PPI) when a byte of information is ready to be transferred to the 8085A CPU (i.e., input buffer is full) or a byte of information has been transferred to a peripheral device (i.e., output buffer is empty). Two jumper-selectable interrupt requests can be automatically generated by the USART when a character is ready to be transferred to the 8085A CPU (i.e., receive channel buffer is full) or when a character is ready to be transmitted (i.e., transmit channel data buffer is empty). A jumper-selectable interrupt request can be generated by two of the programmable counters and by the Universal Peripheral Interface (UPI). Eight additional interrupt request lines are available to the user for direct interfaces to user designated periphral devices via the Multibus, and two interrupt request lines may be jumper routed directly from peripherals via the parallel I/O driver/ terminator section.

Control logic is also included for generation of a Power-Fail Interrupt, which works in conjunction with an AC LOW signal from an Intel iSBC 635 Power Supply or equivalent.

The iSBC 80/30 includes the resources for supporting a variety of OEM system requirements. For those applications requiring additional processing capacity and the benefits of multiprocessing (i.e., several CPU's and/or controllers logically sharing systems tasks with communication over the Multibus), the iSBC 80/30 provides full bus arbitration control logic. This control logic allows up to three bus masters (e.g., any combination of iSBC 80/30, iSBC 80/20, DMA controller, diskette controller, etc.) to share the Multibus in serial (daisy-chain) fashion or up to 16 bus masters to share the Multibus using an external parallel priority resolving network.

The Multibus arbitration logic operates synchronously with the bus clock, which is derived either from the iSBC 80/30 or can be optionally generated by some other bus master. Data, however, is transferred via a handshake between the controlling master and the addressed slave module. This arrangement allows different speed controllers to share resources on the same bus, and transfers via the bus proceed asynchronously. Thus, the transfer speed is dependent on transmitting and receiving devices only. This design prevents slower master modules from being handicapped in their attempts to gain control of the bus, but does not restrict the speed at which faster modules can transfer data via the same bus. The most obvious applications for the master-slave capabilities of the bus are multiprocessor configurations, high-speed direct memory access (DMA) operations, and high-speed peripheral control, but are by no means limited to these three.

1-3. SYSTEM SOFTWARE DEVELOPMENT

Intel's RMX/80 Real-Time Multitasking Software, specifically designed for Intel iSBC 80 single board computers, provides the capability to monitor and control multiple asynchronous external events. The RMX/80 Executive, which synchronizes and controls the execution of multiple tasks, is provided as a linkable and relocatable module that requires only 2K bytes of memory. Optional linkage and relocatable modules for teletypewriter and CRT control, diskette file system, high-speed mathematics unit, and analog subsystems are also available.

The development cycle of iSBC 80/30 based products may be significantly reduced using an Intel Intellec Microcomputer Development System. The resident macroassembler, text editor, and system monitor greatly simplify the design, development, and debug of iSBC 80/30 system software. An optional diskette operating system provides a relocating macroassembler, relocating loader and linkage editor, and a library manager. A unique In-Circuit Emulator (ICE-85) option provides the capability of developing and debugging software directly on the iSBC 80/30.

Intel's high level programming language, PL/M, is also available as a resident Intellec Microcomputer Development System option. PL/M provides the capability to program in a natural, algorithmic language and eliminates the need to manage register usage or allocate memory. PL/M programs can be written in a much shorter time than assembly language programs for a given application.

1-4. EQUIPMENT SUPPLIED

The following are supplied with the iSBC 80/30 Single Board Computer:

- a. Schematic diagram, dwg no. 2002132
- b. Assembly drawing, dwg no. 1001576

1-5. EQUIPMENT REQUIRED

Because the iSBC 80/30 is designed to satisfy a variety of applications, the user must purchase and install only those components required to satisfy his particular needs. A list of components required to configure all the intended applications of the iSBC 80/30 is provided in table 2-1.

1-6. SPECIFICATIONS

Specifications of the iSBC 80/30 Single Board Computer are listed in table 1-1.

WORD SIZE Instruction: Data:	8, 16, or 24 bits. 8 bits.
CYCLE TIME:	1.44 μ sec ±0.1% for fastest executable instruction; i.e., four clock cycles.
MEMORY CAPACITY On-Board ROM/EPROM:	Up to 8K bytes; user installed in 1K, 2K, or 4K increments.
On-Board RAM:	16K bytes of dynamic RAM; integrity maintained during power failure with user- furnished batteries.
Off-Board Expansion:	Up to 65K bytes of user-specified combinations of RAM, ROM, and EPROM.
MEMORY ADDRESSING On-Board ROM/EPROM:	0-07FF (using 2708 or 2758 EPROM's or 8308 ROM's); 0-0FFF (using 2716 EPROM's or 2316E ROM's); 0-1FFF (using 2332 ROM's).
On-Board RAM (CPU Access):	Jumpers allow on-board CPU to access either 8K or 16K. For 8K RAM access, addresses may be set on 8K boundaries 2000, 4000, E000. For 16K access, addresses may be set on 16K boundaries 4000, 8000, or C000. One or both 8K segments may be reserved for CPU use only.
On-Board RAM (Multibus Access):	Jumpers allow board to act as slave for 8K or 16K RAM access by another bus master; 16-bit or 20-bit addressing is accommodated and addresses are irrespective of addresses used for CPU access. For 16-bit addressing, boundaries may be set on any 8K or 16K segment of 65K byte address space. For 20-bit addressing, boundaries may be set on any 8K or 16K segment of 1M byte address space.

Table 1-1. Specifications

Table 1-1. Specifications (Continued)

SERIAL COMMUNICATIONS Synchronous:

5-, 6-, 7-, or 8-bit characters. Internal; 1 or 2 sync characters. Automatic sync insertion.

Asynchronous:

5-, 6-, 7-, or 8-bit characters. Break character generation. 1, 1¹/₂, or 2 stop bits. False start bit detection.

Sample Baud Rate:

Frequency1	Baud Rate (Hz) ²				
(kHz, Software Selectable)	Synchronous	Asynch	ironous		
		÷16	÷64		
153.6		9600	2400		
76.8	-	4800	1200		
38.4	38400	2400	600		
19.2	19200	1200	300		
9.6	9600	600	150		
4.8	4800	300	75		
2.4	2400	150			
1.76	1760	110	—		

Notes: 1. Frequency selected by I/O writes of appropriate 16-bit frequency factor to Baud Rate Register.

 Baud rates shown here are only a sample subset of possible softwareprogrammable rates available. Any frequency from 18.75 Hz to 614.4 kHz may be generated utilizing on-board crystal oscillator and 16-bit Programmable Interval Timer (used here as frequency divider).

INTERVAL TIMER AND BAUD RATE GENERATOR

Input Frequency (selectable):

2.46 MHz \pm 0.1% (0.41 μ sec period nominal), 1.23 MHz \pm 0.1% (0.82 μ sec period nominal), and 153.6 kHz \pm 0.1% (6.5 μ sec period nominal).

Function	Single	Timer	Dual Timers (Two Timers Cascaded)		
	Min.	Max.	Min.	Max.	
Real-Time Interrupt Interval	1.63 µsec	426 msec	3.26 µsec	465 28 minutes	
Rate Generator (Frequency)	2.348 Hz	614.4 kHz	0.000036 Hz	307.2 kHz	

SYSTEM CLOCK (8085A CPU):

Output Frequencies:

2.7648 MHz ±0.1%.

1-5

Table 1-1. Specifications (Continued)

I/O ADDRESSING:	All communication to Parallel I/O and Serial I/O Ports, Timer, and Interrupt Controller is via read and write commands from on-board 8085A CPU. Refer to table 3-2.			
INTERFACE COMPATIBILITY Serial I/O:	EIA Standard RS Carrier Detec Clear to Seno Data Set Rea Data Termina Request to So Receive Cloc	232C signals provi t Reco d Ring dy Sec I Ready Sec end Tran k Tran	ded and supported: eive Data I Indicator ondary Receive Data ondary Transmit Data smit Clock smit Data	
Parallel I/O:	24 programmable IC sockets includ required for interfa	lines (8 lines per po ed for user installat ace ports. Refer to	rt); one port includes ion of line drivers ar table 2-1.	bidirectional bus driver. nd/or I/O terminators as
Optional I/O:	IC socket include included are IC so required for interfa	d for Intel 8041/874 ckets for user instal ace. Refer to table 2	1 Universal Periphe lation of line drivers a 2-1.	ral Interface (UPI). Also Ind/or I/O terminators as
INTERRUPTS:	8085A CPU includ	des five interrupt inp nemory locations fo	outs, each of which v r entry point to servi	rectors processor to the ce routine:
	Interrupt Input	Vector Address	Priority	Туре
	TRAP RST 7.5 RST 6.5 RST 5.5 INTR	24H 3CH 34H 2CH Note	Highest	Non-maskable Maskable Maskable Maskable Maskable Maskable
	Note: INTR in vide ve	nput provided by 82 ector CALL address	59A PIC, which is pro of service routine for in	ogrammable to pro- nterrupting device.
	Jumpers allow sel be programmed to	ection of 12 priority o respond to edge-	interrupts from 18 inte sensitive or level-ser	errupt sources. PIC may sitive inputs.
COMPATIBLE CONNECTORS/CABLES:	Refer to table 2-2 2-30 for recomme	for compatible conr nded types and len	nector details. Refer ligths of I/O cables.	to paragraphs 2-29 and
ENVIRONMENTAL REQUIREMENTS Operating Temperature:	0° to 55°C (32° to	131°F).		
Relative Humidity:	To 90% without c	ondensation.		
PHYSICAL CHARACTERISTICS Width: Depth: Thickness: Weight:	30.48 cm (12.00 ii 17.15 cm (6.75 in 1.27 cm (0.50 incl 425 gm (15 ounce	nches). ches). h). ss).		•

POWER REQUIREMENTS	S:			
CONFIGURATION	$V_{CC} = +5V \pm 5\%$	$V_{DD} = +12V \pm 5\%$	$V_{BB} = -5V \pm 5\%$	$V_{AA} = -12V \pm 5\%$
Without EPROM ¹	3.5A	220 mA		50 mA
With 8041/8741 UPI ²	3.6A	220 mA		50 mA
RAM Only ³	350 mA	20 mA	2.5 mA	—
With iSBC 5304	3.5A	320 mA		150 mA
With 2K EPROM⁵ (using 8708)	4.4A	350 mA	95 mA	40 mA
With 2K EPROM⁵ (Using 8758)	4.6A	220 mA		50 mA
With 4K EPROM⁵ (Using 2716)	(4.6A)	(220 mA)	_	(50 mA ⟩
With 8K ROM⁵ (Using 2332)	4.6A	220 mA		50 mA

Table 1-1. Specifications (Continued)

Notes: 1. Does not include power for optional ROM/EPROM, 8041/8741 UPI, I/O drivers, and I/O terminators.

2. Does not include power required for optional ROM/EPROM, I/O drivers and I/O terminators.

3. RAM chips powered via auxiliary power bus.

4. Does not include power for optional ROM/EPROM, 8041/8741 UPI, I/O drivers, and I/O terminators. Power for iSBC 530 is supplied via serial port connector.

5. Includes power required for two ROM/EPROM chips, 8041/8741 UPI, and I/O terminators installed for 34 I/O lines; all terminator inputs low.



CHAPTER 2 PREPARATION FOR USE

2-1. INTRODUCTION

This chapter provides instructions for preparing the iSBC 80/30 Single Board Computer for use in the user-defined environment. It is advisable that the contents of Chapters 1 and 3 be fully understood before beginning the configuration and installation procedures provided in this chapter.

2-2. UNPACKING AND INSPECTION

Inspect the shipping carton immediately upon receipt for evidence of mishandling during transit. If the shipping carton is severely damaged or waterstained, request that the carrier's agent be present when the carton is opened. If the carrier's agent is not present when the carton is opened and the contents of the carton are damaged, keep the carton and packing material for the agent's inspection.

For repairs to a product damaged in shipment, contact the Intel Technical Support Center (see paragraph 5-4) to obtain a Return Authorization Number and further instructions. A purchase order will be required to complete the repair. A copy of the purchase order should be submitted to the carrier with your claim.

It is suggested that salvageable shipping cartons and packing material be saved for future use in the event the product must be reshipped. Important criteria for installing and interfacing the iSBC 80/30 in the above environments are presented in following paragraphs.

2-4. USER-FURNISHED COMPONENTS

Because the iSBC 80/30 is designed to satisfy a variety of applications, the user need purchase and install only those components required to satisfy his particular configuration. A list of components required to configure all the intended applications of the iSBC 80/30 are listed in table 2-1. Table 2-2 lists details, types, and vendors of those connectors referenced in table 2-1.

2-5. POWER REQUIREMENTS

The iSBC 80/30 requires +5V, -5V, +12V, and -12V power supply inputs. The currents required from these supplies are listed in table 1-1. (The -5V supply is *mandatory* only if Intel 2708 EPROM chips are installed; an on-board regulator that operates off the -12V supply can otherwise supply the -5V power.)

2-6. COOLING REQUIREMENT

The iSBC 80/30 dissipates 401 gram-calories/minute (1.62 Btu/minute) and adequate circulation of air must be provided to prevent a temperature rise above 55° C (131°F). The System 80 enclosures and the Intellec System include fans to provide adequate intake and exhaust of ventilating air.

2-3. INSTALLATION CONSIDERATIONS

The iSBC 80/30 is designed for use in one of the following configurations:

- a. Standalone (single-board) system.
- b. Bus master in a single bus master system.
- c. Bus master in a multiple bus master system.

2-7. PHYSICAL DIMENSIONS

Physical dimensions of the iSBC 80/30 are as follows:

- a. Width: 30.48 cm (12.00 inches).
- b. Height: 17.15 cm (6.75 inches).
- c. Thickness: 1.25 cm (0.50 inch).

ltem No.	Item	Description				Use		
1	iSBC 604	Modular Backplane and Cardcage. In- cludes four slots with bus terminators.				Provides power input pins and Multibus signal interface between iSBC 80/30 and		
		(See figure 5-3.)				three additional boards in a multiple board system.		
2	iSBC 614	Modular Backplane and Cardcage. In- cludes four slots without bus terminators. (See figure 5-4.)				Provides four-slot extension of iSBC 604.		
3	Connector (mates with P1)	See Multit table 2-2.	ous conne	ector	details in	Power inputs and Multibus signal interface Not required if iSBC 80/30 is installed in in an iSBC 604/614.		
4	Connector (mates with P2)	See Auxili table 2-2.	ary conne	ector	details in	Auxiliary backup battery inputs and asso- ciated memory protect functions.		
5	Connector (mates with J1)	See paralle table 2-2.	el I/O coni	necto	or details in	Interfaces parallel I/O ports to Intel 8255A Programmable Peripheral Interface (PPI).		
6	Connector (mates with J2)	See paralle table 2-2.	el I/O con	necto	or details in	Interfaces I/O ports to optional Intel 8041/ 8741A Universal Peripheral Interface (UPI).		
7	Connector (mates with J3)	See serial I/O connector details in table 2-2.			r details in	Interfaces serial I/O port to Intel 8251A Programmable Communications Interface (USART).		
8	ROM/EPROM Chips	Chips One or two each of the following Intel ROM/EPROM chips:		On-board UV erasable PROM for program development and/or dedicated program				
		ROM	EPRO	М	BITS	employed. Use either ROM or EPROM; do not mix.		
		8308	2708		1K × 8			
		2316E 2332	2756		2K × 8 4K × 8			
9	Intel 8041/8741A	Universal Pe	eripheral In	terfac	æ (UPI).	Single chip microcomputer with program memory, data memory, CPU, event timer, I/O, and clock oscillator. Inter- faces two 8-bit I/O ports; two additional input bits (T0 and T1) for conditional branch and event timer functions.		
10	Line Drivers	Туре			Current	Used for interface to Intel 8255A and		
		SN7403 I, C SN7400 I SN7408 NI SN7409 NI,	403 I, OC 16 mA 400 I 16 mA 408 NI 16 mA 409 NI, OC 16 mA		16 mA 16 mA 16 mA 16 mA	optional Intel 8041/8741. Requires two line driver IC's for each 8-bit parallel output port. (Exception: refer to para- graph 2-11.)		
		Types selected as typical; I = invert- ing, NI = noninverting, and OC = open collector.			I = invert- OC = open	-		
11	I/O Terminators	Intel iSBC 901 Divider or iSBC 902 Pull-Up: iSBC 901 iSBC 901 iSBC 901 iSBC 901 iSBC 902 iSBC 902			r iSBC 902 / ∽ / ~	Used for interface to Intel 8255A and optional Intel 8041/8741A. Requires two 901's or two 902's for each 8-bit parallel input port. (Exception: refer to paragraph 2-11.) Additional 901 or 902 required for 8041/8741 if T0 and T1 inputs are used for conditional branch or event timer functions.		
12	Capacitors	Seven capacitors as required.				Rise time/noise capacitors for serial I/O port.		

Function	No. Of Pairs/ Pins	Centers (inches)	Connector Type	Vendor	Vendor Part No.	Intel Part No.
Parallel I/O Connector	25/50	0.1	Flat Crimp	3M 3M AMP ANSLEY SAE	3415-0000 WITH EARS 3415-0001 W/O EARS 88083-1 609-5015 SD6750 SERIES	iSBC 956 Cable Set
Parallel I/O Connector	25/50	0.1	Soldered	AMP VIKING TI	2-583485-6 3VH25/1JV5 H312125	N/A
Parallel I/O Connector	25/50	0.1	Wirewrap ¹	TI VIKING CDC ³ ITT CANNON	H311125 3VH25/1JND5 VPB01B25D00A1 EC4A050A1A	N/A
Serial I/O Connector	13/26	0.1	Flat Crimp	3M AMP ANSLEY SAE	3462-0001 88106-1 609-2615 SD6726 SERIES	iSBC 955 Cable Set
Serial I/O Connector	13/26	0.1	Soldered	TI AMP	H312113 1-583485-5	N/A
Serial I/O Connector	13/26	0.1	Wirewrap ¹	TI	H311113	N/A
Multibus Connector	43/86	0.156	Soldered ¹	CDC ³ MICRO PLASTICS ARCO VIKING	VPB01E43D00A1 MP-0156-43-BW-4 AE443WP1 LESS EARS 2VH43/1AV5	N/A
Multibus Connector	43/86	0.156	Wirewrap ^{1,2}	CDC ³ CDC ³ VIKING	VFB01E43D00A1 or VPB01E43A00A1 2VH43/1AV5	MDS 985
Auxiliary Connector	30/60	0.1	Soldered ¹	TI VIKING	H312130 3VH30/1JN5	N/A
Auxiliary Connector	30/60	0.1	Wirewrap ^{1,2}	CDC ³ TI	VPB01B30A00A2 H311130	N/A

Table 2-2.	User-Furnished	Connector	Details
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NOTES:

1. Connector heights are not guaranteed to conform to OEM packaging equipment.

2. Wirewrap pin lengths are not guaranteed to conform to OEM packaging equipment.

3. CDC VPB01 ..., VPB02 ..., VPB04 ..., etc. are identical connectors with different electroplating thicknesses or metal surfaces.

2-8. COMPONENT INSTALLATION

Instructions for installing the optional ROM/EPROM, Intel 8041/8741A Universal Peripheral Interface, line drivers, I/O terminators, and rise time/noise capacitors are given in following paragraphs. When installing the optional chips, be sure to orient pin 1 of the chip adjacent to the white dot located near pin 1 of the associated IC socket. The grid location on figure 5-1 (parts location diagram) and figure 5-2 (schematic diagram) are specified for each user-installed component. Because the schematic diagram consists of nine sheets, grid references to figure 5-2 consist of four alphanumeric characters. For example, grid reference 6ZD4 signifies sheet 6 Zone D4.

2-9. ROM/EPROM CHIPS

Install the ROM/EPROM chips in IC sockets A25 and A37. (Refer to figure 5-1 zone ZC3 and figure 5-2 zone 3ZA3.) Sockets A25 and A37, respectively, accommodate the low order and high-order addresses of the ROM/EPROM chip pair. For instance, if two Intel 2716 EPROM chips are installed, the chip installed in IC socket A25 is assigned addresses 0000-07FF; the chip installed in IC socket A37 is assigned addresses 0800-0FFF.

The default (factory connected) jumpers are configured for Intel 2316E ROM or 2716 EPROM. If different type chips are installed, reconfigure the jumpers as described in paragraph 2-14.

2-10. UNIVERSAL PERIPHERAL INTERFACE

Install the optional Intel 8041/8741A Universal Peripheral Interface (UPI) chip in socket A20. (Refer to figure 5-1 zone ZC6 and figure 5-2 zone 5ZC4.)

2-11. LINE DRIVERS AND I/O TERMINATORS

Table 2-3 lists the I/O ports and the location of associated IC sockets for installing either line drivers or I/O terminators. (Refer to table 2-1 items 10 and 11.) Port E8 is factory equipped with Intel 8226 Bidirectional Bus Drivers and requires no additional components. (Refer to paragraphs 2-22 and 2-23.)

2-12. RISE TIME/NOISE CAPACITORS

Eye pads are provided so that rise time/noise capacitors may be installed as required on the individual serial I/O pins. The selection of capacitor values is at the option of the user and is normally a function of the particular environment. The location of these eye pads are as follows:

Capacitor	Fig. 5-1	Fig. 5-2	
C11	ZD4	6ZD4	
C12	ZD4	6ZB3	
C13	ZD3	6ZC4	
C14	ZD4	6ZD4	
C16	ZD3	6ZD6	
C17	ZD3	6ZC6	
C18	ZD3	6ZD6	

2-13. JUMPER CONFIGURATION

The iSBC 80/30 includes a variety of jumper-selectable options to allow the user to configure the board for his particular application. Table 2-4 summarizes these jumper-selectable options and lists the grid reference locations of the jumpers as shown in figure 5-1 (parts location diagram) and figure 5-2 (schematic diagram). Because the schematic consists of nine sheets, grid references to figure 5-2 consists of four alphanumeric characters. For example, grid reference 3ZB7 signifies sheet 3 Zone B7.

Study table 2-5 carefully while making reference to figures 5-1 and 5-2. If the default (factory configured) jumper wiring is appropriate for a particular function, no further actions is required for that function. If, however, a different configuration is required, remove the default jumper(s) and install an optional jumpers(s) as specified. For most options, the information in table 2-4 is sufficient for proper configuration. Additional information, where necessary for clarity, is described in subsequent paragraphs.

2-14. ROM/EPROM CONFIGURATION

Table 2-5 lists the jumper configurations and associated address block for the various types of compatible Intel ROM/EPROM chips.

2-15. ON-BOARD RAM ADDRESSES

This on-board RAM can be accessed by the on-board 8085A microprocessor (CPU) as well as by other bus masters in the system via the Multibus. Addresses for on-board 8085A access and for system access are assigned as described in paragraphs 2-16 and 2-17, respectively.

	I/O Port	Bits	Driver/ Terminator	Fig. 5-1 Grid Ref.	Fig. 5-2 Grid Ref
	E8	0-7	None Required	_	
8255A PPI	E9	0-3 4-7	A5 A6	ZD6 ZD6	4ZA3 4ZA3
Interface	EA	0-3 4-7	A4 A3	ZD7 ZD7	4ZC3 4ZB3
8041/8741 UPI Interface (Optional)	1	0-3 4-7	A7 A8	ZD6 ZD6	5ZD3 5ZC3
	2	0-3 4-7	A10 A11	ZD5 ZD5	5ZB3 5ZB3
		T0, T1	A9	ZD5	5ZC3

Table 2-3. Line Driver and I/O Terminator Locations

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Table 2-4.	Jumper	Selectable	Options
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Function	Fig. 5-1 Grid Ref.	Fig. 5-2 Grid Ref.	Description		
ROM/EPROM Configuration			Six jumpers accommodate one of four types of user-installed ROM or EPROM chips. One jumper required between two posts in each of the following six groups of posts:		
	ZC3 ZC3 ZD2 ZD2 ZD2 ZD2 ZD2	3ZB7 3ZC6 3ZB4 3ZA3 3ZA4 3ZA3	112 thru 114 157 thru 159 153 thru 156 84 thru 88 99 thru 101 102 thru 105		
On-Board RAM (On-Board Access)	ZD2	3ZD7, 3ZD6	One jumper wire selects 8K or 16K access and one jumper wire selects base address foron-board 8085A access of on-board RAM. Jumper W default position *A-B selects 16K access and default jumper *98-92 selects base address 4000H. Refer to paragraphs 2-15 and 2-16 if reconfiguration is required.		
On-Board RAM (System Access)	ZB7 ZB7 ZB7 ZB7	5ZC6 5ZB6 5ZB5	 16-Bit Address System: Three jumper wires select base address for system access: W5 (one): Must be set to position *K-L. W4 (one): 8K or 16K access. 171 thru 180 (one): Base address. Refer to paragraphs 2-17 and 2-18. 		
	ZB7 ZB7 ZB7 ZB7 ZB7	5ZC7 5ZC6 5ZB6 5ZB5	 20-Bit Address System: Five jumper wires select base address for system access: W6 (two): Upper or lower 512K space. W5 (one): 65K page select. W4 (one): 8K or 16K access. 171 thru 180 (one): Base address. Refer to paragraphs 2-17 and 2-19. 		
Bus Clock	ZB7	5ZA4	Default jumper *165-166 routes Bus Clock signal BCLK/ to the Multibus. (Refer to table 2-13.) Remove this jumper <i>only</i> if another bus master supplies this signal.		
Constant Clock	ZB7	5ZA4	Default jumper *167-168 routes Constant Clock signal CCLK/ to the Multibus. (Refer to table 2-13.) Remove this jumper <i>only</i> if another bus master supplies this signal.		
Bus Priority Out	ZB7	8ZD2	Default jumper *169-170 routes Bus Priority Out signal BPRO/ to the Multibus (Refer to table 2-13.) Remove this jumper <i>only</i> in those systems employing a parallel priority bus resolution scheme. (Refer to paragraph 2-27.)		
Bus Priority Resolution	ZB7	8ZD6	One jumper defines one of two modes of resolving bus contention in a multiple bus master system: *162-163: Can request Multibus as needed.		
		<u>-</u>	163-164: Always requesting Multibus; should only be used when iSBC 80/30 has lowest priority.		
Auxiliary Backup Batteries	ZB5	1ZC5, 1ZB5	If auxiliary backup batteries are employed to sustain memory during ac power outages, remove default jumpers *W8, *W9, and *W10.		
On-Board –5V Regulator	ZB6	1ZB2	The 80/30 requires a $-5V$ supply for Intel 2708 EPROM chips and a $-5V$ AUX input for the on-board RAM chips. The system $-5V$ supply is mandatory only if Intel 2708 EPROM chips are employed. If Intel 2708 EPROM chips are not employed and no system $-5V$ supply is available, the -5 AUX input to the on-board RAM chips can be supplied by the on-board $-5V$ regulator or by an auxiliary backup battery. (The on-board $-5V$ regulator operates from the system $-12V$ supply.) If neither a system $-5V$ supply nor an auxiliary backup battery is used, enable the $-5V$ regulator by connecting jumper *W10.		

Table	2-4.	Jumper	Selectable	Options	(Continued)

Function	Fig. 5-1 Grid Ref.	Fig. 5-2 Grid Ref.	Description
Failsafe Timer	ZC8	1ZC6	If the on-board 8085A CPU addresses either a system or an on-board memory or I/O device and that device does not return an acknow- ledge signal, the 8085A will hang up in a wait state. A failsafe timer is triggered during T ₁ of every machine cycle and, if not retriggered within 10 milliseconds, the resultant time-out pulse can be used to allow the 8085A to exit the wait state. If this feature is desired, connect jumper 115-116.
RAM Refresh			The 8202 Memory Controller for on-board RAM is jumper selectable as follows to select the automatic or invisible refresh mode:
	ZC4	2ZA3	*110-111: Automatic refresh (normal) mode. 110-106: Invisible refresh (on request) mode.
			In the <i>automatic</i> refresh mode, a wait state can be imposed on the on-board 8085A if a memory access occurs while a memory refresh cycle is in progress. In this case, the 8085A is forced to wait until the refresh cycle is complete. In the <i>invisible</i> refresh mode, the 8202 works around memory accesses by refreshing memory during the instruction decode clock cycle which follows each instruction fetch.
Timer Input Frequency			Input frequencies to the 8253 Programmable Interval Timer counters are jumper selectable as follows:
			Counter 0 (8041/8741 Event Clock)
	ZD5	7ZB3	*47-52: Same as Counter 2. 47-51: 153.6 kHz. 47-48: External Event Clock (from Port EA).
			Counter 1 (Count Out)
	ZD5	7ZA4	 *46-54: 1.2288 MHz. 46-52: Same as Counter 2. 46-50: Counter 0 output. 46-48: External Event Clock (from Port EA). Jumper 46-50 effectively connects Counter 0 and 1 in series in which the output of Counter 0 serves as the input clock to Counter 1. This permits lower clock rates to Counter 1 and, in turn, Counter 1 provides longer time intervals.
			Counter 2 (8251A Baud Rate Clock)
	ZD5	7ZB5	*53-54: 1.2288 MHz. 53-49: 2.4576 MHz.
Event Clock			The Event Clock is a jumper option for Port C (Port EA) of the 8255A PPI. The Event Clock may be provided by the following optional sources:
	ZD5	7ZB4	48-50: Same as PIT Counter 0 (8041/8741 Event Clock). 48-51: 153.6 kHz. 48-52: Same as PIT Counter 2 clock.
Priority Interrupt			A jumper matrix provides a wide selection of interrupts to be interfaced to the on-board 8085A and the Multibus. The matrix includes the following jumpers (refer to paragraph 2-20):
	ZC6 ZC6 ZB6	Sheet 7	117 thru 134 136 thru 152 181 thru 190
8251A Serial I/O Port Configuration		Sheet 6	Jumpers are used to input the serial I/O port transmit clock, receive clock, ring indicator, carrier detect, etc., from several sources. Refer to paragraph 2-21.
8255A Parallel I/O Port Configuration		Sheet 4	Jumpers are used to configure Ports E8, E9, and EA for the desired operating mode. Refer to paragraph 2-22.
8041/8741A Universal Peripheral Interface	_	Sheet 5	Jumpers are used to select various input and/or output signals depending on the application. Refer to paragraph 2-23.

Table 2-5.	ROM/EPROM	Configuration	Jumpers
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ROM/EPROM Type	Jumpers	Address Block
2708 EPROM or 8308 ROM	112-113, 158-159, 100-101, 104-105, 155-154, 86-84	0000-07FF
2758 EPROM	112-113, 158-159, 100-101, 104-103, 155-154, 86-87	0000-07FF
*2716 EPROM or *2316E ROM	112-113, 157-158, 100-101, 104-103, 155-156, 86-85	0000-0FFF
2332 ROM	112-114, 157-158, 100-99, 104-102, 155-153, 86-85	0000-1FFF
*Default jumpers 2316E ROM. D necessary if inst	are connected for type 2 isconnect and reconfigu alling different type ROM	716 EPROM or ire jumpers as /EPROM.

2-16. ON-BOARD 8085A ACCESS. Jumper W1 position B-C allows the on-board 8085A to access 8K of the 16K RAM; jumper W1 default position A-B allows access to all 16K. For 8K access, the RAM address block may be configured on 8K boundaries 2000, 4000, . . . E000; for 16K access, the RAM address block may be configured on 16K boundaries 4000, 8000, or C000. (Address boundary 0000 is reserved for ROM/EPROM.) Default jumper 98-92 selects boundary 4000 for both 8K and 16K access. If only 8K access is desired, remove jumper W1 from position A-B and install it in position B-C. If a different address boundary is desired, reconfigure the jumpers as listed in table 2-6.

2-17. SYSTEM ACCESS. The on-board RAM can be shared by other bus masters in the system via the Multibus. If one or more bus masters have a 20-bit address capability, the extended addressing jumpers allow the onboard RAM to reside anywhere within a 1-megabyte address space. (The on-board 8085A can only access memory in the lower 65K address space.) If it is not desired to have the on-board RAM shared by the system, leave default jumper 180-179 installed. Otherwise, configure the jumpers for 16-bit addressing or 20-bit addressing as described in following paragraphs.

NOTE

Addresses for system access of on-board RAM are completely independent of addresses for onboard 8085A access of on-board RAM.

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Jumper	RAM Add	RAM Address Block ¹				
oumpor	8K Access ²	16K Access ³				
98-91	2000-3FFF					
*98-92	4000-5FFF	4000-7FFF				
98-93	6000-7FFF					
98-94	8000-9FFF	8000-BFFF				
98-95	A000-BFFF					
98-96	C000-DFFF	C000-FFFF				
98-97	E000-FFFF					
NOTES:						
1. Address block 0000-1FFF is reserved for ROM/EPROM.						
2. Rec	Requires jumper W1 position B-C installed.					
3. Rec	Requires jumper W1 position *A-B installed.					
*De is	efault jumper; disconne required.	ct if reconfiguration				

Table 2-6. Jumper Configuration for On-Board8085A Access of On-Board RAM

2-18. 16-BIT ADDRESS SYSTEMS. For system access, the on-board RAM is divided into eight 65K pages. In 16-bit address systems, there is no page selection capability and jumper W5 default position K-L restricts the addresses to Page 0 (the only page in a 16-bit system). (Refer to table 2-7.)

As shown in table 2-8, the system can access either 8K or 16K of the on-board RAM. Jumper W4 default position B-A limits the system to 8K access; jumper W4 position B-C allows access of all 16K of on-board RAM.

One jumper wire places the on-board RAM in the desired 8K or 16K segment of the selected 65K page. To access 8K, for example, the 8K segment can be placed on any 8K boundary 0000 (1st 8K), 2000 (2nd 8K), 4000 (3rd 8K), . . . E000 (8th 8K). To access 16K, the 16K segment can be placed on any 16K boundary 0000 (1st 16K), 4000 (2nd 16K), 8000 (3rd 16K), or C0000 (4th 16K). Figure 2-1 illustrates a step-by-step sequence for establishing RAM addresses for 16-bit address systems.

2-19. 20-BIT ADDRESS SYSTEMS. In 20-bit address systems, the on-board RAM can reside anywhere within a 1-megabyte address space. As shown in table 2-7, the RAM is first placed in the lower or upper 524K by jumper W6. Jumper W5 then selects one of eight 65K pages within the upper or lower 524K bytes.

Next, referring to table 2-8, the system can access either 8K or 16K of the on-board RAM. Default jumper W4 position B-A limits the system to 8K access; jumper W4 position B-C allows access of all 16K of on-board RAM.

Table 2-7. OSK Page System Memory Selection	`able 2-7.	65K Page	System	Memory	Selection
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Low/(High)¹ System Memory	65K Page No.	Address Range ¹			
N/A (Note 2)	0 W5: *K-L	0000-FFFF			
0-524K (525-1048K)	0 W5: K-A	00000-0FFFF (80000-8FFFF)			
W6: *B-C *D-F	1 W5: К-В	10000-1FFFF (90000-9FFFF)			
D-L	W5: ² К-С	20000-2FFFF (A0000-AFFFF)			
W6: (B-E) (D-A)	8 W5: К-D	30000-3FFFF (B0000-BFFFF)			
	4 W5: К-Е	40000-4FFFF (C0000-CFFFF)			
	w5: к-ғ	50000-5FFFF (D0000-DFFFF)			
	6 W5: К-G	60000-6FFFF (E0000-EFFFF)			
	7 W5: К-Н	70000-7FFFF (F0000-FFFFF)			
NOTES:					
 Notation in parentheses applies to high (upper 524K) bytes of 20-bit system address space. 					
2. Syste use ji	ems without 20-bit ad umper W5 in positio	dress capability must n *K-L.			
*Defa requi	ault Jumper; disconne red.	ect if reconfiguration is			
N/A =	 not applicable. 				

8K or 16K	Address Block Within 65K Page (Refer to Table 2-7)		
System Access	Jumper	Block	
8K W4: *B-A	180-172 180-173 180-174 180-175 180-176 180-177 180-178 180-179 *180-171	1st 8K 2nd 8K 3rd 8K 4th 8K 5th 8K 6th 8K 7th 8K 8th 8K No Access	
16K W4: B-C	180-173 180-175 180-177 180-177 180-179 *180-171	1st 16K 2nd 16K 3rd 16K 4th 16K No Access	

Table 2-8. 8K/16K Block SelectionWithin 65K Page

Finally, one jumper wire places the on-board RAM in the desired 8K or 16K segment of the selected 65K page. To access an 8K segment in Page 4 of the lower 524K, for example, the 8K segment can be placed on any 8K boundary 40000 (1st 8K), 42000 (2nd 8K), 44000 (3rd 8K), . . . 4E000 (8th 8K). To access a 16K segment in Page 4 of the lower 524K, the 16K segment can be placed on any 16K boundary 40000 (1st 16K), 44000 (2nd 16K), 48000 (3rd 16K), or 4C000 (4th 16K). Figure 2-2 illustrates a step-by-step sequence for establishing RAM addresses for a 20-bit address system.

2-20. PRIORITY INTERRUPTS

Table 2-9 lists the source (from) and destination (to) of the interrupt matrix shown in figure 5-2 sheet 7. For example, note that the 8259A Programmable Interrupt Controller (PIC) can handle eight positive-true interrupt requests and, after resolving any priority contention, outputs an interrupt request to the INTR input of the 8085A micro-processor.

Study table 2-9 carefully while making reference to figure 5-2 sheet 7 before deciding on a definite priority configuration for the iSBC 80/30. There are two areas that require some explanation: the 8085A TRAP and RST 7.5 interrupts.

Default jumper 137-145 grounds the TRAP interrupt input to prevent the possibility of false interrupts being generated by noise spikes. Since the TRAP interrupt is not maskable, cannot be disabled by the program, and has the highest priority, it should be used only to detect a catastrophic event such as a power failure or a bus failure.

The RST 7.5 interrupt input is edge-sensitive only and is default jumpered to the COUNT OUT output of Counter 1. If it is desired to interrupt the 8085A if the Failsafe Timer times out, remove jumper 123-138 (COUNT OUT) and connect jumper 122-138 (BUS TIME OUT). (The BUS TIME OUT signal is generated when the Failsafe Timer is retriggered after timing out.)

NOTE

The 8259A PIC can be programmed to respond to either edge-sensitive or level-sensitive interrupt requests. If the PIC is programmed to respond to edge-sensitive interrupt requests, the PIC will respond only to a low-to-high transition on any one of the individual IR input lines.



(3) Jumper 180-178 selects 7th 8K segment of 65K page 0 (the only page in a 16-bit address system).

Thus, the lower 8K of on-board memory is assigned system addresses C000 thru DFFF. In this example, this same 8K of memory has on-board 8085A addresses A000 thru 5FFF.

Note that for 8K access, the 1st, 3rd, 5th, and 7th 8K bytes access the lower 8K of on-board RAM. The 2nd, 4th, 6th, and 8th 8K bytes access the upper 8K of on-board RAM. With jumper W4 in position B-C, all 16K of on-board RAM is accessed by the system; addresses for system use are established by jumpers 180-173 (1st 16K), 180-175 (2nd 16K), etc.

611-2 Figure 2-1. Jumper Configuration for Multibus Access of On-Board RAM (16-Bit Address System)



Dumper W5 position K-G selects page 6 of lower 524K.

(3.) Jumper W4 position B-A selects 8K access of page 6.

(4.) Jumper 180-173 selects 2nd 8K segment of page 6.

Thus, the upper 8K of on-board memory is assigned system addresses 62000 thru 63FFF. In this example, this same 8K of memory has on-board 8085A addresses A000 thru BFFF.

8000

16-BIT ON-BOARD ADDRESS LINES

Note that for 8K access, the 1st, 3rd, 5th, and 7th 8K bytes access the lower 8K of on-board RAM. The 2nd, 4th, 6th, and 8th 8K bytes access the upper 8K of on-board RAM. With jumper W4 in position B-C, all 16K of on-board RAM is accessed by the system; addresses for system use are established by the jumpers 180-173 (1st 16K), 180-175 (2nd 16K), etc.

611-3 Figure 2-2. Jumper Configuration for Multibus Access of On-Board RAM (20-Bit Address System)

Interrupt Request (From)			. II	nterrupt Request (To)	
Source	Signal	Post	Device	Signal	Post
Multibus	INT0/ (1) INT1/ (1) INT2/ (1) INT3/ (1) INT4/ (1) INT5/ (1) INT6/ (1) INT7/ (1)	148 147 152 151 150 149 146 136	8259A PIC (7)	IR0 IR1 IR2 IR3 IR4 IR5 IR6 IR7	133 132 131 130 129 128 127 126
External Via J2-50	EXT INTR1/ (1)	144	8085A CPU	TRAP (8)	137
External Via J1-50	EXT INTR0/ (1)	125		RST 7.5 (9) RST 6.5 (10)	138 139
Power Fail Logic Via P2-19	PFI/ (1)(2)	134		RST 5.5 (10) INTR (11)	140 None
Failsafe Timer	BUS TIME OUT/ (3)	122			
8255A PPI Port A (Port E8) Port B (Port E9) 8251A USART	55PAI (1)(4) 55PBI (1)(4) 51TXR (1)	117 118 142	Multibus	INTO/ (12) INT1/ (12) INT2/ (12) INT3/ (12) INT3/ (12)	181 182 183 184 197
	51RXR (1)	143		INT4/ (12)	188
8253 PIT Counter 0 Out Counter 1 Out	8041/8741 EVENT CLK (1) COUNT OUT (1)(5)	141 123		INT6/ (12) INT7/ (12) ADR10/ (12)	189 190 186
8259A PIC	INTR (6)	None			
8041/8741A UPI	41INTR (1)	124	119 -		
8255A PPI	INTROUT (13)	185	120 ⊶		

Table 2-9. Priority Interrupt Jumper Matrix

NOTES:

- (1) Signal is positive-true at associated jumper post.
- (2) Disconnect 137-145 and connect 134-137 (8085A TRAP).
- (3) Disconnect 123-138 and connect 122-138 (8085A RST 7.5); jumper 115-116 must also be connected. See note (5).
- (4) Used primarily in strobed I/O applications.
- (5) Default jumper 123-138 connects signal to 8085A RST 7.5.
- (6) Applied directly to 8085A INTR input.
- (7) May be programmed for either edge-sensitive or level-sensitive input; IR lines are not individually programmable.
- (8) Default jumper 137-145 disables (grounds) input. TRAP is highest priority, non-maskable, and is both level and edge sensitive.
- (9) Default jumper 123-138 connects input to COUNT OUT. RST 7.5 is second highest priority and is edge sensitive only.
- (10) RST 6.5 and RST 5.5, respectively, are third and fourth highest priority. Both inputs are level sensitive.
- (11) INTR is connected directly to INTR output of 8259A PIC.
- (12) To system via Multibus; requires ground-true signal.
- (13) Signal is ground-true at associated jumper post.
- (14) One two-input OR gate is provided in interrupt matrix.

2-21. 8251A PORT CONFIGURATION

Table 2-10 lists the signals, signal functions, and the jumpers required (if necessary) to input or output a particular signal to or from the serial I/O port (Intel 8251A Programmable Communication Interface).

2-22. 8255A PORT CONFIGURATION

Table 2-11 lists the jumper configuration for three parallel I/O ports. Note that each of the three ports (E8, E9, and EA) can be configured in a variety of ways to suit the

individual requirement. Recommended line drivers and I/O terminators for user applications are listed in table 2-1.

2-23. 8041/8741A PORT CONFIGURATION

The optional Intel 8041/8741A Universal Peripheral Interface can be programmed and, hence, configured to perform serial or parallel I/O functions. Refer to figure 5-2 sheet 5 for jumper details. Applications for the 8041/8741 are presented in the *Intel UPI-41 User's Manual*, Order No. 9800504.

Pin ¹	Signal	Function	Jumper In	Jumper Out
2	PROTECTIVE GND	Protective Ground	69–70	
3	TRANS SIG ELE TIMING (IN)	8251A TXC in (Note 2)	56–57	*55–56
4	TRANSMITTED DATA	8251A RXD in or 8251A TXD out	*82–83 81–82	 *82_83, *80_81
5	SEC REC DATA	8255A STXD out³ or 8741A RS232 out⁴	76–79, 73–71 76–79, 73–74	-
6	RECEIVED DATA	8251A TXD out or 8251A RXD in	*80–81 80–83	 *80-81, *82-83
7	REC SIG ELE TIMING	8251A RXC in (Note 2)	59–60	*58–59
8	REQUEST TO SEND	8251A CTS in (Note 5)	—	
—	(Note 5)	8251A RTS out to CTS in	67–68	
10	CLEAR TO SEND	8251A RTS out (Note 5)	—	—
	(Note 5)	8251A RTS out to CTS in	67–68	—
12	DATA SET READY	8251A DTR out		—
13	DATA TERMINAL RDY	8251A DSR in	—	—
14	GND	Ground		_
16	REG LINE SIG DET	8255A Carrier Detect in ³	_	_
17	RING INDICATOR	8255A Ring Indicator in ³	—	—
19	–12V	-12V out	64–65	
21	TRANS SIG ELE TIMING (OUT)	Same as 8251A TXC in	76–75, 72–73	_
22	+12V	+12V out	6366	
23	+5V	+5V out	61–62	-
25	GND	Ground	—	—
26	SEC CLEAR TO SEND	8255A STXD in ³	76–77	_

Table 2-10. Connector 15 Fin Assignments vs. Jumper Conneura	Table 2-10.	Connector J3	Pin Ass	ignments Vs.	Jumper	Configurat
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NOTES:

1. All odd-numbered pins (1, 3, 5, ... 25) are on component side of the board. Pin 1 is the right-most pin when viewed from the component side of the board with the extractors at the top.

 Default jumpers *55–56 and *58–59 connect 8253 BAUD RATE CLK to 8251 TXC and RXC inputs, respectively. See Timer Input Frequency (Counter 2) in table 2-4.

3. Optional jumper-selected output function for Intel 8255 Programmable Peripheral Interface. Refer to figure 5-2 sheet 4.

4. Optional jumper-selected output function for 8041/8741 Universal Peripheral Interface. Refer to figure 5-2 sheet 5.

5. Jumper 67-68 connects 8251A RTS output back to CTS input for those applications without CTS capability.

*Default jumpers connected at the factory.

				Jumper	Configuration		· · · · · · · · · · · · · · · · · · ·
Port	Mode	Terminator (T)	Delete	Add	Effect	Port	Restrictions
E8	0 Input	8226: A1,A2	None	*78	8226 = input enabled.	E9	None; can be in Mode 0 or 1, input or output.
						EA	None; can be in Mode 0, input or output, unless Port E9 is in Mode 1.
E8	0 Output (latched)	8226: A1,A2	*7–8	4–8	8226 = output enabled.	E9	None; can be in Mode 0 or 1, input or output.
						EA	None; can be in Mode 0, input or output, unless Port E9 is in Mode 1.
E8	1 Input (strobed)	8226: A1,A2 T: A3		*7–8	8226 = input enabled.	E9	None; can be in Mode 0 or 1, input or output.
		D: A4		*21–22	Connects J1–26 to STB _A / input.	EA	Port EA bits perform the following:
			*15–16 and *17–18	15–17	Connects IBF _A output to J1–18.		• Bits 0,1,2 — Control for Port E9 if in Mode 1.
							 Bit 3 — Port E8 Interrupt (55PAI) to interrupt jumper matrix.
]							 Bit 4 — Port E8 Strobe (STB/) input.
							 Bit 5 — Port E8 Input Buffer Full (IBF) output.
							 Bits 6,7 — Port EA input or output (both must be in same direction).
E8	1 Output (latched)	8226: A1,A2 T: A3 D: A4	*7–8	4–8	8226: output enabled.	E9	None; can be in Mode 0 or 1, input or output.
				*13–14	Connects J1-30 to ACK₄/ input.	EA	Port EA bits perform the following:
			*9–10 and *15–16	9–15	Connects OBF_A output to J1–18.		• Bits 0,1,2 — Control for Port E9 if in Mode 1.
							 Bit 3 — Port E8 Interrupt (55PAI) to interrupt jumper matrix.
						į	 Bits 4,5 — Input or output (both must be in same direction).
							 Bit 6 — Port E8 Acknowl- edge (ACK/) input.
							 Bit 7 — Port E8 Output Buffer Full (OBF/) output.

			Jumper Configuration				
Port	Mode	Terminator (T)	Delete	Add	Effect	Port	Restrictions
E8	2 (hidiractional)	8226: A1,A2	*7–8	8–13	Allows ACK _A / input to	E9	None.
	(bidirectional)	D: A4			direction.	EA	Port EA bits perform the following:
				*21–22	Connects J1–26 to STB _A / input.		 Bit 0 — Can only be used for jumper option (see figure 5–2 zone 4ZC4).
			*17–18 and *25–26	17–25	Connects IBF _A output to J1–24.		 Bits 1,2 — Can be used for input or output if Port E9 is in Mode 0.
				*13–14	Connects J1–30 to ACK _A / input.		 Bit 3 — Port E8 Interrupt (55PAI) to interrupt jumper matrix.
			*9–10 and *15–16	9–15	Connects OBF _A / output to J1–18.		 Bit 4 — Port E8 Strobe (STB/) input.
	ļ						 Bit 5 — Port E8 Input Buffer Full (IBF) output.
							 Bit 6 — Port E8 Acknowl- edge (ACK/) input.
							Bit 7— Port E8 Output Buffer Full (OBF/) output.
E9	0 Input	T: A5,A6	None	None		E8	None.
						EA	None; Port EA can be in Mode 0, input or output, if Port E8 is also in Mode 0.
E9	0 Output	D: A5,A6	None	None		E8	None.
	(latched)					EA	None; Port EA can be in Mode 0, input or output, if Port E8 is also in Mode 0.
E9	1 Input	T: A3,A5,A6		*23–24	Connects IBF _B output	E8	None.
	(strobed)	D: A4			to J1–22.	EA	Port EA bits perform the following:
			*19–20 and *9–10	10–20	Connects J1–32 to STB _B / input.		 Bit 0 — Port E9 Interrupt (55PBI) to interrupt jumper matrix.
		·			i i		 Bit 1 — Port E9 Input Buffer Full (IBF) output.
							 Bit 2 — Port E9 Strobe (STB/) input.
							• Bit 3 — If Port E8 is in Mode 0, bit 3 can be input or output. Otherwise, bit 3 is reserved.
							Bits 4,5 — Depends on Port E8 mode.
							 Bits 6,7 — Input or output (both must be in same direction).

Table 2-11.	8255A	Port	Configuration	Jumpers	(Continued)
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				Jumper	Configuration		
Port	Mode	Terminator (D)/	Delete	Add	Effect	Port	Restrictions
E9	1 Output	T: A3	*25–26	*23–24	Connects OBF _B / output	E8	None.
	(latched)	D: A4,A5,A6			J1–22.	EA	Port EA bits perform the following:
			*19–20 and *9–10	10–20	Connects J1–32 to ACK _B / input.		 Bit 0 — Port E9 Interrupt (55PBI) to interrupt jumper matrix.
							 Bit 1 — Port E9 Output Buffer Full (OBF/) output.
							 Bit 2 — Port E9 Acknowl- edge (ACK/) input.
							 Bit 3 — If Port E8 is in Mode 0, bit 3 can be input or output. Otherwise, bit 3 is reserved.
							 Bits 4,5 — Input or output (both must be in same direction).
						:	 Bits 6,7 — Depends on Port E8 mode.
EA (upper)	0 Input	Т: АЗ	None	*21–22 *17–18	Connects bit 4 to J1–26. Connects bit 5 to J1–28.	E8	Port E8 must be in Mode 0 for all four bits to be available.
				*9–10	Connects bit 7 to J1–32.	E9	Port E9 must be in Mode 0 for all four bits to be available.
EA (lower)	0 Input	T: A4	None	*25–26 *23–24 *19–20	Connects bit 0 to J1–24. Connects bit 1 to J1–22. Connects bit 2 to J1–20	E8	Port E8 must be in Mode 0 for all four bits to be available.
				*15–16	Connects bit 3 to J1–18.	E9	Port E9 must be in Mode 0 for all four bits to be available.
EA (upper)	0 Output (latched)	D: A3	None	Same as Mode 0	s for Port EA (upper) Input.	EA8	Same as for Port EA (upper) Mode 0 Input.
EA (lower)	0 Output (latched)	D: A4	None	Same as Mode 0	s for Port EA (lower) Input.	E9	Same as for Port EA (lower) Mode 0 Input.
*Default	iumper conne	cted at the factory.					

Table	2-11.	8255A	Port	Configuration	Jumpers	(Continued)
				0	-	• •

2-24. MULTIBUS CONFIGURATION

For systems applications, the iSBC 80/30 is designed for installation in a standard Intel iSBC 604/614 Modular Backplane and Cardcage. (Refer to table 2-1 items 1 and 2.) Alternatively, the iSBC 80/30 can be interfaced to a user-designed system backplane by means of an 86-pin connector. (Refer to table 2-1 item 3.) Multibus signal characteristics and methods of implementing a serial or parallel priority resolution scheme for resolving bus contention in a multiple bus master system are described in the following paragraphs.



Always turn off the system power supply before installing the board in or removing the board

from the backplane. Failure to observe this precaution can cause damage to the board.

2-25. SIGNAL CHARACTERISTICS

As shown in figure 1-1, connector P1 interfaces the iSBC 80/30 to the Multibus. Connector P1 pin assignments are listed in table 2-12 and descriptions of the signal functions are provided in table 2-13.

The dc characteristics of the iSBC 80/30 bus interface signals are provided in table 2-14. The ac characteristics of the iSBC 80/30 when operating in the master mode and slave mode are provided in tables 2-15 and 2-16, respectively. Bus exchange timing diagrams are provided in figures 2-3 and 2-4.

Pin*	Signal	Function	Pin*	Signal	Function
1	GND		44	ADRF/	
2	GND	} Ground	45	ADRC/	
3	+5V		46	ADRD/	
4	+5V		47	ADRA/	
5	+5V	4	48	ADRB/	
6	+5V	Dever in set	49	ADR8/	
7	+12V	> Power input	50	ADR9/	
8	+12V		51	ADR6/	Address bus
9	-5V		52	ADR7/	
10	-5V)	53	ADR4/	
11	GND	} Ground	54	ADR5/	
12	GND) Cround	55	ADR2/	
13	BCLK/	Bus Clock	56	ADR3/	
14	INIT/	System Initialize	57	ADR0/	
15	BPRN/	Bus Priority In	58	ADR1/)
16	BPRO/	Bus Priority Out	59		
17	BUSY/	Bus Busy	60		
18	BREQ/	Bus Request	61		
19	MRDC/	Memory Read Command	62		
20	MWTC/	Memory Write Command	63		с
21	IORC/	I/O Read Command	64		
22	IOWC/	I/O Write Command	65		
23	XACK/	Transfer Acknowledge	66		
24	INH1/	Inhibit RAM	67	DAT6/	
25			68	DAT7/	
26			69	DAT4/	
27			70	DAT5/	Data Bus
28	ADR10/	Extended Address bus	71	DAT2/	Data Bus
29	CBRQ/	Common Bus Request	72	DAT3/	
30	ADR11/	Extended Address bus	73	DAT0/	
31	CCLK/	Constant Clock	74	DAT1/)
32	ADR12/	Extended Address bus	75	GND	
33			76	GND	Ground
34	ADB13/	Extended Address bus	70	GND	,
35	INT6/	Interrupt request on level 6	78		
36	INT7/	Interrupt request on level 7	79	+12V	
37	INT4/	Interrupt request on level 4	80	-12V	
38	INT5/	Interrupt request on level 5	81	+5V	
00					Power input
39	IN [2/	Interrupt request on level 2	82	+5V	
40	IN [3/	Interrupt request on level 3	83	+5V	J I
41	IN FO/	Interrupt request on level 0	84	+5V	
42	IN I 1/	Interrupt request on level 1	85	GND	Ground
43	ADRE/		86	GND)

Table 2-12. Multibus Connector P1 Pin Assignments

*All odd-numbered pins (1, 3, 5... 85) are on component side of the board. Pin 1 is the left-most pin when viewed from the component side of the board with the extractors at the top. All unassigned pins are reserved.

Table 2-13. Multibus Signal Functions

Signal	Functional Description
ADR0/-ADRF/ ADR10/-ADR13/	Address. These 20 lines transmit the address of the memory location or I/O port to be accessed. ADRF/ is the most-significant bit except where ADR10/ through ADR13/ are used. ADR10/ through ADR13/ are transmitted only by those bus masters capable of addressing beyond 65K of memory. In this case, ADR13/ is the most-significant bit.
BCLK/	<i>Bus Clock.</i> Used to synchronize the bus contention logic on all bus masters. When generated by the iSBC 80/30, BCLK/ has a period of 108 nanoseconds (9.22 MHz) with a 35-65 percent duty cycle.
BPRN/	Bus Priority In. Indicates to a particular bus master that no higher priority bus master is requesting use of the bus. BPRN/ is synchronized with BCLK/.
BPRO/	Bus Priority Out. In serial (daisy chain) priority resolution schemes, BPRO/ must be connected to the BPRN/ input of the bus master with the next lower bus priority.
BREQ/	Bus Request. In parallel priority resolution schemes, BREQ/ indicates that a particular bus master requires control of the bus for one or more data transfers. BREQ/ is synchronized with BCLK/.
BUSY/	Bus Busy. Indicates that the bus is in use and prevents all other bus masters from gaining control of the bus. BUSY/ is synchronized with BCLK/.
CBRQ/	Common Bus Request. Indicates that a bus master wishes control of the bus but does not presently have control. As soon as control of the bus is obtained, the requesting bus controller raises the CBRQ/ signal.
CCLK/	Constant Clock. Provides a clock signal of constant frequency for use by other system modules. When generated by the iSBC 80/30, CCLK/ has a period of 108 nanoseconds (9.22 MHz) with a 35-65 percent duty cycle.
DAT0/-DAT7/	Data. These eight bidirectional data lines transmit and receive data to and from the addressed memory location or I/O port. DAT7/ is the most-significant bit.
INH1/	Inhibit Ram. Prevents system access to on-board RAM; allows system addresses to overlay on-board RAM for certain applications. This signal has no effect on on-board access of on-board RAM.
INIT/	Initialization. Resets the entire system to a known internal state.
INTO/-INT7/	Interrupt. These eight lines are for inputting interrupt requests to the iSBC 80/30. INTO/ has the highest priority; INT7/ has the lowest priority.
IORC/	I/O Read Command. Indicates that the address of an I/O port is on the Multibus address lines and that the output of that port is to be read (placed) onto the Multibus data lines.
IOWC/	I/O Write Command. Indicates that the address of an I/O port is on the Multibus address lines and that the contents on the Multibus data lines are to be accepted by the addressed port.
MRDC/	Memory Read Command. Indicates that the address of a memory location is on the Multibus address lines and that the contents of that location are to be read (placed) on the Multibus data lines.
MWTC/	Memory Write Command. Indicates that the address of a memory location is on the Multibus address lines and that the contents on the Multibus data lines are to be written into that location.
XACK/	Transfer Acknowledge. Indicates that the addressed memory location has completed the specified read or write operation. That is, data has been placed onto or accepted from the Multibus data lines.

Signals	Symbol	Parameter Description	Test Conditions	Min.	Max.	Units
ADR0/-ADRC/	V _{OL}	Output Low Voltage	I _{OL} = 50 mA		0.5	v
	V _{OH}	Output High Voltage	l _{OH} = -10 mA	2.4		v
	VIL	Input Low Voltage			0.8	v
	VIH	Input High Voltage		2.0		v
	I _{IL}	Input Current at Low V	V _{IN} = 0.45V		-0.25	mA
	liH	Input Current at High V	V _{IN} = 5.25V		80	μA
	ILH	Output Leakage High	$V_0 = 5.25V$		200	μΑ
	, ILL	Output Leakage Low	$V_0 = 0.45V$		200	μA
	*CL	Capacitive Load			18	pF
ADRD/-ADRF	V _{OL}	Output Low Voltage	l _{OL} = 50 mA		0.5	v
	V _{OH}	Output High Voltage	l _{OH} = 10 mA	2.4		v
	VIL	Input Low Voltage			0.8	v
	VIH	Input High Voltage		2.0		v
	Ι _{ΙL}	Input Current at Low V	V _{IN} = 0.45V		-0.5	mA
	ιн	Input Current at High V	V _{IN} = 5.25V		200	μA
	*CL	Capacitive Load			25	pF
ADR10/-ADR13/	V _{IL}	Input Low Voltage		-	0.85	v
	[·] V _{IH}	Input High Voltage		2.0		v
	կլ	Input Current at Low V	V _{IN} = 0.45V		-0.25	mA
	lın	Input Current at High V	V _{IN} = 5.25V		10	μΑ
	*CL	Capacitive Load			18	pF
BCLK/	VOL	Output Low Voltage	l _{OL} = 59.5 mA		0.5	v
	V _{OH}	Output High Voltage	I _{OH} = –3 mA	2.7		v
	VIL	Input Low Voltage			0.8	v
	VIH	Input High Voltage		2.0		v
	IL.	Input Current at Low V	V _{IN} = 0.45V		-0.5	mA
	lін	Input Current at High V	V _{IN} = 5.25V		100	μA
	*CL	Capacitive Load			15	pF
BPRN/	VIL	Input Low Voltage	· · · · · · · · · · · · · · · · · · ·	1	0.8	v
	VIH	Input High Voltage		2.0		· v
	 I _{IL}	Input Current at Low V	V _{IN} = 0.4V		-1.6	mA
	 Iн	Input Current at High V	$V_{IN} = 2.4V$		40	μA
	*CL	Capacitive Load			18	pF
<u> </u>	I	1	L,			L

Table 2-14. iSBC 80/30 DC Characteristics

Signals	Symbol	Parameter Description	Test Conditions	Min.	Max.	Units
BPRO/	V _{OL}	Output Low Voltage	I _{OL} = 3.2 mA		0.45	v
	V _{OH}	Output High Voltage	l _{OH} = -400 μA	2.4		v
	ILH	Output Leakage High	V _O = 5.25V]	100	μA
	ILL	Output Leakage Low	V _O = 0.45V	1	-100	μA
	*CL	Capacitive Load			15	pF
BREQ/	V _{OL}	Output Low Voltage	I _{OL} = 20 mA		0.45	v
	V _{OH}	Output High Voltage	l _{OH} = -400 μA	2.4		v
	VIL	Input Low Voltage	V ₀ = 5.25V		100	μΑ
	VIH	Input High Voltage	V _O = 0.45V		-100	μΑ
	*CL	Capacitive Load			15	pF
BUSY/, CBRQ/,	V _{OL}	Output Low Voltage	l _{OL} = 20 mA		0.45	v
(OPEN COLLECTOR)	*CL	Capacitive Load			20	pF
CCLK/	V _{OL}	Output Low Voltage	l _{OL} = 60 mA		0.5	v
	V _{OH}	Output High Voltage	I _{OH} = –3 mA	2.7		v
	*CL	Capacitive Load			15	pF
DAT0/-DAT7/	V _{OL}	Output Low Voltage	l _{OL} = 50 mA		0.6	v
	V _{OH}	Output High Voltage	I _{OH} = -10 mA	2.4		v
	VIL	Input Low Voltage			0.95	v
	V _{IH}	Input High Voltage		2.0		v
	կլ	Input Current at Low V	$V_{IN} = 0.45V$		-0.25	mA
	Іцн	Output Leakage High	V ₀ = 5.25V		100	μA
	الد	Output Leakage Low	V _O = 0.45V		100	μA
	*CL	Capacitive Load			18	pF
INH1/	VIL	Input Low Voltage			0.8	v
	VIH	Input High Voltage		2.0		V
	կլ	Input Current at Low V	$V_{IN} = 0.5V$		-2.0	mA
	lін	Input Current at High V	$V_{IN} = 2.7V$		50	μA
	*CL	Capacitive Load			18	pF
INIT/ (SYSTEM RESET)	V _{OL}	Output Low Voltage	$I_{OL} = 46 \text{ mA}$		0.4	v
	V _{OH}	Output High Voltage	OPEN COLLECTOR			
	VIL	Input Low Voltage			0.8	v
*Capacitive load values are a	approximations.		·	I	L	L

Table 2-14. iSBC 80/30 DC Characteristics (Continued)

Signals	Symbol	Parameter Description	Test Conditions	Min.	Max.	Units
INIT/	VIH	Input High Voltage		2.0		V
(Continued)	Ι _{ΙL}	Input Current at Low V	V _{IN} = 0.5V		-2.0	mA
	lін	Input Current at High V	V _{IN} = 2.7V		90	μA
	*CL	Capacitive Load			15	pF
INTO/-INT7	VIL	Input Low Voltage			0.8	V
	VIH	Input High Voltage		2.0		v
	կլ	Input Current at Low V	$V_{IN} = 0.4V$		-0.4	mA
	Ιн	Input Current at High V	V _{IN} = 2.7V		20	μA
	*CL	Capacitive Load			15	pF
IORC/, IOWC/	V _{OL}	Output Low Voltage	I _{OL} = 32 mA		0.45	v
	V _{OH}	Output High Voltage	I _{OH} = -2 mA	2.4		v
	Lн	Output Leakage High	V _O = 5.25V		100	μΑ
	ارر	Output Leakage Low	V _O = 0.45V		-100	μΑ
	*CL	Capacitive Load			15	pF
MRDC/, MWTC/	V _{OL}	Output Low Voltage	I _{OL} = 32 mA		0.45	V
	V _{OH}	Output High Voltage	I _{OH} = -2 mA	2.4		v
	VIL	Input Low Voltage		i	0.8	v
	V _{IH}	Input High Voltage	1	2.0		v
	կլ	Input Current at Low V	V _{IN} = 0.4V		-0.5	mA
	l _{ін}	Input Current at High V	V _{IN} = 2.7V		150	μA
	*CL	Capacitive Load			15	pF
XACK/	Vol	Output Low Voltage	I _{OL} = 32 mA		0.4	V
	V _{OH}	Output High Voltage	I _{OH} = -5 mA	2.4		v
	VIL	Input Low Voltage		1	0.8	v
	ViH	Input High Voltage		2.0		v
	Ι _{ΙL}	Input Current at Low V	V _{IN} = 0.4V		-0.95	mA
	Чн	Input Current at High V	V _{IN} = 2.4V		60	μA
	*CL	Capacitive Load			25	pF
*Capacitive load values are approximations.						

Table 2-14. iSBC 80/30 DC Characteristics (Continued)

Overall Read Write	
ParameterMin.Max.Min.Max.Min.Max.(ns)(ns)(ns)(ns)(ns)(ns)	Remarks
t _{AS} 50 50 50 Address setup time to command	
t _{AH} 50 50 50 Address hold time from command	
t _{DS} 50 50 Data setup to command	
tDHW 50 50 Data hold time from command	
t _{CY} 358 365 CPU cycle time	
tCMDB 670 Read command width	With 1 wait state
t _{CMDW} 390 Write command width	With 1 wait state
t _{CSWR} 575 Read-to-write command separation	In override mode
t _{CSRR} 465 Read-to-read command separation	In override mode
t _{CSWW} 575 Write-to-write command separation	In override mode
t _{CSRW} 465 Write-to-read command separation	In override mode
txACK1 –195 Read command to XACK 1st sample poir	In override mode
txACK2 –509 Write command to XACK 1st sample poin	t In override mode
t _{SAM} 350 375 Time between XACK samples	In override mode
tACKRD 75 AACK to valid read data	When AACK is used
tACKWT 175 AACK to write command inactive	When AACK is used
t _{DHR} 0 Read data hold time	
t _{DXL} –75 Read data setup to XACK	
t _{XKH} 0 0 0 XACK hold time	
t _{BW} 35 Bus clock low or high interval	Supplied by system
t _{BS} 23 BPRN to BCLK setup time	
t _{DBY} 55 BCLK to BUSY delay	
t _{NOD} 30 BPRN to BPRO delay	
tBCY 108 109 Bus clock period (BCLK)	From iSBC 80/30 when terminated
t _{BW} 35 74 Bus clock low or high interval	From iSBC 80/30 when terminated
tINIT 3000 Initialization width	After all voltages have stabilized

	Table 2-15.	iSBC 80/30	AC Characteristics	(Master	Mode)
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Table 2-10. ISBC 80/30 AC Characteristics (Slave Mode	Table 2-16.	iSBC 80/30 A	AC	Characteristics	(Slave	Mode)
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Parameter	Minimum (ns)	Maximum (ns)	Description	Remarks
t _{AS}	50		Address setup to command	From address to command
t _{DS}	-200		Write data setup to command	
t _{OB1}		1500	On-board memory cycle delay	No refresh
t _{ACK}	480	720	Command to XACK	
t _{CMD}	720		Command width	
t _{AH}	0		Address hold time	
tohw	0		Write data hold time	
t _{DHR}	0		Read data hold time	
tхтн	0	45	Acknowledge hold time	Acknowledge turnoff delay
*tacc		645	Access time to read data	
ŧн	50		Inhibit time from command trailing edge	Blocks AACK if $t_{IS} > t_{IS}$ min.
tipw	100		Inhibit pulse width	
*tcy	680	920	Minimum cycle time	$t_{CY} = t_{ACK} + t_{SEP}$
tOB2		1865	On board memory cycle delay	Refresh delaying SACK
t _{RD}	535	555	Refresh delay time	
t _{DXL}	25		Read data setup to XACK	
t _{SEP}	200	· ·	Command separation	
tis	50		Inhibit setup to command	Blocks RAM cycle and tACK
	L	L	L	1

*When an asynchronous refresh cycle occurs, t_{RD} is added to these parameters; when on-board memory cycle occurs, t_{OB1} is also added.



Figure 2-3. Bus Exchange Timing (Master Mode)

611-4



Figure 2-4. Bus Exchange Timing (Slave Mode)

2-26. SERIAL PRIORITY RESOLUTION

In a multiple bus master system, bus contention can be resolved in an iSBC 604 Modular Backplane and Cardcage by implementing a serial priority resolution scheme as shown in figure 2-5. Due to the propagation delay of the BPRO/ signal path, this scheme is limited to a maximum of three bus masters capable of acquiring and controlling the Multibus. In the configuration shown in figure 2-5, the bus master installed in slot J2 has the highest priority and is able to acquire control of the Multibus at any time because its BPRN/input is always enabled (tied to ground) through jumpers B and N on the backplane. (See figure 5-3.)

If the bus master in slot J2 desires control of the Multibus, it drives its BPRO/ output high and inhibits the BPRN/ input to all lower-priority bus masters. When finished using the Multibus, the J2 bus master pulls its BPRO/ output low and gives the J3 bus master the opportunity to take control of the Multibus. If the J3 bus master does not


Figure 2-5. Serial Priority Resolution Scheme

desire to control the Multibus at this time, it pulls its BPRO/ output low and gives the lowest priority bus master in slot J4 the opportunity to assume control of the Multibus.

The serial priority scheme can be implemented in a userdesigned system bus if the chaining of BPRO/ and BPRN/ signals are wired as shown in figure 5-3.

2-27. PARALLEL PRIORITY RESOLUTION

A parallel priority resolution scheme allows up to 16 bus masters to acquire and control the Multibus. Figure 2-6 illustrates one method of implementing such a scheme for resolving bus contention in a system containing eight bus masters installed in an iSBC 604/614. Notice that the two highest and two lowest priority bus masters are shown installed in the iSBC 604.

In the scheme shown in figure 2-6, the priority encoder is a Texas Instruments 74148 and the priority decoder is an Intel 8205. Input connections to the priority encoder determine the bus priority, with input 7 having the highest priority and input 0 having the lowest priority. Here, the J3 bus master has the highest priority and the J5 bus master has the lowest priority.

IMPORTANT: In a parallel priority resolution scheme, the BPRO/ output must be disabled on all bus masters. On the iSBC 80/30, disable the BPRO/ output signal by removing jumper 169-170. If a similar jumper cannot be removed on the other bus masters, either clip the IC pin that supplies the BPRO/ output signal to the Multibus or cut the signal trace.

2-28. POWER FAIL/MEMORY PROTECT CONFIGURATION

A mating connector must be installed in the iSBC 604/604 Modular Cardcage and Backplane to accommodate auxiliary connector P2. (Refer to figure 1-1.) Table 2-2 lists

484-2



some 60-pin connectors that can be used for this purpose; flat crimp, solder, and wirewrap connector types are listed. Table 2-17 correlates the signals and pin numbers

Procure the appropriate mating connector for P2 and secure it in place as follows:

on the connector.

- a. Position holes in P2 mating connector over mounting holes that are in line with corresponding P1 mating connector.
- b. From top of connector, insert two 0.5-inch #4-40 pan head screws down through connector and mounting holes.
- c. Install a flat washer, lock washer, and star-type nut on each screw; then tighten the nuts.

When the mating connector for P2 is in place, wire the power fail signals to the appropriate pins of the connector as listed in table 2-17. (The dc characteristics of the auxiliary signals are given in table 2-18.) In a typical system, these signals would be wired as follows:

a. Connect auxiliary signal common and returns for +5V, -5V, and +12V backup batteries to P2 pins 1 and 2.

- b. Connect +5V battery input to P2 pins 3 and 4; -5V battery input to P2 pins 7 and 8; and +12V battery input to P2 pins 11 and 12. Remove jumpers W7, W8, and W9.
- c. Connect PFS/input to P2 pin 17 and MEM PROT/ input to P2 pin 20.
- d. Connect PFI/ input to P2 pin 19; this signal is inverted and applied to the priority interrupt matrix. To assign the PFI/ input as the highest priority interrupt (8085A TRAP), remove jumper 137-145 and connect jumper 134-137.
- e. Connect HLT/ output at P2 pin 28 to external HALT indicator, which is typically a light-emitting diode (LED) mounted on the system enclosure.
- f. Connect BTMO output at P2 pin 34 to external TIME OUT indicator, which is typically a LED mounted on the system enclosure.
- g. Connect AUX RESET/ input to P2 pin 38. This signal is usually supplied by a momentary closure switch mounted on the system enclosure.
- h. Connect WAIT/ output at P2 pin 32 to external WAIT indicator, which is typically a LED mounted on the system enclosure.

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Pin*	Signal	Definition
1	GND	
2	GND	Auxiliary common
3	+5V AUX	
4	+5V AUX	
7	-5V AUX	Annillant beeling better constru
8	5V AUX	Auxiliary backup battery supply
11	+12V AUX	
12	+12 AUX)
17	PFS/	Power Fail Status. This externally supplied signal is applied to the SID input of the 8085A microprocessor to allow the program to check the current status of the system power fail circuit. The status is checked by periodically executing a RIM instruction.
19	PFI/	Power Fail Interrupt This externally supplied signal is applied to the priority interrupt matrix. This signal should normally be jumpered to the 8085A microprocessor TRAP input.
20	MEM PROT/	Memory Protect. This externally supplied signal prevents access to RAM during battery backup operation.
28	HLT/	Halt. This output signal indicates that the 8085A microprocessor is halted.
- 32	WAIT/	Wait. This output signal, which is the same as the CPU ALE signal, indicates that the 8085A microprocessor is either halted or in a wait state.
34	BTMO	Bus Time Out. This output signal indicates that the 8085A microprocessor has either halted or is hung up in a wait state (i.e., waiting for an acknowledge signal in response to the previous I/O or Memory Command).
38	AUX RESET/	Auxiliary Reset. This externally supplied signal initiates a pseudo power-up sequence; i.e., initializes the board and resets the entire system to a known internal state.
*All odd-i compon	numbered pins (1, 3, 5 ent side of the board wit	59) are on component side of the board. Pin 1 is the left-most pin when viewed from the the the extractors at the top.

Table 2-17.	Auxiliary	Connector	P2 Pin	Assignments
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[•] Table 2-18. Auxiliary Signal (Connector P2) DC Characteristics

Signals	Symbol	Parameter Description	Test Conditions	Min.	Max.	Units
PFS/, MEM PROT/	Vլլ Vլн կլ կլ Cլ	Input Low Voltage Input High Voltage Input Current at Low V Input Current at High V Capacitive Load	V _{IN} = 0.4V V _{IN} = 2.4V	2.4	0.8 2 50 10	V V mA μ pF
AUX RESET/	Vil Viн կi Сi	Input Low Voltage Input High Voltage Input Current at Low V Input Current at High V Capacitive Load	V _{IN} = 0.45V V _{IN} = 5.25V	2.6	0.8 0.25 10 10	V V mA μA μF
WAIT/	V _{OL} V _{OH} C _L	Output Low Voltage Output High Voltage Capacitive Load	$I_{OL} = 17 \text{ mA}$ $I_{OH} = -950 \mu \text{A}$	2.7	0.5 15	V V pF
PFI/	VIL VIH III III CL	Input Low Voltage Input High Voltage Input Current at Low V Input Current at High V Capacitive Load	$V_{IN} = 0.40V$ $V_{IN} = 2.7V$	2.0	0.80 -0.4 20 15	V V mA μA pF

2-29. PARALLEL I/O CABLING

Parallel I/O ports E8, E9, and EA are controlled by the Intel 8255A Programmable Peripheral Interface (PPI) and interfaced via edge connector J1. Parallel I/O ports 1 and 2 are controlled by the optional Intel 8041/8741 Universal Peripheral Interface and interfaced via edge connector J2. (Refer to figure 1-1.) Pin assignments for J1 and J2 are listed in tables 2-19 and 2-20, respectively; dc characteristics of the parallel I/O signals are given in table 2-21. Table 2-2 lists some 50-pin edge connectors that can be used for interface to J1 and J2; flat crimp, solder, and wirewrap connector types are listed.

The transmission path from the I/O source to the iSBC 80/30 should be limited to 3 meters (10 feet) maximum.

The following bulk cable types (or equivalent) are recommended for interfacing with the parallel I/O ports:

- a. Cable, flat, 50-conductor, 3M 3306-50.
- b. Cable, flat, 50-conductor (with ground plane), 3M 3380-50.
- c. Cable, woven, 25-pair, 3M 3321-25.

An Intel iSBC 956 Cable Set, consisting of two cable assemblies, is recommended for parallel I/O interfacing.

 Table 2-19.
 Connector J1 Pin Assignments

Pin*	Function	Pin*	Function
1 3 5 7 9 11 13 15	Ground	2 4 8 10 12 14 16	Port E9 bit 7 Port E9 bit 6 Port E9 bit 5 Port E9 bit 4 Port E9 bit 3 Port E9 bit 2 Port E9 bit 1 Port E9 bit 0
17 19 21 23 25 27 29 31	Ground	18 20 22 24 26 28 30 32	Port EA bit 3 Port EA bit 2 Port EA bit 1 Port EA bit 0 Port EA bit 4 Port EA bit 5 Port EA bit 6 Port EA bit 7
33 35 37 39 41 43 45 47	Ground	34 36 38 40 42 44 46 48	Port E8 bit 7 Port E8 bit 6 Port E8 bit 5 Port E8 bit 4 Port E8 bit 3 Port E8 bit 2 Port E8 bit 1 Port E8 bit 0
49	Ground	EXT INTRO/	
*All odd-numbered pins (1, 3, 5, 49) are on com- ponent side of the board. Pin 1 is the right-most pin when viewed form the component side of the board with the extractors at the top.			

Both cable assemblies consist of a 50-conductor flat cable with a 50-pin PC connector at one end. When attaching the cable to J1 or J2, be sure that the connector is oriented properly with respect to pin 1 on the edge connector. (Refer to the footnotes in tables 2-19 and 2-20.)

2-30. SERIAL I/O CABLING

Pin assignments and signal definitions for RS232C serial I/O interface are listed in table 2-10. An Intel iSBC 955 Cable Set is recommended for RS232C interfacing. One cable assembly consists of a 25-conductor flat cable with a 26-pin PC connector at one end and an RS232C interface connector at the other end. The second cable assembly includes an RS232C connector at one end and has spade lugs at the other end; the spade lugs are used to interface to a teletypewriter. (See Appendix B for ASR-33 TTY interface instructions.)

For OEM applications where cables will be made for the iSBC 80/30, it is important to note that the mating connector for J3 has 26 pins whereas the RS232C connector has 25 pins. Consequently, when connecting the 26-pin mating connector to 25-conductor flat cable, be sure that the

Table 2-20. Connector J2 Pin Assignments

Pin*	Function	Pin*	Function
1 3 5 7 9 11 13 15	Ground	2 4 6 8 10 12 14 16	Port 2 bit 7 Port 2 bit 6 Port 2 bit 5 Port 2 bit 4** Port 2 bit 3** Port 2 bit 3** Port 2 bit 2** Port 2 bit 1** Port 2 bit 0**
17 19 21 23 25 27 29 31	Ground	18 20 22 24 26 28 30 32	T0 Input T1 Input Not Used Not Used Not Used Not Used 41RS232 Out* * 41RS232 In* *
33 35 37 39 41 43 45 45	Ground	34 36 38 40 42 44 46 48	Port 1 bit 7 Port 1 bit 6 Port 1 bit 5 Port 1 bit 4 Port 1 bit 3 Port 1 bit 2 Port 1 bit 2 Port 1 bit 1 Port 1 bit 0
49	Ground	50	EXT INTR1/

*All odd-numbered pins (1, 3, 5, ... 49) are on component side of the board. Pin 1 is the right-most pin when viewed from the component side of the board with the extractors at the top.

**Jumpered pin or function. Refer to figure 5-2 sheet 5.

Signals	Symbol	Parameter Test Description Condition		Min.	Max.	Units
Port E8 Bidiroctional	V _{OL}	Output Low Voltage	l _{OL} = 20 mA		0.45	V
Drivers	V _{он}	Output High Voltage	l _{он} = –12 mA	2.4		v
	VIL	Input Low Voltage			0.95	V
	V _{IH}	Input High Voltage		2.0		v
	IιL	Input Current at Low V	$V_{IN} = 0.45V$		-5.25	mA
	CL	Capacitive Load			18	pF
8255A Driver/Passiver	VoL	Output Low Voltage	l _{OL} = 1.7 mA		0.45	v
Diver/neceiver	VoH	Output High Voltage	$I_{OH} = -200 \ \mu A$	2.4		v
	VIL	Input Low Voltage	· .		0.8	v
	V _{IH}	Input High Voltage		2.0		v
	i _{IL}	Input Current at Low V	$V_{IN} = 0.45$		10	μA
	I _{IH}	Input Current at High V	V _{IN} = 5.0		10	μA
	CL	Capacitive Load			18	pF
8041/8741A Driver/Receiver	V _{OL}	Output Low Voltage	l _{OL} = 1.6 mA		0.45	v
Divernieceiver	V _{OH}	Output High Voltage	$I_{OH} = 50 \ \mu A$	2.4		v
	VIL	Input Low Voltage			0.8	v
	V _{IH}	Input High Voltage		2.0		v
	l _{IL} '	Input Current at Low V	$V_{IN} = 0.8V$		0.4	μΑ
	Чн	Input Current at High V	$V_{IN} = 5.25V$		10	μΑ
·	CL	Capacitive Load			18	pF
EXT INTO	VIL	Input Low Voltage			0.8	v
	V _{IH}	Input High Voltage		2.0		v
	I _{IL}	Input Current at Low V	$V_{IN} = 0.4V$		-5.5	3 mA
	Чн	Input Current at High V	$V_{IN} = 2.7V$		20	3 μΑ
	CL	Capacitive Load			15	pF
EXT INT1	VIL	Input Low Voaltage			0.8	v
	ViH	Input High Voltage		2.0		v
	l _{IL}	Input Current at Low V	$V_{IN} = 0.4V$		-0.4	mA
	Чн	Input Current at High V	$V_{IN} = 2.7V$		20	μΑ
	CL	Capacitive Load			15	pF

	Table 2-21.	Parallel I/O Signal	(Connectors J1/J2)) DC	Characteristics
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cable makes contact with pins 1 and 2 of the mating connector and not with pin 26. Table 2-22 provides pin correspondence between connector J3 and an RS232C connector. When attaching the cable to J3, be sure that the PC connector is oriented properly with respect to pin 1 on the edge connector. (Refer to the footnote in table 2-10.)

Table 2-22. Connector J3 Vs RS232CPin Correspondence

PC Conn. J3	RS232C Conn.	PC Conn. J3	RS232C Conn.
1	14	14	7
2	1	15	21
3	15	16	8
4	2	17	22
5	16	18	9
6	3	19	23
7	17	20	10
8	4	21	24
9	18	22	11
10	5	23	25
11	19	24	12
12	6	25	N/C
13	20	26	13

2-31. BOARD INSTALLATION



Always turn off the computer system power supply before installing or removing the iSBC 80/30 board and before installing or removing device interface cables. Failure to take these precautions can result in damage to the board.

In an iSBC 80 Single Board Computer based system, install the iSBC 80/30 in any slot that has not been wired for a dedicated function. In an Intellec System, install the iSBC 80/30 in any slot except slot 1 or 2. Make sure that auxiliary connector P2 (if used) mates with the user-installed mating connector. Attach the appropriate cable assemblies to connectors J1 through J3.



CHAPTER 3 PROGRAMMING INFORMATION

3-1. INTRODUCTION

This chapter lists the on-board memory and I/O address assignments, describes the effects of a system initialize command, and provides programming information for the following programmable chips:

- a. Intel 8251A USART (Universal Synchronous/ Asynchronous Receiver/Transmitter) that controls the serial I/O port.
- b. Intel 8253 PIT (Programmable Interval Timer) that controls various frequency and timing functions.
- c. Intel 8255A PPI (Programmable Peripheral Interface) that controls the three parallel I/O ports.
- d. Intel 8259A PIC (Programmable Interrupt Controller) that can handle up to eight vectored priority interrupts for the on-board 8085A microprocessor (CPU).
- e. Intel 8041/8741 UPI (Universal Peripheral Interface).

This chapter also discusses the Intel 8085A Microprocessor interrupt capability. The instruction set for the 8085A is included in Appendix A; a complete description of programming with Intel's assembly language is given in the 8080/8085 Assembly Language Programming Manual, Order No. 9800310.

This chapter does not provide assembly language programming information for the optional Intel 8041/8741 UPI (Universal Peripheral Interface). This information is available in the *UPI-41 User's Manual*, Order No. 9800504A.

3-2. FAILSAFE TIMER

The 8085A microprocessor (CPU) expects an acknowledge signal to be returned from the addressed I/O or memory device in response to each Read or Write Command. The iSBC 80/30 includes a Failsafe Timer that is triggered during T_1 of every machine cycle. If the Failsafe Timer is enabled by hardwire jumper as described in table 2-4, and an acknowledge signal is not received within 10 milliseconds, the Failsafe Timer will time out and allow the CPU to exit the wait state. As described in Chapter 2, provision is made so that the Failsafe Timer output (BUS TIME OUT) can optionally be used to interrupt the CPU and/or to drive a front panel indicator.

If the Failsafe Timer is not enabled, and an acknowledge signal is not returned for any reason, the CPU will hang up

in a wait state. In this situation, the only way to free the CPU is to initialize the system as described in paragraph 3-5.

3-3. MEMORY ADDRESSING

The iSBC 80/30 includes 16K of dynamic RAM and two IC sockets to accommodate up to 8K of user-installed ROM/PROM. The iSBC 80/30 features a two-port RAM access arrangement in which the on-board RAM can be accessed by the on-board 8085A microprocessor (CPU) or by another bus master board via the Multibus. The ROM/PROM can be accessed only by the CPU.

The on-board RAM can be accessed by another bus master that currently has control of the Multibus. It should be noted, however, that even though another bus master may be continuously accessing the iSBC 80/30 on-board RAM, this does not lock out the CPU from accessing the on-board RAM. In this situation, Memory Commands from the CPU and the controlling bus master are interleaved. This, of course, will impose CPU wait states until the current access by the controlling bus master is completed.

Addresses for CPU access of ROM/PROM and on-board RAM are provided in table 3-1. Note that the ROM/ PROM address space depends on the user's configura-

 Table 3-1. On-Board Memory Addresses (For CPU Access)

On-Board Memory	Configuration	Legal Address	Illegal Address		
ROM/PROM	One 1K × 8 chip Two 1K × 8 chips	0000-03FF 0000-07FF	0400-07FF —		
ROM/PROM	One 2K × 8 chip Two 2K × 8 chips	0000-07FF 0000-0FFF	0800-0FFF —		
ROM/PROM	One 4K × 8 chip Two 4K × 8 chips	0000-0FFF 0000-1FFF	1000-1FFF —		
RAM	8K Access	2000-3FFF *4000-5FFF 6000-7FFF 8000-9FFF A000-BFFF C000-DFFF E000-FFFF	None None None None None None		
RAM *16K Access *4000-7FFF None 8000-BFFF None C000-FFFF None					
*Default (factory connected) jumper; refer to paragraph 2-16. Address- ing RAM outside the jumper-selected block results in an off-board request via the Multibus					

tion, and that the RAM address space depends on whether the board jumpers are configured to allow the CPU to access 8K or 16K of on-board RAM.

For Multibus access, the on-board RAM may be mapped into any 8K or 16K segment within the addressing constraints of the controlling bus master. In other words, for 16-bit Multibus addressing, the RAM may be mapped into any 8K or 16K segment of the 64K byte address space. For 20-bit Multibus addressing, the RAM may be mapped into any 8K or 16K segment of the 1-megabyte address space. Additional information is provided in paragraphs 2-17 through 2-19.

When the CPU is addressing *on-board* memory (ROM/ PROM or RAM), an internal PROM or RAM Advanced Acknowledge (AACK/) is automatically generated to prevent imposing a CPU wait state. When the CPU is addressing *system* memory via the Multibus, the CPU must first gain control of the Multibus and, after the Memory Read or Memory Write Command is given, must wait for a Transfer Acknowledge (XACK/) to be received from the addressed memory device. The Failsafe Timer, if enabled, will prevent a CPU hang-up in the event of a memory device equipment failure or a bus failure.

It should be noted in table 3-1 that it is possible to configure ROM/PROM such as to create *illegal* addresses. If an illegal address is used in conjunction with a Memory Write Command to ROM/PROM, a PROM AACK/ signal is generated as though the address was legal and the CPU will continue executing the program. However, in this case, erroneous data will be returned.

3-4. I/O ADDRESSING

The on-board 8085A microprocessor (CPU) communicates with the programmable chips through a sequence of I/O Read and I/O Write Commands. As shown in table 3-2, each of these chips recognizes four separate hexadecimal I/O addresses that are used to control the various programmable functions. Where two hexadecimal addresses are listed for a single function, either address may be used. For example, an I/O Read Command to ED or EF will read the status of the 8251A USART.

NOTE

The on-board I/O functions are not accessible to another bus master via the Multibus.

3-5. SYSTEM INITIALIZATION

When power is initially applied to the system, an Initialize (INIT/) signal is automatically generated that clears the internal Program Counter, Instruction Register, and Interrupt Enable flip-flop and "resets" the 8251A USART, 8255A PPI, and optional 8041/8741A UPI as follows:

Table	3-2.	I/O	Address	Assignments
-------	------	-----	---------	-------------

I/O Address	Chip Select	Function
D8 or DA		Write: ICW1, OCW2, and OCW3 Read: Status and Poll
D9 or DB	8259A PIC	Write: ICW2 and OCW1 (Mask) Read: OCW1 (Mask)
DC		Write: Counter 0 (Load Count ÷ N) Read: Counter 0
DD	9052 DIT	Write: Counter 1 (Load Count ÷ N) Read: Counter 1
DE	8253 PIT	Write: Counter 2 (Load Count ÷ N) Read: Counter 2
DF		Write: Control Read: None
E4 or E6	8041/8741A UPI	Write: Data (J2) Read: Data (J2) Read: Data
E5 or E7		Write: Command Read: Status
E8		Write: Port A (J1) Read: Port A (J1)
E9	8255A PPI	Write: Port B (J1) Read: Port B (J1)
EA		Write: Port C (J1) Read: Port C (J1)
EB		Write: Control Read: None
EC or EE	8251A	Write: Data (J3) Read: Data (J3)
ED or EF	USART	Write: Mode or Command Read: Status

- a. The 8251A USART is set to an "idle" mode, waiting for a set of Command Words to program the desired function.
- b. All three ports of the 8255A PPI are set to the input mode.
- c. The 8041/8741A UPI internal Program Counter and Status flip-flops are cleared.

The 8253 PIT and 8259A PIC are not affected by the power-up sequence.

The INIT/ signal is also gated onto the Multibus to set the remainder of the system components to a known internal state.

The INIT/ signal can also be generated by an auxiliary RESET switch. Pressing and releasing the RESET switch produces the same effect as the INIT/ signal described above.

3-6. 8251A USART PROGRAMMING

The USART converts parallel output data into virtually any serial output data format (including IBM Bi-Sync) for half- or full-duplex operation. The USART also converts serial input data into parallel data format.

Prior to starting transmitting or receiving data, the USART must be loaded with a set of control words. These control words, which define the complete functional operation of the USART, must immediately follow a reset (internal or external). The control words are either a Mode instruction or a Command instruction.

3-7. MODE INSTRUCTION FORMAT

The Mode instruction word defines the general characteristics of the USART and must follow a reset operation. Once the Mode instruction word has been written into the USART, sync characters or command instructions may be inserted. The Mode instruction word defines the following:

- a. For Sync Mode:
 - (1) Character length
 - (2) Parity enable
 - (3) Even/odd parity generation and check
 - (4) External sync detect (not supported by iSBC 80/30)
 - (5) Single or double character sync
- b. For Async Mode:
 - (1) Baud rate factor (X1, X16, or X64)
 - (2) Character length
 - (3) Parity enable
 - (4) Even/odd parity generation and check
 - (5) Number of stop bits

Instruction word and data transmission formats for synchronous and asynchronous modes are shown in figures 3-1 through 3-4.

3-8. SYNC CHARACTERS

Sync characters are written to the USART in the synchronous mode only. The USART can be programmed for either one or two sync characters; the format of the sync characters is at the option of the programmer.

3-9. COMMAND INSTRUCTION FORMAT

The Command instruction word shown in figure 3-5 controls the operation of the addressed USART. A Command











Figure 3-3. USART Asynchronous Mode Instruction Word Format



•NOTE: IF CHARACTER LENGTH IS DEFINED AS 5, 6 OR 7 BITS THE UNUSED BITS ARE SET TO "ZERO".

Figure 3-4. USART Asynchronous Mode Transmission Format



Figure 3-5. USART Command Instruction Word Format

instruction must follow the mode and/or sync words and, once the Command instruction is written, data can be transmitted or received by the USART.

It is not necessary for a Command instruction to precede all data transactions; only those transmissions that require a change in the Command instruction. An example is a change in the enable transmit bit or enable receive bit. Command instructions can be written to the USART at any time after one or more data operations.

After initialization, always read the chip status and check for the TXRDY bit prior to writing either data or command words to the USART. This ensures that any prior input is not overwritten and lost. Note that issuing a Command instruction with bit 6 (IR) set will return the USART to the Mode instruction format.

3-10. RESET

To change the Mode instruction word, the USART must receive a Reset command. The next word written to the USART after a Reset command is assumed to be a Mode instruction. Similarly, for sync mode, the next word after a Mode instruction is assumed to be one or more sync characters. All control words written into the USART after the Mode instruction (and/or the sync character) are assumed to be Command instructions.

3-11. ADDRESSING

The USART chip uses two consecutive pairs of addresses. The lower of the two addresses in each pair is used to read and write I/O data; the upper address in each pair is used to write mode and command words and to read the USART status. (Refer to table 3-2.)

3-12. INITIALIZATION

A typical USART initialization and I/O data sequence is presented in figure 3-6. The USART chip is initialized in four steps:

- a. Reset USART to Mode instruction format.
- b. Write Mode instruction word. One function of mode word is to specify synchronous or asynchronous operation.
- c. If synchronous mode is selected, write one or two sync characters as required.
- d. Write Command instruction word.

To avoid spurious interrupts during USART initialization, disable the USART interrupt. This can be done by either masking the appropriate interrupt request input at the 8259A PIC or by disabling the CPU interrupts by executing a DI instruction. First, reset the USART chip by writing a Command instruction to location ED (or EF). The Command instruction must have bit 6 set (IR6 = 1); all other bits are immaterial.

NOTE

This reset procedure should be used only if the USART has been completely initialized, or if the initialization procedure has reached the point that the USART is ready to receive a Command word. For example, if the reset command is written when the initialization sequence calls for a sync character, then subsequent programming will be in error.

Next, write a Mode instruction word to the USART. (See figures 3-1 through 3-4.) A typical subroutine for writing both Mode and Command instructions is given in table 3-3.

If the USART is programmed for the synchronous mode, write one or two sync characters depending on the transmission format.

Finally, write a Command instruction word to the USART. Refer to figure 3-5 and table 3-3.

3-13. OPERATION

Normal operating procedures use data I/O read and write, status read, and Command instruction write operations. Programming and addressing procedures for the above are summarized in following paragraphs.

NOTE

After the USART has been initialized, always check the status of the TXRDY bit *prior* to writing data or writing a new command word to the USART. The TXRDY bit *must* be *true* to prevent overwriting and subsequent loss of command or data words. The TXRDY bit is inactive until initialization has been completed; therefore, do not check TXRDY until after the command word, which concludes the initialization procedure, has been written.

Prior to any operating change, a new command word must be written with command bits changed as appropriate. (Refer to figure 3-5 and table 3-3.)



*The second sync character is skipped if Mode instruction has programmed USART to single character internal sync mode. Both sync characters are skipped if Mode instruction has programmed USART to async mode.

Figure 3-6. Typical USART 611-6 Initialization and Data I/O Sequence

3-14. DATA INPUT/OUTPUT. For data receive or transmit operations perform a read or write, respectively, to the USART. Tables 3-4 and 3-5 provide examples of typical character read and write subroutines.

During normal transmit operation, the USART generates a Transmit Ready (TXRDY) signal that indicates that the

 Table 3-3. Typical USART Mode or Command Instruction Subroutine

;CMD2 OU ;USES-A,	D2 OUTPUTS CONTROL WORD TO USART. ES-A, STAT; DESTROYS-NOTHING.				
	EXTRN	STAT			
CMD2:	PUSH CALL ANI JZ POP OUT RET END	PSW STAT 1 LP PSW 0ED	;CHECK TXRDY ;TXRDY MUST BE TRUE ;ENTER HERE FOR INITIALIZATION		





Table 3-5.	Typical	USART Dat	a Character	Write	Subroutine
------------	---------	------------------	-------------	-------	------------

;TX1 WRITES I ;USES-STAT; E	;TX1 WRITES DATA CHARACTER FROM REG A TO USART. ;USES-STAT; DESTROYS-A,FLAGS.						
EXTRN	STAT						
TX1: PUSH TX11: CALL	PSW STAT	;SAVE DATA					
ANI JZ POP	1 TX11 PSW	;CHECK FOR TXRDY					
OUT RET	OEC	;ENTER HERE IF TXRDY IS TRUE					
END							

USART is ready to accept a data character for transmission. TXRDY is automatically reset when the CPU loads a character into the USART.

Similarly, during normal receive operation, the USART generates a Receive Ready (RXRDY) signal that indicates that a character has been received and is ready for input to the CPU. RXRDY is automatically reset when a character is read by the CPU.

The TXRDY and RXRDY outputs of the USART are available at the priority interrupt jumper matrix. If, for instance, TXRDY and RXRDY are input to the 8259A PIC, the PIC resolves the priority and drives the INTR input high to the CPU. TXRDY and RXRDY are also available in the status word. (Refer to paragraph 3-15.)

3-15. STATUS READ. The CPU can determine the status of the serial I/O port by issuing an I/O Read Command to the upper address (ED or EF) of the USART chip. The format of the status word is shown in figure 3-7. A typical status read subroutine is given in table 3-6.

3-16. 8253 PIT PROGRAMMING

A 22.1184-MHz crystal oscillator supplies the basic clock frequency for the programmable chips. This clock frequency is divided by 9, 18, and 144 to produce three jumper-selectable clocks: 2.4576 MHz, 1.2288 MHz, and 153.6 kHz. These clocks are available for input to Counter 0, Counter 1, and Counter 2 of the 8253 PIT. The default (factory connected) and optional jumpers for selecting the clock inputs to the three counters are listed in table 2-4.

Default jumpers connect the output of Counter 2 to the TXC and RXC inputs of the 8251A USART. Jumpers are included so that Counters 0 and 1 can provide real-time interrupts or provide an event clock to the 8041/8741 UPI and a count out clock to 8255A PPI Port EA I/O driver.

Before programming the 8253 PIT, ascertain the input clock frequency and the output function of each of the three counters. These factors are determined and established by the user during the installation.



Table 3-6. Typical USART Status Read Subroutine

;STAT REA ;USES-NO	NDS STATU THING; DE			
;STAT	IN RET	0ED	;GET STATUS	
	END			

3-17. MODE CONTROL WORD AND COUNT

All three counters must be initialized separately prior to their use. The initialization for each counter consists of two steps:

- a. A mode control word (figure 3-8) is written to the control register for each individual counter.
- b. A down-count number is loaded into each counter; number is in one or two 8-bit bytes as determined by mode control word.

The mode control word (figure 3-8) does the following:

- a. Selects counter to be loaded.
- b. Selects counter operating mode.
- c. Selects one of the following four counter read/load functions:
 - (1) Counter latch (for stable read operation).
 - (2) Read or load most-significant byte only.
 - (3) Read or load least-significant byte only.
 - (4) Read or load least-significant byte first, then most-significant byte.
- d. Sets counter for either binary or BCD count.

D7	D ₆	D_5	D4	D3	D ₂	D1	D ₀				
SC1	SC0	RL1	RL0	M2	M1	мо	вCD				
		L				ı	· [' (BINA	RY/E	BCD)	
	ĺ							0		Bi	nary Counter (16-bits)
								1		Bi (4	nary Coded Decimal (BCD) Counter Decades)
								M2	M1	MO	(MODE)
								0 0 X	0 0 1	0 1 0	Mode 0 Mode 1 Mode 2
									0	0	Mode 3 Mode 4 Mode 5
								RL1		RL0	(READ/LOAD)
								0		0	Counter Latching operation (refer to paragraph 3-29).
								1		0	Read/Load most significant byte only.
								0		1	Read/Load least significant byte only.
								1		1	Read/Load least significant byte first, then most significant byte.
								SC1		SC	0 (SELECT COUNTER)
								0		0	Select Counter 0
								0		1	Select Counter 1
								1		0	Select Counter 2
								1		1	lllegal
					Figu	ire 3-	8. PI	Т Мо	de C	ontrol	Word Format

The mode control word and the count register bytes for any given counter must be entered in the following sequence:

- a. Mode control word.
- b. Least-significant count register byte.
- c. Most-significant count register byte.

As long as the above procedure is followed *for each counter*, the chip can be programmed in any convenient sequence. For example, mode control words first can be loaded into each of three counters, followed by the least-significant byte, etc. Figure 3-9 shows the two programming sequences described above.

Since all counters in the PIT chip are downcounters, the value loaded in the count registers is decremented. Loading all zeroes into a count register results in a maximum count of 2^{16} for binary numbers or 10^4 for BCD numbers.

When a selected count register is to be loaded, it *must* be loaded with the number of bytes programmed in the mode control word. One or two bytes can be loaded, depending on the appropriate down count. These two bytes can be programmed at any time following the mode control word, as long as the correct number of bytes is loaded in order.

The count mode selected in the control word controls the counter output. As shown in figure 3-8, the PIT chip can operate in any of six modes:

PROGRAMMING FORMAT

Step

1		Mode Control Word Counter n
2	LSB	Count Register Byte Counter n
3	MSB	Count Register Byte Counter n

ALTERNATE PROGRAMMING FORMAT

	Mode Control Word Counter 0
	Mode Control Word Counter 1
	Mode Control Word Counter 2
LSB	Counter Register Byte Counter 1
MSB	Count Register Byte Counter 1
LSB	Count Register Byte Counter 2
MSB	Count Register Byte Counter 2
LSB	Count Register Byte Counter 0
MSB	Count Register Byte Counter 0
	LSB MSB LSB MSB LSB MSB

450-18

Figure 3-9. PIT Programming Sequence Examples

- a. Mode 0: Interrupt on terminal count. In this mode, Counters 1 and 2 can be used for auxiliary functions such as generating real-time interrupt intervals. After the count value is loaded into the count register, the counter output goes low and remains low until the terminal count is reached. The output then goes high until either the count register or the mode control register is reloaded.
- b. Mode 1: Programmable one-shot. In this mode, the output of Counter 1 and/or Counter 2 will go low on the count following the rising edge of the GATE input from Port EA. The output will go high on the terminal count. If a new count value is loaded while the output is low, it will not affect the duration of the one-shot pulse until the succeeding trigger. The current count can be read at any time without affecting the one-shot pulse. The one-shot is retriggerable, hence the output will remain low for the full count after any rising edge of the gate input.
- c. Mode 2: Rate generator. In this mode, the output of Counter 1 and/or Counter 2 will be low for one period of the clock input. The period from one output pulse to the next equals the number of input counts in the count register. If the count register is reloaded between output pulses, the present period will not be affected but the subsequent period will reflect the new value. The gate input, when low, will force the output high. When the gate input goes high, the

counter will start from the initial count. Thus, the gate input can be used to synchronize the counter. When Mode 2 is set, the output will remain high until after the count register is loaded; thus, the count can be synchronized by software.

- d. Mode 3: Square wave generator. Mode 3, which is the primary operating mode for Counter 2, is used for generating Baud rate clock signals. In this mode, the counter output remains high until one-half of the count value in the count register has been decremented (for even numbers). The output then goes low for the other half of the count. If the value in the count register is odd, the counter output is high for (N + 1)/2 counts, and low for (N - 1)/2 counts.
- e. Mode 4. Software triggered strobe. After this mode is set, the output will be high. When the count is loaded, the counter begins counting. On terminal count, the output will go low for one input clock period and then go high again. If the count register is reloaded between output pulses, the present count will not be affected, but the subsequent period will reflect the new value. The count will be inhibited while the gate input is low. Reloading the count register will restart the counting for the new value.
- f. Mode 5: Hardware triggered strobe. The counter will start counting on the rising edge of the gate input and the output will go low for one clock period when

the terminal count is reached. The counter is retriggerable; the output will not go low until the full count after the rising edge of the gate input.

Table 3-7 provides a summary of the counter operation versus the gate inputs. The gate inputs to Counters 0 and 2 are tied high by default jumpers; these gates may optionally be controlled by Port EA. The gate input to Counter 2 is floating high and not optionally controlled.

Table 3-7. PIT Counter Op	peration Vs.	Gate Inputs
---------------------------	--------------	--------------------

Signal Status Modes	Low Or Going Low	Rising	High
0	Disables counting	1	Enables counting
1		 Initiates counting Resets output after next clock 	—
2	 Disables counting Sets output high immediately 	Initiates counting	Enables counting
3	 Disables counting Sets output high immediately 	Initiates counting	Enables counting
4	Disables counting		Enables counting
5		Initiates counting	_

3-18. ADDRESSING

As listed in table 3-2, the PIT uses four consecutive I/O addresses: DC through DF. Addresses DC, DD, and DE, respectively, are used in loading and reading the count in Counters 0, 1, and 2. Address DF is used in writing the mode control word to the desired counter.

3-19. INITIALIZATION

To initialize the PIT chips, perform the following:

- a. Write mode control word for Counter 0 to DF. Note that *all* mode control words are written to DF, since mode control word must specify which counter is being programmed. (Refer to figure 3-8.) Table 3-8 provides a sample subroutine for writing mode control words to all three counters.
- b. Assuming mode control word has selected a 2-byte load, load least-significant byte of count into Counter 0 at DC. (Count value to be loaded is described in paragraphs 3-22 through 3-24). Table 3-9 provides a sample subroutine for loading 2-byte count value.
- c. Load most-significant byte of count into Counter 0 at DC.

NOTE

Be sure to enter the downcount in two bytes if the counter was programmed for a two-byte entry in the mode control word. Similarly, enter the downcount value in BCD if the counter was so programmed.

d. Repeat steps a, b, and c for Counters 1 and 2.

Table 3-8. Typical PIT Control Word Subroutine

;INTTMR INITIALIZES COUNTERS 0, 1, 2. ;COUNTERS 0 AND 1 ARE INITIALIZED AS INTERRUPT TIMERS. ;COUNTER 2 IS INITIALIZED AS BAUD RATE GENERATOR. ;ALL THREE COUNTERS ARE SET UP FOR 16-BIT OPERATION. ;USES-NOTHING; DESTROYS-A.							
INTTMR:	MVI OUT	A,30H 0DF	;MODE CONTROL WORD FOR COUNTER 0				
	MVI OUT	A,70H 0DF	;MODE CONTROL WORD FOR COUNTER 1				
	MVI OUT RET	A, B6H 0DF	;MODE CONTROL WORD FOR COUNTER 2				
	END						

;LOAD0 ;USES-E	LOADS CO),E; DEST	DUNTER 0 FF ROYS-A.	ROM D&E,D IS MSB, E IS LSB.	
LOAD0:	MOV OUT MOV OUT RET	A,E ODC A,D ODC	;GET LSB ;GET MSB	
	END			

Table 3-9. Typical PIT Count Value Load Subroutine

3-20. OPERATION

The following paragraphs describe operating procedures for a counter read, clock frequency divider/ratio selection, and interrupt timer count selection.

3-21. COUNTER READ. There are two methods that can be used to read the contents of a particular counter. The first method involves a simple read of the desired counter. The only requirement with this method is that, in order to ensure a stable count reading, the desired counter must be *inhibited* by controlling its gate input. Only Counter 0 and Counter 1 can be read using this method because the gate input to Counter 2 is not controllable.

The second method allows the counter to be read "onthe-fly." The recommended procedure is to use a mode control word to latch the contents of the count register; this ensures that the count reading is accurate and stable. The latched value of the count can then be read.

NOTE

If a counter is read during count, it is mandatory to complete the read procedure; that is, if two bytes were programmed to the counter, then two bytes *must* be read before any other operations are performed with that counter.

To read the count of a particular counter, proceed as follows (a typical counter read subroutine is given in table 3-10):

- a. Write counter register latch control word (figure 3-10) to DF. Control word specifies desired counter and selects counter latching operation.
- b. Perform a read operation of desired counter; refer to table 3-2 for counter addresses.

NOTE

Be sure to read one or two bytes, whichever was specified in the initialization mode control word. For two bytes, read in the order specified.

3-22. CLOCK FREQUENCY/DIVIDE RATIO SELECTION. Table 2-4 lists the default and optional timer input frequencies to Counters 0 through 2. The timer input frequencies are divided by the counters to generate the 8041/8741 Event Clock (Counter 0), Count Out (Counter 1), and the 8251A Baud Rate Clock (Counter 2).

Each counter must be programmed with a downcount number, or count value N. When count value N is loaded into a counter, it becomes the clock divisor. To derive N for either synchronous or asynchronous RS232C operation, use the procedures described in following paragraphs.

3-23. Synchronous Mode. In the synchronous mode, the TXC and/or RXC rates equal the Baud rate. Therefore, the count value is determined by

N = C/B

;READ1 ;USES-N	READS CONTHING;	OUNTER 1 ON-THE-FL DESTROYS-A,D,E.	Y INTO D&E. MSB IN D, LSB IN E.	
READ1:	MVI OUT	A,40H 0DF	;MODE WORD FOR LATCHING COUNTER 1 VALUE	
	IN MOV	0DD E.A	LSB OF COUNTER	
	IN MOV RET	0DD D,A	;MSB OF COUNTER	
· · · · · · · · · · · · · · · · · · ·	END			

Table 3-10. Typical PIT Counter Read Subroutine





where N is the count value,

B is the desired Baud rate, and

C is 1.23 MHz, the input clock frequency.

Thus, for a 4800 Baud rate, the required count value (N) is:

$$N = \frac{1.23 \times 10^6}{4800} = \underline{256}.$$

If the binary equivalent of count value N = 256 is loaded into Counter 2, then the output frequency is 4800 Hz, which is the desired clock rate for synchronous mode operation.

3-24. ASYNCHRONOUS MODE. In the asynchronous mode, the TXC and/or RXC rates equal the Baud rate times one of the following multipliers: X1, X16, or X64. Therefore, the count value is determined by:

N = C/BM

where N is the count value,

B is the desired Baud rate,

M is the Baud rate multiplier (1, 16, or 64), and C is 1.23 MHz, the input clock frequency.

Thus, for a 4800 Baud rate, the required count value (N) is

$$N = \frac{1.23 \times 10^6}{4800 \times 16} = \underline{\underline{16.}}$$

If the binary equivalent of count value N = 16 is loaded into Counter 2, then the output frequency is 4800×16 Hz, which is the desired clock rate for asynchronous mode operation. Count values (N) versus rate multiplier (M) for each Baud rate are listed in table 3-11. NOTE

During initialization, be sure to load the count value (N) into the appropriate counter and the Baud rate multiplier (M) into the 8251A USART.

Table 3-11.	PIT	Count	Value	Vs	Rate	Multiplier	for
		Each	Baud	Rat	te		

Baud Rate	*Co	unt Value (N)	For
(B)	M = 1	M = 16	M = 64
75 110 150 300 600 1200 2400 4800 9600 19200 38400 76800	16384 11171 8192 4096 2048 1024 512 256 128 64 32 16	1024 698 512 256 128 64 32 16 8 4 2	256 175 128 64 32 16 8 4 2
*Count Value Double Cou Count Value decimal.	es (N) assum Int Values (N s (N) and Ra	ne clock is 1) for 2.46 M te Multipliers	I.23 MHz. IHz clock. (M) are in

3-25. RATE GENERATOR/INTERVAL TIMER. Table 3-12 shows the maximum and minimum rate generator frequencies and timer intervals for Counters 0 and 1 when these counters, respectively, have 1.23-MHz and 153.6-KHz clock inputs. The table also provides the maximum and minimum generator frequencies and time intervals and may be obtained by connecting Counters 0 and 1 in series.

3-26. INTERRUPT TIMER. To program an interval timer for an interruption terminal count, program the appropriate timer for the correct operating mode (Mode 0) in the control word. Then load the count value (N), which is derived by

$$N = TC$$

where

N is the count value for Counter 0,

T is the desired interrupt time interval in seconds, and

C is the internal clock frequency (Hz).

Table 3-13 shows the count value (N) required for several time intervals (T) that can be generated for Counters 0 and 1.

	Single Timer ¹ (Counter 0)		Singl (Co	e Timer ² unter 1)	Dual Timer ³ (0 and 1 in Series)		
	Minimum	Maximum	Minimum	Maximum	Minimum	Maximum	
Rate Generator (Frequency)	18.75 Hz	614.4 kHz	2.344 Hz	76.8 kHz	0.00029 Hz	307.2 kHz	
Real-Time Interrupt (Interval)	1.63 µsec	53.3 msec	13 µsec	426.67 msec	3.26 µsec	58.25 minutes	
(Interval) NOTES: 1. Assuming a	1.23-MHz clock i			<u> </u>	<u> </u>	<u> </u>	
2. Assuming a 2. Assuming a 3. Assuming C	1.23-MHZ Clock 1 153.6-kHz clock ounter 0 has 1.23	nput input. 3-MHz clock input					

Table	3-)	12.	PIT	Rate	Generator	Frequencie	s and	Timer	Intervals
-------	-----	-----	-----	------	-----------	------------	-------	-------	-----------

3-27. 8255A PPI PROGRAMMING

The three parallel I/O ports interfaced to connector J1 are controlled by an Intel 8255A Programmable Peripheral Interface. Port A includes bidirectional data buffers and Ports B and C include IC sockets for installation of either input terminators or output drivers depending on the user's application.

Default jumpers set the Port A bidirectional data buffers to the input mode. Optional jumpers allow the bidirectional data buffers to be set to the output mode or to allow any one of the eight Port C bits to selectively set the Port A bidirectional data buffers to the input or output mode.

Table 2-11 lists the various operating modes for the three PPI parallel I/O ports. Note that Port A (E8) can be operated in Modes 0, 1, or 2; Port B (E9) and Port C (EA) can be operated in Mode 0 or 1.

3-28. CONTROL WORD FORMAT

The control word format shown in figure 3-11 is used to initialize the PPI to define the operating mode of the three ports. Note that the ports are separated into two groups. Group A (control word bits 3 through 6) defines the

Table 3-13. PIT Timer Intervals Vs Timer Counts

Т	N*
10 μsec	12
100 μsec	123
1 msec	1229
10 msec	12288
50 msec	61440
*Count Values (N) assun	ne clock is 1.23 MHz.
Double Count Values	(N) for 2.46 MHz
clock. Count Values (N	J) are in decimal.

operating mode for Port A (E8) and the upper four bits of Port C (EA). Group B (control word bits 0 through 2) defines the operating mode for Port B (E9) and the lower four bits of Port C (EA). Bit 7 of the control word controls the mode set flag.



Table 3-14. Typical PPI Initialization Subroutine

;INT PAR I ;USES-NO	NITIALIZES F THING; DEST	ARALLEL PORTS. ROYS-A.	
INT PAR:	MVI OUT RET	A,92H 0EBH	;MODE WORD TO PPI PORT A&B IN,C OUT
	END		

3-29. ADDRESSING

The PPI uses four consecutive addresses (E8 through EB) for data transfer, obtaining the status of Port C (EA), and for port control. (Refer to table 3-2.)

3-30. INITIALIZATION

To initialize the PPI, write a control word to E8. Refer to figure 3-11 and table 3-14 and assume that the control word is 92 (hexadecimal). This initializes the PLI as follows:

- a. Mode Set Flag active
- b. Port A (E8) set to Mode 0 Input
- c. Port C (EA) upper set to Mode 0 Output
- d. Port B (E9) set to Mode 0 Input
- e. Port C (EA) lower set to Mode 0 Output

3-31. OPERATION

After the PPI has been initialized, the operation is simply performing a read or a write to the appropriate port.

3-32. READ OPERATION. A typical read subroutine for Port A is given in table 3-15.

3-33. WRITE OPERATION. A typical write subroutine for Port C is given in table 3-16. As shown in figure 3-12, and of the Port C bits can be selectively set or cleared by writing a control word to EB.

3-34. 8259A PIC PROGRAMMING

As described in paragraph 2-20, the PIC monitors the interrupt requests from eight separate sources. When one or more of the interrupt requests are active (true), the PIC determines the following:

- a. Which input signal has the highest priority.
- b. Whether the input signal has a higher priority than the interrupt presently being serviced by the main processor. If so, the interrupt being serviced is interrupted; if not, the input signal is held for later output.
- c. Whether the interrupt input bit is masked.

Thus, the basic functions of the PIC are (1) resolve the priority of interrupt requests and (2) issue a single interrupt request to the 8085A microprocessor (CPU) based on that priority. The output of the PIC is applied directly to the INTR input of the CPU. (Refer to paragraph 3-50.)

3-35. INTERRUPT PRIORITY MODES

The PIC has two modes for resolving the priority of interrupt inputs: (1) fully nested mode and (2) rotating mode. The rotating mode has two variations: auto-rotating and specific rotating.

3-36. FULLY NESTED MODE. In this mode the PIC input signals are assigned priority from 0 through 7. The PIC operates in this mode unless specifically programmed otherwise. Interrupt IR0 has the highest priority and IR7 has the lowest priority. When an interrupt is acknowledged, the highest priority request is available to the CPU. Lower priority interrupts are inhibited; higher prior-

Table 3-15. Typical PPI Port Read Subroutine

:	AREAD READ USES-A; DES	SABYTE STROYS-/	FROM PORT A INTO REGA. A.
A	AREAD:	IN RET	0E8H
		END	· · · · · · · · · · · · · · · · · · ·

Table 3-16.	Typical	PPI	Port	Write	Subroutine

;COUT C ;USES-A	UTPUTS A	A BYTE FROM REG A TO PORT C. YS-NOTHING.	
COUT:	OUT RET	OEAH	
	END		

ity interrupts will be able to generate an interrupt that will be acknowledged if the CPU has enabled its own interrupt input through software. An End-Of-Interrupt (EOI) command from the CPU is required to reset the PIC for the next interrupt.

3-37. AUTO-ROTATING MODE. In this mode the interrupt priority rotates. Once an interrupt on a given input is serviced, that interrupt assumes the lowest priority. Thus, if there are a number of simultaneous interrupts, the priority will rotate among the interrupts in numerical order. For example, if interrupts IR4 and IR6 request service simultaneously, IR4 will receive the highest priority. After service, the priority level rotates so that IR4 has the lowest priority and IR5 assumes the highest priority. In the worst case, seven other interrupts are serviced before IR4 again has the highest priority. Of course, if IR4 is the only request, it is serviced promptly. In the Auto-Rotating Mode, priority shifts when the PIC chip receives an End-of-Interrupt (EOI) command.

3-38. SPECIFIC ROTATING MODE. In this mode the software can change interrupt priority by specifying the bottom priority, which automatically sets the highest priority. For example, if IR5 is assigned the bottom priority, IR6 assumes the highest priority. In the specific rotat-





ing mode, the priority can be rotated by writing a Specific Rotate at EOI (SEOI) command to the PIC chip. This command contains the BCD code of the interrupt being serviced; that interrupt is reset as the bottom priority. In addition, the bottom priority interrupt can be fixed at any time by writing a command word to the PIC chip.

3-39. INTERRUPT MASK

One or more of the eight interrupt request inputs can be individually masked during the PIC initialization or at any subsequent time. If an interrupt is masked while it is being serviced, lower priority interrupts are inhibited. There are two ways to enable the lower priority interrupts:

- a. Write an End-of-Interrupt (EOI) command.
- b. Set the Special Mask Mode.

The Special Mask Mode is useful when one or more interrupts are masked. If for any reason an input is masked while it is being serviced, the lower priority interrupts are disabled. However, it is possible to enable the lower priority interrupt with the Special Mask Mode. In this mode, the lower priority lines are enabled until the Special Mask Mode is reset. Higher priorities are not affected.

3-40. STATUS READ

Interrupt request inputs is handled by the following two internal PIC registers:

- a. Interrupt Request Register (IRR), which stores all interrupt levels that are requesting service.
- b. In-Service Register (ISR), which stores all interrupt levels that are being serviced.

Either register can be read by writing a suitable command word and then performing a read operation.

3-41. INITIALIZATION COMMAND WORDS

The eight devices serviced by the PIC have eight addresses equally spaced in memory that can be programmed at intervals of four or eight bytes. Interrupt service routines for these eight devices thus occupy a 32- or 64-byte block, respectively, of memory. The address format for device interrupt service routine is shown in figure 3-13.



Figure 3-13. PIC Device Interrupt Addresses

Bits 0-4 are automatically inserted by the PIC whereas bits 6-15 are programmed by Initialization Command Words ICW1 and ICW2. If the address interval is eight bytes, bit 5 is automatically inserted by the PIC; if the address interval if four bytes, bit 5 is programmed in ICW1. Thus, the 32-byte or 64-byte block of addresses reserved for interupt service routines can be located anywhere in the available memory space. Table 3-17 shows the address format inserted by the PIC for each device.

Initialization of the PIC consists of writing two or three 8-bit Initialization Command words as shown in figure 3-14. Since there are no slave PIC's, the initialization for the one PIC consists of writing two Initialization Command Words as follows:

- a. The first Initialization Command Word (ICW1) consists of the following:
 - (1) Bits 5-7 specify the most-significant bits of the lower address byte of the interrupt service routine.
 - (2) Bit 3 establishes whether the interrupts are requested by a positive-true or requested by a low-to-high transition input. This applies to all input requests handled by the PIC. In other words, if bit 3-1, a low-to-high transition is required to request an interrupt on any of the eight levels handled by the PIC.
 - (3) Bit 2 specifies a 4-byte or 8-byte address interval.

- (4) Bit 1 specifies whether or not there are slave (cascaded) PIC's. Since there are no slave PIC's, set bit 1 = 1.
- (5) Bits 0 and 4 identify the word as ICW1.
- b. The second word (ICW2) specifies the upper byte (bits 8-15) of the interrupt service routine.

3-42. OPERATION COMMAND WORDS

After being initialized, the PIC can be programmed at any time for various interrupt modes. The Operation Command Word (OCW) formats are shown in figure 3-15 and discussed in paragraph 3-45.

3-43. ADDRESSING

The PIC uses two consecutive addresses for writing to and reading internal registers. Address functions pertinent to programming are identified in table 3-2.

3-44. INITIALIZATION

To initialize the PIC, proceed as follows (table 3-17 provides a typical initialization subroutine);

- a. Disable system interrupts by executing a DI (Disable Interrupts) instruction.
- b. Write ICW1 to D8.
- c. Write ICW2 to D9.
- d. Enable system interrupts by executing an EI (Enable Interrupts) instruction.

NOTE

The PIC operates in the fully nested mode after the initialization sequence without requiring any Operation Control Word (OCW).

3-45. OPERATION

After initialization, the PIC chips can be programmed at any time for the following operations:

Dovice							Lowe	r Routin	e Addres	s Byte						
Device				Inter	val = 4	1						Interv	al = 8			
	D7	D6	D5	.D4	D3	D2	D1	D0	D7	D6	D5	D4	D3	D2	D1	D0
IR7	A7	A6	A5	1	1	1	0	0	A7	A6	1	1	1	0	0	0
IR6	A7	A6	A5	1	1	0	0	0	A7	A6	1	1	0	Ō	0	0
, IR5	A7	A6	A5	1	0	1	0	0	A7	A6	1	Ó	1	Ō	Ó	0
IR4	A7	A6	A5	1	0	0	0	0	A7	A6	1	Ó	0	Ō	0	0
IR3	A7	A6	A5	0	1	1	Ō	0	A7	A6	Ó	1	1	ŏ	Ō	Ō
IR2	A7	A6	A5	0	1	Ó	ŏ	0	A7	A6	ō	1	ò	ō	ŏ	Ō
IR1	A7	A6	A5	Ō	Ó	1	Ō	0	A7	A6	Ō	Ó	1	õ	Ō	Ō
IR0	A7	A6	A5	0	Ō	Ó	Ō	Ō	A7	A6	Ō	ō	Ó	ō	Ō	Ō

Table 3-17. PIC Device Address Insertion



611-8

- a. Auto-rotating priority.
- b. Specific rotating priority.
- c. Status read of Interrupt Request Register (IRR).
- d. Status read of In-Service Register (ISR).
- e. Interrupt mask bits set, reset, or read.
- f. Special mask mode set or reset.

Table 3-19 lists details of the above operations. Note that an End-Of-Interrupt (EOI) or a Special End-Of-Interrupt (SEOI) command is required at the end of each interrupt service routine to reset the ISR. The EOI command is sued in the fully nested and auto-rotating priority modes and the SEOI command, which specifies the bit to be reset, is used in the specific rotating priority mode. Tables 3-20 through 3-24 provide typical subroutines for the following:

- a. Read IRR (table 3-20).
- b. Read ISR (table 3-21).
- c. Set mask register (table 3-22).
- d. Read mask register (table 3-23).
- e. Issue EOI command (table 3-24).

3-46. 8041/8741A UPI PROGRAMMING

Table 3-2 lists the addresses used for writing data and commands and for reading data and status words. Assembly language programming information for the Intel 8041/8741A is given in the *Intel UPI-41 User's Manual*, Order No. 9800504, and *Control With UPI-41*, Application Note AP-27.

Figure 3-16 illustrates the internal configuration of the UPI data bus buffer and status register. The CPU accesses the UPI data bus buffer register via I/O Read and Write Commands to E4 or E6. The CPU accesses the UPI status via an I/O Read Command to E5 or E7.

 Table 3-18. Typical PIC Initialization Subroutine

<u>en en e</u>	· · · · · · · · · · · · · · · · · · ·	· · · · · ·		
;INT59 IN ;4000H II ;PIC MA ;USES-S	NITIALIZES THE S SET UP FOR SK IS SET, DIS MASK; DESTR	E PIC. A 64-BY INTERRUPT (ABLING ALL P OYS-A.	TE ADDRESS BLOCK BEGINNING WITH SERVICE ROUTINES. IC INTERRUPTS.	
	EXTRN	SMASK		
INT59:	CALL MVI OUT MVI OUT MVI CALL RET	SETD A,12H 0D8H A,40H 0D9H 0FFH SMASK	ICW1 TO PIC	
	END			







3-18

Operation	Procedure
Auto-Rotating Priority Mode	To set: In OCW2, write a Rotate Priority at EOI command (A0H) to D8.
	Terminate interrupt and rotate priority: In OCW2, write EOI command (20H) to D8.
Specific Rotating Priority Mode	To set: In OCW2, write a Rotate Priority at SEOI command in the following format to D8:
	D7 D6 D5 D4 D3 D2 D1 D0
	BCD of IR line to be reset and/or put into lowest priority.
	To terminate interrupt and rotate priority: In OCW2, write an SEOI command in the following format to D8.
	D7 D6 D5 D4 D3 D2 D1 D0
	0 1 1 0 0 <u>L2 L1 L0</u>
	BCD of ISR flip-flop to be reset.
	To rotate priority without EOI: In OCW2, write a command word in the following format to D8.
	D7 D6 D5 D4 D3 D2 D1 D0
	1 1 0 0 0 <u>L2 L1 L0</u>
	BCD of bottom priority IR line.
Interrupt Request Register (IRR)	The IRR stores a "1" in the associated bit for each IR input line that is requesting an interrupt. To read the IRR (refer to footnote):
Sidius	(1) Write 0BH to D8.(2) Read D8. Status format is as follows:
	D7 D6 D5 D4 D3 D2 D1 D0
	IR Line: 7 6 5 4 3 2 1 0
In-Service Register (ISR) Status	The ISR stores a "1" in the associated bit for priority inputs that are being serviced. The ISR is updated when an EOI command is issued. To read the ISR (refer to footnote):
	(1) Write 0AH to D8.(2) Read D8. Status format is as follows:
	D7 D6 D5 D4 D3 D2 D1 D0
•	IR Line: 7 6 5 4 3 2 1 0
	Be sure to reset ISR bit at end-of-interrupt when in the following modes: Auto-Rotating (both types) and Special Mask. To reset ISR in OCW2, write:
	D7 D6 D5 D4 D3 D2 D1 D0
	BCD identifies bit to be reset.

Table 3-19. PIC Operation Procedures

Interrupt Mask Register	To set mask bits in OCW1, write the following mask byte to D9:									
	D7 D6 D5 D4 D3 D2 D1 D0									
	IR Bit Mask: M7 M6 M5 M4 M3 M2 M1 M0 $1 =$ Mask Set, $0 =$ Mask Reset									
	To read mask bits, read D9.									
Special Mask Mode	The Special Mask Mode enables desired bits that have been previously masked; lower priority bits are also enabled.									
	To set, write 68H to D8.									
	To reset, write 48H to D8.									

Table 3-19. PIC Operation Procedures (Continued)

Table 3-20. Typical PIC Interrupt Request Register Read Subroutine

;RR0 READS PIC INTERRUPT REQUEST REGISTER. ;USES-NOTHING; DESTROYS-A.								
 RR0:	MVI OUT IN RET	A,0AH 0D8H 0D8H	OCW3 RR INSTRUCTION TO PIC					
	END							

Table 3-21. Typical PIC In-Service Register Read Subroutine

;RISO RE ;USES-N	;RIS0 READS PIC IN-SERVICE REGISTER. ;USES-NOTHING; DESTROYS-A.									
RIS0:	MVI OUT IN RET	A,08H 0D8H 0D8H	OCW3 RIS INSTRUCTION TO PIC							
	END									

Table 3-22. Typical PIC Set Mask Register Subroutine

;SMASK S ;A ONE MA ;USES-A; [;SMASK STORES A REG INTO PIC MASK REGISTER. ;A ONE MASKS OUT AN INTERRUPT, A ZERO ENABLES IT. ;USES-A; DESTROYS-NOTHING. SMASK: OUT 0D9H RET END	
SMASK:	OUT RET	0D9H
	END	

;RMASK F ;USES-NC	EADS PIC THING; DE	MASK REGISTER. STROYS-A.	· · · · · · · · · · · · · · · · · · ·	
RMASK:	IN RET	0D9H		
	END			

Table 3-23. Typical PIC Mask Register Read Subroutine



;EOI ISSUE ;USES-NOT	S END-OF-INT THING; DESTR	1C.	
EOI:	MVI OUT RET	A,20H 0D8H	NON-SPECIFIC EOI
	END		

NOTE

The data bus buffer configuration does not allow the CPU to read back data it has previously written into the data bus buffer register. Also, the UPI cannot input data from the data bus buffer register that it previously output. The CPU can only read data that the UPI has output and the UPI can only input data that the CPU has written. The 4-bit status register indicates the status of the data bus buffer register and implements the handshaking protocol required for two-way data transfer. The four bits comprising the status register are defined as follows:

a. Bit 0: Output Buffer Full. The OBF flag is automatically set when the UPI loads the data bus buffer register and cleared when the CPU reads the data bus buffer register.



•				
;IN41 R ;USES-I				
IN41:	IN ANI JZ IN RET	0E5H 1 IN41 0E4H	;INPUT STATUS ;CHECK OBF FLAG ;WAIT IF EMPTY	
	END			

Table 3-25. Typical UPI Data Byte Read Subroutine



 ;OUT 41 :USES-N					
OUT41:	IN ANI JZ OUT	0E5H 2 OUT41 0E4H	;INPUT STATUS ;CHECK IBF FLAG ;WAIT IF NOT READY	•	
	RET			T	
	END				

- b. Bit 1: Input Buffer Full. The IBF flag is set when the CPU writes a character to the data bus buffer register and cleared when the UPI inputs the contents of the data bus buffer register to its accumulator.
- c. Bit 2: F0. This general purpose flag, which can be cleared or toggled under UPI software control, is used to transfer UPI status information to the CPU.
- d. Bit 3: F1 (Control/Data). This flag is set to the condition of the A0 input line when the CPU writes a character to the data bus buffer register. This flag can be cleared or toggled under UPI software control.

Tables 3-25 and 3-26, respectively, provides typical routines for inputting and outputting a data byte from and to the UPI.

3-47. INTERRUPT HANDLING

The iSBC 80/30 includes fige interrupt inputs: TRAP, RST 7.5, RST 6.5, RST 5.5, and INTR. The three "restart" interrupts (7.5, 6.5, and 5.5) are maskable; the TRAP is also a "restart" interrupt but is nonmaskable.

The RST 7.5, RST 6.5, and RST 5.5 interrupts cause the internal execution of an RST instruction if the interrupts are enabled and if the interrupt mask is not set by a previously executed SIM instruction. The nonmaskable TRAP interrupt causes the execution of an RST instruc-

tion independent of the state of the interrupt enable or interrupt mask. The priority and vector location of each of the restart interrupts are given in table 3-27.

Table 3-27. 8085A CPU Restart Interrupt Vectors

Interrupt	Vector Location	Priority		
TRAP	24	Highest		
RST 7.5	3C	2nd		
RST 6.5	34	3rd		
RST 5.5	2C	4th		

3-48. TRAP INPUT

There are special considerations that must be understood when the TRAP interrupt is used. The fact that the TRAP interrupt is nonmaskable can present problems in at least two areas.

Interrupt driven systems often contain parameters that must be modified only within critical regions. A critical region can be roughly defined as a section of code that once begun must complete execution before it or another critical regin that corresponds to the same system parameter(s) can be executed. A TRAP interrupt handler cannot safely alter such parameters either directly or indirectly by causing the execution of procedures or tasks that may alter such parameters. If the hardware generates a TRAP interrupt on power up or power fail, the system must be able to process the TRAP interrupt before it is completely initialized. It should also take into account that an interrupt routine that runs with interrupts disabled can still be interrupted by a TRAP.

Because of these considerations, it is recommended that the TRAP interrupt only be used for system startup and/or catastrophic error handling.

It should be noted that TRAP does not destroy a previously established interrupt enable status. Executing the first RIM instruction following a TRAP interrupt yields the interrupt enable status as it was before the TRAP occurred. Following this first *mandatory* RIM instruction, subsequently executed RIM instructions provide current interrupt enable status.

3-49. RST 7.5, 6.5, 5.5 INPUTS

These interrupts can be individually masked by a SIM instruction and can thus be prevented from interrupting the CPU. The priorities shown in table 3-27 does not take into account the priority of a routine that was started by a higher priority interrupt. An RST 5.5 interrupt can interrupt an RST 7.5 routine if the interrupts are *re-enabled* before the end of the RST 7.5 routine.

The RST 6.5 and 5.5 interrupts are *high level sensitive* and, in order to be recognized, must remain high. The

RST 7.5 interrupt is *rising edge sensitive* and only a pulse is required to set the interrupt request; this request will be remembered until the request is serviced or reset by a SIM instruction.

The TRAP interrupt, which is both edge and level sensitive, must have a leading (positive-going) edge and then remain high until serviced.

3-50. INTR INPUT

The INTR input from the 8259A PIC has the lowest priority and is sampled only during the last clock cycle of a given instruction. When INTR is active, the CPU Program Counter (PC) is inhibited and three bytes, timed by INTA/ pulses, are transferred from the PIC to the CPU:

a. Byte 1 = CALL instruction (11001101)

b. Byte 2 = Lower byte of routine address

c. Byte 3 =Upper byte of routine address.

The 16-bit routine address must be on a 32-byte boundard or 64-byte boundary as determined by the Initialization Command Words previously written to the PIC. (Refer to paragraph 3-41.)

The INTR is enabled or disabled by software. It is disabled by "reset" and immediately after an interrupt is accepted.



4-1. INTRODUCTION

This chapter provides a functional description and a circuit analysis of the iSBC 80/30 Single Board Computer. Figures 4-1 and 4-2, located at the end of this chapter, are simplified foldout block diagrams that illustrate the functional interface between the 8085A microprocessor (CPU) and the on-board facilities and between the CPU and the system facilities via the Multibus. Also shown in figure 4-2 is the dual port control logic that allows the iSBC 80/30 to function in a master/slave relationship with the Multibus to allow another bus master to access the on-board RAM.

4-2. FUNCTIONAL DESCRIPTION

A brief description of the functional blocks comprising the iSBC 80/30 is given in following paragraphs. An operational circuit analysis is given beginning with paragraph 4-14.

4-3. CLOCK CIRCUITS

The clock circuit composed of A12, A13, and A21 is stabilized by a 22.1184-MHz crystal. This circuit, which provides the various clock frequencies required for onboard activities, generates a power-up Reset signal to initialize the system to a known internal state; a Reset signal can also be initiated by a signal supplied via auxiliary connector P2.

The clock circuit composed of A55 and A57 is stabilized by an 18.432-MHz crystal. This frequency is divided by two to provide the 9.22-MHz Bus Clock (BCLK/) and Constant Clock (CCLK/) to the Multibus. (The BCLK/ signal is also used by Bus Controller A67.) Removable jumpers are provided to allow this on-board clock circuit to be disabled if some other source supplies BCLK/ and CCLK/ to the Multibus.

4-4. CENTRAL PROCESSOR UNIT

The 8085A Microprocessor (CPU), which is the heart of the single board computer, performs system processing functions and generates the address and control signals required to access memory and I/O devices. The ADO-AD7 pins are used to multiplex the 8-bit input/output data and the lower eight bits of the address. During the first part of a machine cycle, for example, the lower eight bits of address are strobed into Latch A23 by the Address Latch Enable (ALE) signal; the outputs of A23 are combined with the upper eight bits of the address to form the 16-bit address bus. During the remainder of the machine cycle, AD0-AD7 pins of the CPU are used for data input/ output.

4-5. INTERVAL TIMER

The 8253 Programmable Interval Timer (PIT) includes three independently controlled counters that provide optional (jumper selectable) timing inputs to the on-board I/O devices and the CPU interrupts. The clock frequency of 2.46 MHz, 1.23 MHz, or 153.6 kHz, which is derived from the clock circuit composed of A12, A13, and A21, provides the basic timing input.

Counter 2 provides timing for the serial I/O port (8251A USART). This counter, in conjunction with the USART, can provide programmable Baud rates from 110 to 9600. Counter 0 can be used as a timing input to the optional 8041/8741 UPI which, in turn, can connect this signal to its own internal timer input. Counter 0, if not employed by an optional UPI, can be used as an interval timer to generate a CPU interrupt. Counter 1, which is the system interval timer and can also generate a CPU interrupt, has a range of 1.6 microseconds to 853.3 milliseconds. If longer times are required, Counters 0 and 1 can be cascaded to provide a single timer with a maximum delay of more than 50 hours.

4-6. SERIAL I/O

The 8251A Universal Synchronous/Asynchronous Receiver/Transmitter (USART) provides RS232C compatibility and is configured as a data terminal. Synchronous or asynchronous mode, character size, parity bits, stop bits, and Baud rates are all programmable. Data, clocks, and control lines to and from connector J3 are buffered with drivers and receivers.

A second serial I/O channel may be derived from programming the optional 8041/8741 UPI to simulate the USART. This second channel is limited in speed and the number of RS232C interface lines (one driver and one receiver) but is sufficient to communicate with a simple CRT interface.

4-7. PARALLEL I/O

The 8255A Programmable Peripherel Interface provides 24 programmable I/O lines. A bidirectional bus driver is included to interface eight of the I/O lines to connector J1. Two IC sockets are provided so that, depending on the application, TTL drivers or I/O terminators may be installed to complete the interface to connector J1. The 24 lines are grouped into three ports of eight lines each; these ports can be programmed to be simple I/O ports or strobed I/O ports with handshaking. One port can be programmed as a bidirectional port with control lines. The iSBC 80/30 includes various optional features such as RS232C interface lines, timer gates, and interrupts that can be controlled by the parallel I/O lines.

4-8. UNIVERSAL PERIPHERAL INTERFACE

The optional 8041/8741A Universal Peripheral Interface (UPI) is a complete single chip microcomputer that interfaces directly to the 8085A CPU through the I/O structure. The UPI can interface devices such as a printer controller or a keyboard scanner while relying on the CPU only for data transfer.

4-9. INTERRUPT CONTROL

The interrupt section provides 12 priority interrupts. Eight interrupts connect to the inputs of the 8259 Programmable Interrupt Controller (PIC); the remaining four interrupts connect directly to the 8085A CPU. All interrupts, except TRAP, are maskable; the TRAP interrupt is intended to handle a power fail (PFI/) signal input via auxiliary connector P2.

There are 18 jumper-selectable interrupt sources: PPI (2), UPI (1), USART (2), PIT (2), external via J1 and J2 (2), power fail via P2 (1), and Multibus (8).

4-10. ROM/EPROM CONFIGURATION

IC sockets A25 and A37 are provided for user installation of ROM/EPROM chips. Jumpers are provided to accommodate either 1K, 2K, or 4K chips. The on-board ROM/ EPROM address space begins at 0000. This space is normally reserved for ROM/EPROM use only; however, ROM/EPROM may be disabled through an optional jumper connect on the 8255A PPI output port.

4-11. RAM CONFIGURATION

The iSBC 80/30 includes 16K of dynamic RAM implemented with eight Intel 2117 chips and an Intel 8202 Dynamic RAM Controller. Dual-port control logic interfaces this RAM with the Multibus so that the iSBC 80/30 can perform as a slave RAM device when not acting as a bus master. The slave RAM decode logic allows extended Multibus addressing of up to 20 address lines. This allows bus masters with 20-bit addressing capability to partition the iSBC 80/30 into any 8K or 16K segment in a 1-megabyte address space. The 8085A CPU, however, has only a 16-bit address space. Hardware jumpers allow the 8085A CPU or the system to access either 8K or 16K of on-board RAM.

4-12. BUS INTERFACE

The interface to the Multibus includes an Intel Bus Controller, bidirectional address and data bus drivers, and the bus interrupt driver/receivers. The bus controller allows the iSBC 80/30 to operate as a bus master in a serial or parallel priority arrangement with other bus masters in the system in which the 8085A CPU can request the Multibus only when a bus resource is needed.

4-13. DUAL PORT CONTROL

The dual-port control logic allows the iSBC 80/30 to function as a slave RAM device when not acting as a bus master. The dual-port control logic enables or disables the internal data and address buses, depending on whether the access is from the 8085A CPU or from the Multibus. For RAM access, the 8085A CPU has priority over a Multibus request.

An extended addressing facility is provided to allow the 16K RAM to reside within a 1-megabyte address space when accessed from the Multibus. This is useful when one or more bus masters have a 20-bit addressing capability. The 8085A CPU, however, can only access this memory in the lowest 64K address space.

4-14. CIRCUIT ANALYSIS

The schematic diagram for the iSBC 80/30 is given in figure 5-2. The schematic diagram consists of nine sheets, each of which includes grid coordinates. Signals that transverse from one sheet to another are assigned grid coordinates at both the signal source and signal destination. For example, the grid coordinates 2ZB1 locate a signal source (or signal destination as the case may be) on sheet 2 Zone B1.

Both active-high and active-low signals are used. A signal mnemonic that ends with a virgule (e.g., DAT7/) denotes that the signal is active low (≤ 0.4 V). Conversely, a signal mnemonic without a virgule (e.g., ALE) denotes that the signal is active high (≥ 2.0 V).

Figures 4-1 and 4-2 at the end of this chapter are simplified block diagrams of the input/output, interrupt, and memory sections. These block diagrams will be help-ful in understanding both the addressing scheme and the internal bus structure of the board.

4-15. INITIALIZATION

When power is applied in a start-up sequence, the contents of the 8085A CPU program counter, instruction register, and interrupt enable flip-flop are subject to random factors and cannot be predicted. For this reason, a power-up sequence is used to set the CPU, bus controller, and I/O ports to a known internal state.

When power is initially applied to the iSBC 80/30, capacitor C7 (7ZA7) begins to charge through resistor R2. The charge developed across C7 is sensed by a Schmitt trigger, which is internal to Clock Generator A13. The Schmitt trigger converts the slow transition appearing at pin 2 into a clean, fast-rising synchronized RESET output signal at pin 1. The RESET signal is inverted by A29-11

(2ZC7) to produce the Initialize signal INIT/. The INIT/ signal clears the CPU program counter, instruction register, and interrupt enable flip-flop; resets the parallel I/O ports to the input mode; resets the serial I/O port to the "idle" mode; clears the UPI internal program counter and status flags; resets the bus controller; and, via the Multibus, sets the system to a known internal state.

The initialization described above can be performed at any time by inputting an AUX RESET/ signal via connector P2.

4-16. CLOCK CIRCUITS

The time base for Bus Clock signal BCLK/ and Constant Clock signal CCLK/ is provided by Clock Generator A57 (5ZA6) and crystal Y2. The 18.43-MHz output of A57 is divided by A55 and driven onto the Multibus through jumpers 165-166 and 167-168. The BCLK/ signal is also used as the clock input to Bus Controller A67.

The time base for all other functions on the board is provided by Clock Generator A13 and crystal Y1. The crystal frequency of 22.12-MHz appearing at the OSC output of A13 is buffered and applied to the dual port control logic (9ZC8) and to RAM Controller A66 (3Z5D). The 22.12-MHz OSC output is also divided by A44 and A31 to develop the 5.53-MHz clock for 8085 CPU A36 (2ZD6) and for optional UPI A20 (5Z4D).

The crystal frequency is divided by nine internally by A13 to produce a 2.46-MHz clock at its $\Phi 2$ TTL output. This signal clocks Divider A12 and supplies a selectable clock signal to PIT A21. Divider A12 divides the clock input by two and by 16 to produce 1.23-MHz and 153.6-kHz selectable clocks for PIT A21 and the parallel I/O port.

4-17. 8085A CPU TIMING

The 8085A CPU internally divides the 5.53-MHz clock input by two to develop the timing requirements for the various time-dependent functions. These functions are described in following paragraphs.

4-18. INSTRUCTION TIMING. The execution of any program consists of read and write operations, where each operation transfers one byte of data between the CPU and memory or between the CPU and an I/O device. Although the CPU can vary the address, data, type, and sequence of operations, it is capable of performing only a basic read or write operation. With the exception of a few control lines, such as Address Latch Enable (ALE), these read and write operations are the only communication necessary between the CPU and the other components to execute any instruction.

An *instruction cycle* is the time required to fetch and execute an instruction. During the fetch phase, the

selected instruction (consisting of up to three bytes) is read from memory and stored in the operating registers of the CPU. During the execution phase, the instruction is decoded by the CPU and translated into specific processing activities.

Each instruction cycle consists of up to five machine cycles. A *machine cycle* is required each time the CPU accesses memory or an I/O device. The fetch phase requires one machine cycle for each byte to be fetched. Some instructions do not require any machine cycles other than those necessary to fetch the instructions from memory; other instructions, however, require an additional machine cycle(s) to write or read data to or from memory or I/O devices.

Every instruction cycle has at least one reference to memory during which time the instruction is fetched. An instruction cycle must always have a fetch, even if the execution of that instruction requires no reference to memory. The first machine cycle in every instruction cycle is therefore a fetch, and beyond that there are no specific rules. For instance, the IN (input) and OUT (output) instructions each require three machine cycles: fetch (to obtain the instruction), memory read (to obtain the I/O address of the device), and an input or output machine cycle (to complete the transfer).

Each machine cycle consists of a minimum of three and a maximum of six states designated T_1 through T_6 . A *state* is the smallest unit of processing activity and is defined as the interval between two successive falling edges of the CPU clock. Each state (or CPU clock cycle) has a duration of 350 nanoseconds (derived by dividing the 5.53-MHz frequency by two).

Every machine cycle normally consists of three T-states with the exception of an opcode fetch, which consists of either four or six T-states. The actual number of states required to execute any instruction depends on the instruction being executed, the particular machine cycle within the instruction cycle, and the number of *wait* states inserted into the machine cycle. The wait state state is initiated when the READY input to the CPU is pulsed low.

There are no wait states imposed when the CPU is addressing on-board ROM/PROM or on-board I/O. There will be wait states imposed, however, in the following operations:

- a. When addressing on-board RAM and a refresh cycle in progress.
- b. When addressing on-board RAM and a slave memory read or write is in progress. (One wait state is always imposed when performing a write to on-board RAM.)
- c. When addressing system RAM, PROM, or I/O devices.

Figure 4-3 is presented to show the relationship between an instruction cycle, machine cycle, and T-state. This example shows the execution of a Store Accumulator Direct (STA) instruction involving on-board memory. Notice that for this instruction the opcode fetch (machine cycle M_1) requires four T-states and the remaining three cycles each require three T-states.

The opcode fetch is the only machine cycle that requires more than three T-states. This is because the CPU must interpret the requirements of the opcode fetched during T_1 through T_3 before it can decide what must be done in the remaining T-state(s).

There are seven types of machine cycles, each of which can be differentiated by the states of three CPU status lines (IO/\overline{M} , S0, and S1) and three CPU control lines (\overline{RD} , \overline{WR} , and \overline{INTA}). Table 4-1 lists the states of the CPU status and control lines during each of the seven machine cycles and during a CPU halt.

Table 4-1. CPU Status and Control Lines

		Status	5	Control		
Machine Cycle	10/M	S0	S1	RD	WR	INTA
Opcode Fetch Memory Read Memory Write I/O Read I/O Write INTR Acknowledge Bus Idle Halt	0 0 1 1 X TS	1 0 1 0 1 X 0	1 1 0 1 0 1 X 0	0 0 1 0 1 1 TS	1 0 1 0 1 TS	1 1 1 1 0 1

4-19. OPCODE FETCH TIMING. Figure 4-4 shows the timing relationship of a typical opcode fetch machine cycle. At the beginning of T_1 of *every* machine cycle, the CPU performs the following:

- a. Pulls IO/\overline{M} low to signify that the machine cycle is a memory reference operation.
- b. Drives status lines S0 and S1 high to identify the machine cycle as an opcode fetch.
- c. Places high-order bits (PCH) of program counter onto address lines A8-A15. These address bits will remain true until at least T_4 .
- d. Places low-order bits (PCL) of program counter onto address/data lines AD0-AD7. These address bits will remain true for only one clock cycle, after which AD0-AD7 go to their high-impedance state as indicated by the dashed line in figure 4-3.
- e. Activates the Address Latch Enable (ALE) signal.

At the beginning of T_2 , the CPU pulls the RD/ line low to enable the addressed memory device. The device will then drive the ADO-AD7 lines. After a period of time, as determined by the access time of the addressed memory device, valid data (the DCX instruction in this example) will be present on the D0-D7 lines. During T_3 the CPU loads the data on the D0-D7 lines into its instruction register and drives RD/ high, disabling the addressed memory device. During T_4 the CPU decodes the opcode and decides whether or not to enter T_5 on the next clock cycle or start a new machine cycle and enter T_5 and then T_6 before beginning a new machine cycle.

Figure 4-5 is identical to figure 4-4 with one exception, which is the use of the READY input to the CPU. As shown in figure 4-5, the CPU examines the state of the



Figure 4-3. Typical CPU Instruction Cycle







Figure 4-5. Opcode Fetch Machine Cycle (With Wait)

READY input during T_2 . If the READY input is high, the CPU will proceed to T_3 as shown in figure 4-4. If the READY input is low, however, the CPU will enter the T_{wait} state and stay there until READY goes high. When READY goes high, the CPU will exit the T_{wait} state and enter T_3 . The external effect of using the READY input is to preserve the exact state of the CPU signals at the end of T_3 for an integral number of clock periods before finishing the machine cycle. This 'stretching' of the system timing,

in effect, increases the allowable access time for memory or I/O devices. By inserting T_{wait} states, the CPU can accommodate slower memory or slower I/O devices. It should be noted, however, that access to the on-board ROM/PROM and I/O ports does not impose a T_{wait} state. However, as mentioned previously, T_{wait} states are imposed in certain instances when accessing on-board RAM; T_{wait} states are always imposed when accessing system memory or I/O devices via the Multibus. 4-20. MEMORY READ TIMING. Figure 4-6 shows the timing of two successive memory read machine cycles, the first without a Twait state and the second with one $T_{wait}\xspace$ state. Disregarding the states of the S0 and S1 lines, the timing during T_1 through T_3 is identical with the opcode fetch machine cycle shown in figure 4-4. The major difference between the opcode fetch and memory read cycles is that an opcode fetch machine cycle requires four or six T-states whereas the memory read machine cycle requires only three T-states. One minor difference between the two cycles is that the memory address used for the opcode fetch cycle is always the contents of the program counter (PC), which points to the current instruction; the address used for a memory read cycle can be one of several origins. Also, the data read from memory is placed into the appropriate register istead of the instruction register. Note that a Twait state is not imposed during a read of on-board ROM/PROM.

4-21. I/O READ TIMING. Figure 4-6 also illustrates the timing of two successive I/O read machine cycles, the first without a T_{wait} state and the second with one T_{wait} state. With the exception of the IO/ \overline{M} status signal, the timing of a memory read cycle and an I/O read cycle is identical. For an I/O read, IO/ \overline{M} is driven high to identify that the current machine cycle is referencing an I/O port. One other minor exception is that the address used for an

I/O read cycle is derived from the second byte of an IN instruction; this address is duplicated onto both the A8-A15 and AD0-AD7 lines. The data read from the I/O port is always placed in the accumulator specified by the IN instruction. Note that a T_{wait} is not imposed during the access of on-board I/O devices; T_{wait} states are imposed during the access of system I/O devices via the Multibus.

4-22. MEMORY WRITE TIMING. Figure 4-7 shows the timing of two successive memory write machine cycles, the first without a Twait state. Again, disregarding the states of the S0 and S1 lines, the timing during T_1 is identical to the timing of an opcode fetch. memory read, and I/O read cycles. The difference occurs, however, at the end of T_1 . For instance, in a memory read cycle the AD0-AD7 lines are disabled (high impedance) at the beginning of T_2 in anticipation of the returned data. In a memory write cycle, the AD0-AD7 lines are not disabled and the data to be written into memory is placed on these lines at the beginning of T_2 . The Write (WR/) line is driven low at this time to enable the addressed memory device. During T_2 the READY input is checked to determine if a T_{wait} state is required. If the READY input is low, Twait states are inserted until READY goes high. During T_3 , the WR/ line is driven high to disable the addressed memory device and terminate the memory write operation. Note that the contents on the address and data lines do not change until the next T₁ state.



Figure 4-6. Memory Read (or I/O Read) Machine Cycles

4-23. I/O WRITE TIMING. Figure 4-7 also illustrates the timing of two successive I/O write machine cycles, the first without a T_{wait} state and the second with one T_{wait} state. With the exception of the IO/M status signal, the timing of a memory write cycle and an I/O cycle are identical.

4-24. INTERRUPT ACKNOWLEDGE TIM-ING. Figure 2-8 shows the CPU timing in response to the INTR input being driven high by PIC A30 (assuming the CPU interrupt enable flip-flop has been set by a previously executed Enable Interrupt instruction). The status of the TRAP, RST 7.5, RST 6.5, RST 5.5, and INTR inputs are sampled during CLK of the T-state immediately preceding T_1 of M_1 . If INTR was the only valid interrupt, the CPU clears its interrupt enable flip-flop and enters the Interrupt Acknowledge (INA) machine cycle. With two exceptions, the INA machine cycle is identical with the Opcode Fetch (OF) machine cycle. The first exception is that $IO/\overline{M} = 1$, which signifies that the opcode fetched will be from an I/O device. The second exception is that INTA is asserted instead of RD. Although the contents of CPU program counter is sent out on the address lines, the address lines are ignored.

When INTA is asserted, the PIC provides a CALL instruction which causes the CPU to push the contents of the program counter onto the stack before jumping to a new location. After receiving the CALL opcode, the CPU perform a second INA machine cycle (M_2) to access the second byte of the CALL instruction from the PIC. The timing of M_2 is identical with M_1 except that M_2 has three T-states. M_2 is followed by M_3 to access the third byte of the CALL instruction. When all three bytes have been received, the CPU executes the instruction. The CPU inhibits the incrementing of the program counter during the three INA cycles so that the correct program counter value can be pushed onto the stack during M_4 and M_5 .

During M_4 and M_5 , the CPU performs Memory Write (MW) machine cycles to write (push) the contents of the program counter onto the top of the stack. The CPU then places the two bytes accessed during M_2 and M_3 into the upper and lower bytes of the program counter. This has the same effect as jumping the execution of the program to the location specified by the CALL instruction.

After the interrupt service routine is executed, the CPU pops the stack and loads it into the program counter, and resumes system operation at the point of the interrupt. (It is the programmer's responsibility to ensure that the interrupt enable flip-flop is set before returning from the service routine.)



Figure 4-7. Memory Write (or I/O Write) Machine Cycles
,

CICNAL	M2	(MR)				M ₁ (INA)					M2 (1NA)	
SIGNAL	т2	тз	т ₁	т2	TWAIT	т ₃	т ₄	т ₅	т ₆	т ₁	т2	тз
CLK		\bigvee							<u>\</u>			
INTR	$\int $				-							-
INTA				<u> </u>							<u>\</u>	
IO/M, \$1, \$0	(0,	(,0)	X		(1,	 1,1) 					(1,1,1)	
A ₈ -A ₁₅	(PC-	і)н 		P	 сн 						РСН	
AD ₀ -AD ₇	IN D ₀ -	,		\succ	IN D ₀ -D ₇ (CAL							N D ₇ (B2)
ALE												
RD												
WR												
	1	M3 (INA)			M4 (MVV)			M ₅ (MW)			M1 (OF)	
SIGNAL	т1	т2	т3	т ₁	т2	. т _з	тı	т2	т ₃	T ₁	т2	TWAIT
CLK	\bigtriangledown		\backslash			\backslash		\backslash				
INTR		-										
INTA		\										
10/M, S1, S0		(1,1,1)		X	(0,0,1)		X	(0,0,1)		X	(0,1,1)	
A ₈ -A ₁₅		РСН			(SP-1)H			(SP-2)H			PCH(B3)	
AD ₀ -AD ₇			N P7 (B3)	(SP-1)		ώт D ₇ (РСН)	OUT (SP-2)L		UT D ₇ (PCL)	DUT B2	\succ	IN D ₀ .D ₇
ALE						ļ/	$ \land$			$ \land $		
RD										<u> </u>		
WR				-	 	Ļ_/		<u> </u>	Ļ_⁄_			

NOTE: DISREGARD TWAIT STATES; NO TWAIT STATES IMPOSED ON 80/30 APPLICATION OF INTA.



4-25. ADDRESS BUS

The address bus is shown in weighted lines in figures 4-1 and 4-2. The lower eight bits (AD0-AD7) of the memory address or I/O address (depending on whether a memory reference machine cycle or an I/O reference machine cycle is in progress) are output by the CPU during the first clock cycle (T_1). The CPU AD0-AD7 pins become the source to or input from the data bus during the second and third cycles (T_2 and T_3). The trailing edge of the Address Latch Enable (ALE) signal issued by the CPU during T_1 strobes these eight address bits into Latch A23 (2ZD3). The lower eight address bits (AB0-AB7) from A23 are placed on the iSBC 80/30 address bus together with the high-order address bits (AB8-AB15). These address bits are decoded by the following:

- a. AB2-AB7 to I/O Address Decoder A50 (6ZA7).
- b. ABA-ABF to PROM Address Decode Logic (sheet 3).
- c. AB0-ABB to PROM A25/A37 (3ZA3).
- d. ABD-ABF to On-Board RAM Decoder A17 (3ZD7).

4-26. BUS TIME OUT

Bus Time Out one-shot A38 (1ZC7) is retriggered by the leading edge of the ALE signal, which is asserted by the CPU during T_1 of every machine cycle. If the CPU is halted, or if the CPU is hung up in a wait state for approximately 10 milliseconds, A38 times out and asserts the BTMO signal on pin 34 of the auxiliary bus (P2). If jumper 115-116 is installed, the BUS TIME OUT/ signal (A38 pin 6) drives the CPU READY line high through gates A39 and A28 and allows the CPU to exit the wait state.

The BUS TIME OUT/ signal is also applied to interrupt jumper matrix post 122 (7ZC3). If jumpered to the CPU RST 7.5 input and A38 is retriggered following a CPU hang-up state, the rising edge of BUS TIME OUT/ will cause an interrupt to report this event.

4-27. DATA BUS

The CPU AD0-AD7 pins become the source or destination of the data bus DB0-DB7 during T_2 and T_3 clock cycles. Data can be sourced to or input from the following:

- a. Data Buffer A24 (4ZC6).
- b. Data Buffer A52 (9ZD6).

4-28. READ/WRITE SIGNAL GENERATION

The COMMAND/ signal, which is used in conjunction with the various on-board read/write operations, is generated by A45-3 (2ZB4) when the CPU RD/ or WR/ signal is true. The following paragraphs describe how the various I/O and memory control signals are generated.

4-29. I/O CONTROL SIGNALS. The I/O control signals (2ZC2) are derived simply by gating the status of the CPU IO/M, RD/, and WR/ pins. For example, the I/O ADR signal is the buffered IO/ \overline{M} pins status; the I/O WRT/ signal is the logical AND of the IO/ \overline{M} and WR/ pin status.

4-30. MEMORY CONTROL SIGNALS. The MEM ADR, MEMRD/, and MEMWR/ signals (2ZB2, 2ZC2) are also derived simply by gating the status of the CPU IO/\overline{M} , RD/, and WR/ pins. For example, the MEM ADR signal is the buffered and inverted IO/\overline{M} pin status; the MEMWR/ signal is the logical AND of the WR/ and inverted IO/\overline{M} pin status.

The ADV MEM RD/ and ADV MEM WRT/ are derived by multiplexing the CPU S0 and S1 pin status. Refer to table 4-1. During all memory references, the CPU IO/ \overline{M} status pin is low and activates the gate input to 8:4 Multiplexer A35. The select pin of A35 is controlled by the S0 pin status; when S0 = 0, the "A" inputs are selected and, when S0 = 1, the "B" inputs are selected.

For a memory read operation, S0 = 0 and S1 = 1 and a logic 1 appears on A35 pin 12. The trailing edge of the CPU ALE signal clocks Quad "D" Flip-Flop A34 and A34 pin 14 asserts the ADV MEM RD/ signal.

For a memory write operation, S0 = 1 and S1 = 0 and a logic 1 appears on A35 pin 7. The trailing edge of ALE clocks A34 and A34 pin 11 asserts the ADV MEM WRT/ signal.

For a memory fetch operation, S0 = 1 and S1 = 1 and, since the S1 status is applied to both the 4A and 4B inputs, the ADV MEM RD/ signal is generated as described above for the memory read operation.

When the read, write, or fetch operation is complete, the CPU RD/ or WR/ pin goes false and Flip-Flop A44 (2ZB3) is clocked and set. The output at A44-6 clears A34; the next rising edge of ALE clears A44.

4-31. DUAL PORT CONTROL LOGIC

The Dual Port Control logic (figure 5-2 sheet 9) allows the on-board RAM facilities to be shared by the on-board CPU and another bus master via the Multibus. When not acting as a bus master or when not accessing its on-board RAM, the iSBC 80/30 can act as a "slave" RAM device in a multiple bus master system. When accessing its on-board RAM, the on-board CPU has priority over any attempt to access the on-board RAM via the Multibus. In this situation, the bus access is held off until the CPU has

completed its particular read or write operation. When a bus access is in progress, the Dual Port Control Logic enters the "slave" mode and any subsequent CPU request will be held off until the slave mode is terminated. Figures 4-9 and 4-10 are timing diagrams for the Dual Port Control logic.



Figure 4-9. Dual Port Control Bus Access Timing With CPU Lockout

611-9





611-10

4-11

4-32. BUS ACCESS TIMING. Figure 4-9 illustrates Dual Port Control timing for RAM access via the Multibus. (P-periods P0 through P17 are used only for descriptive references.) When SLAVE RAM RQT/ and SLAVE RAM RD/ or SLAVE RAM WRT/ are true, A43-15 goes high and A43-6 goes low on the next rising edge of the clock at the end of P0. (The purpose of A43-15 is to synchronize the asynchronous command and the purpose of A43-6 is to prevent a subsequent CPU request from aborting a bus access after A55-9 goes high at the end of P1).

At the end of P1, A55-9 goes high and A55-8 goes low; A55-8 asserts the SLAVE MODE signal. The outputs of A55-8 and A43-6 are ANDed to hold A55-9 in the preset (high) state. At the end of P2, A43-11 goes low and asserts the OFF BD CS/ signal, which drives the PCS/ input low to RAM Controller A66. The output of A43-11 also enables pin 1 of A60-7, A60-3, A60-5, and A60-9, to gate the RAM RD/ or RAM WRT/ signal to A66. Approximately one-half to two clock cycles later, A66 asserts the RAM AACK/ signal and, at the end of P3, A43-15 goes low.

The RAM Controller asserts RAM XACK/ during P13 and SLAVE XACK/ in driven onto the Multibus. The bus master then first raises the SLAVE RAM RD/ or SLAVE RAM WRT/ signal and then the SLAVE RAM RQT/ signal. When SLAVE RAM RD/ or WRT/ goes false, the RAM Controller raises RAM AACK/ and RAM XACK/. At the end of P15, A43-6 is clocked high. At the end of P16, A55-9 goes low and A59-8 goes high. At the end of P17, A43-11 goes high and terminates the OFF BD CS/ signal.

The foregoing discussion pertains only to the operation of the Dual Port Control for bus access of on-board RAM. The actual addressing and the transfer of data are discussed in paragraph 4-42.

4-33. CPU ACCESS TIMING. Figure 4-10 illustrates the Dual Port Control timing for RAM access by the on-board CPU. (P-periods P0 through P13 are used only for descriptive purposes.) To demonstrate that the CPU has priority in the access of on-board RAM, figure 4-10 shows that the SLAVE RAM RQT/ and a SLAVE RAM RD/ or SLAVE RAM WRT/ are active when the CPU access is initiated. The timing has progressed through P0 during which time A43-15 has been clocked high and A43-6 has been clocked low.

When the ON BD RAM RQT and ADV MEM RD/ are asserted, the ADV MEM RD/ signal is driven through A60-15 to assert the RAM RD/ signal. The RAM Controller then asserts RAM AACK/, which is driven through A62-8 to produce RAM QAACK/. At the end of P1, A43-15 is clocked low to effectively abort the slave access timing cycle. (For a write cycle, the MEM WR/ signal is used instead of ADV MEM WRT/. During P12 the RAM controller asserts RAM XACK/ (not used by the CPU) and, during P13, the ADV MEM RD/ and ON BD RAM go false to complete the CPU access. On the following rising edge of the clock, the SLAVE RAM RQT/ and SLAVE RAM RD/ or SLAVE RAM WRT/ signals, which have been held off during the CPU access, set A43-15 once again to initiate the slave access timing cycle.

The foregoing discussion pertains only to the operation of the Dual Port Control for CPU access of on-board RAM. The actual addressing and the transfer of data are discussed in paragraph 4-41.

4-34. MULTIBUS INTERFACE

The Multibus interface consists of Bus Controller A67 (8ZD5), bidirectional Address Bus Driver A72/A73 (8ZA2), bidirectional Data Bus Driver A74/A75 (8ZB2), and the Slave RAM Address Decode Logic (figure 5-2 sheets 3 and 5).

The Bus Controller allows the iSBC 80/30 to assume the role of a bus master and includes bus arbitration, timing, and read/write command logic. The falling edge of BCLK/ provides the timing reference, and bus arbitration begins when the Qualified Command (QCMD/) signal is asserted and the address decoders have determined that the associated Read or Write Command is not intended for on-board I/O or memory. The QCMD/ signal (1ZD8), which is used only for Multibus requests, is derived directly from the CPU Read (RD/) signal or derived by delaying the CPU Write (WT/) signal half a clock cycle. (Delaying WT/ ensures the adequate setup of the address and data to the Bus Drivers before the Write Command (IORC/ or MWTC/) is asserted.

The QCMD/ signal activates the Transfer Start Request (XSTR) input to the Bus Controller, which drives BREQ/ low and BPRO/ high. The BREQ/ output from each bus master in the system is used by the Multibus when the bus priority is resolved by a parallel priority scheme as described in paragraph 2-27. The BPRO/ output is used by the Multibus when the bus priority is resolved by a serial priority scheme as described in paragraph 2-26.

The iSBC 80/30 gains control of the Multibus when the BPRN/ input to the Bus Controller is driven low. On the next falling edge of BCLK), the Bus Controller drives BUSY/ and ADEN/ low. The BUSY/ output indicates that the bus is in use and that the current bus master in control will not relinquish control until it raises its BUSY/ signal. The ADEN/ output, which can be thought of as a "master bus control" signal, enables Address Bus Driver A72/A73 and Data Bus Driver A74/A75. Since the board is not in the slave mode, the QSLAVE RQT/ signal is false (high) and the Address Bus Driver transmit function is selected. The steering logic composed of A39-6,

Principles of Operation

A59-6, A42-6, and A41-8 (8ZB4) decides which function (transmit or receive) to select for the Data Bus Driver. This decision is based on the mode (master or slave) that the board is in and whether the command is a read or a write. For the master mode, a Write Command selects the transmit function (DIEN is driven high); a Read Command selects the receive function (DIEN is driven low).

The BUS Controller examines its IORR, IOWR, MRDR, and MWTR inputs and drives the appropriate command line low on the Multibus. After the command is acknowledged (signified by the addressed device driving AACK/ or XACK/ low), the CPU terminates the appropriate command and tristates its address lines. This deselects the Address Bus Drivers and Data Bus Drivers and causes QCMD/ to go false (high). The Bus Controller relinquishes control of the Multibus by driving BREQ/ high and BPRO/ low and then raising BUSY/.

It should be noted that, after gaining control of the Multibus, the iSBC 80/30 can invoke an override condition to prevent losing control at a critical time. (For instance, it may be desired to execute several consecutive commands without having to contend for the bus after each command is executed.) Bus override is invoked by executing a SIM instruction with accumulator bits 6 and 7 = 1, which causes the CPU SOD line to drive the Bus Controller OVRD input high.

4-35. I/O OPERATION

The following paragraphs describe on-board and system I/O operations. The actual functions performed by specific read and write commands to on-board I/O devices are described in Chapter 3.

4-36. ON-BOARD I/O OPERATION. Address bits AB2–AB7 are applied to I/O Address Decoder A50 and associated input chip enable gates (6Z7A). Address bits AB4–AB7 are decoded by A48–12, A65–6, and A65–3 to provide E1, E2, and E3 enable inputs to A50; when enabled, A50 decodes address bits AB2–AB4 to provide chip select inputs to the appropriate I/O device. Address bits AB2–AB7 are decoded as follows:

Bits	Addrosooo	Chip Select
765432	Addresses	Signal
1 1 0 1 1 0 1 1 0 1 1 1 1 1 1 0 0 1 1 1 0 0 1 1 1 0 1 0	D8-DB DC-DF E4-E7 E8-EB EC-EF	8259CS/ 8253CS/ 8741CS/ 8255CS/ 8251CS/

When any one of the five outputs of A50 goes low, the I/O ADR signal from the CPU (buffered IO/M) is driven through A46–11 to develop I/O AACK/. The I/O AACK/ signal drives the CPU READY line high and, together

with the COMMAND/ signal (decoded WR or RD), enables Data Buffer A24 (4ZC6). The function of A24 is selected by the CPU S1 output; S1 = 1 for read operations and S1 = 0 for write operations. For example, data is transferred from the DIO0–DIO7 lines to the DB0–DB7 lines when S1 = 1.

After the I/O device has been selected by address bits AB2–AB7, specific functions for the chip are selected by address bits AB0–AB1. (Refer to table 3-2.)

4-37. SYSTEM I/O OPERATION. Address bits AB2–AB7 are decoded by I/O Address Decoder A50 as described in paragraph 4-31. If the address is not for an on-board I/O device, I/O AACK/ remains false (high) and, together with QCMD/, activates the Bus Controller XSTR input. The false SLAVE MODE/ signal from A55-9 (9ZC4) in the Dual Port Control logic enables Address Buffer A53/A54 and, together with the false I/O AACK/ signal, enables Data Buffer A52.

When the Bus Controller gains control of the Multibus as described in paragraph 4-29, it drives ADEN/ low to select Address Bus Driver A72/A73 and Data Bus Driver A74/A75. Since the board is not in the slave mode, the QSLAVE RQT/ signal is false (high) and the Address Bus Driver transmit function is selected. The transmit or receive function of Data Buffer is selected by the CPU S1 line. If the operation is a read, S1 = 1 and the receive function is enabled; for a write, S1 = 0 and the transmit function is enabled. The steering logic composed of A39-6, A58-6, A42-6, and A41-8 (8ZB4) decides which function to select for the Data Bus Driver. Since the board is in the master mode for off-board operations, a write operation selects the transmit function and a read operation selects the receive function.

After gaining control of the Multibus the Bus Controller proceeds with the control tasks for the remainder of the I/O transfer. Refer to paragraph 4-29 for a description of how the Bus Controller terminates bus control.

4-38. ROM/EPROM OPERATION

The two ROM/EPROM chips are installed by the user in IC sockets A25/A37 (3ZA3). Memory addresses 0000-1FFF are reserved exclusively for ROM/EPROM; the actual occupied memory space depends on the ROM/ EPROM chips as follows:

Chip	Chip Addresses In				
Size	A25	A3			
1K × 8 2K × 8 4K × 8	0000-03FF 0000-07FF 0000-0FFF	0400-07FF 0800-0FFF 1000-1FFF			

Address bits AB0-ABB are applied directly to the inputs of A25-A37 and address bits ABA-ABF are applied to the

ROM/EPROM chip select decoders. When address bits ABA-ABF are true for the address ranges specified above, PROM AACK/ signal is true; when the MEMRD/ signal is also asserted during a read operation, the PROM ENABLE/ and PROM AACK/ signals are true. PROM ENABLE/ turns on Data Buffer A24 (4ZC6), whose transmit function is enabled by CPU S1 = 1, and PROM AACK/ drives the CPU READY line high via A28-6 (1DZ5). The decoded output of A49-8 selects the ROM/EPROM chip in A25; the decoded output of A49-6 selects the ROM/EPROM chip in A37. When the chip is enabled, the contents of the location specified by ABO-ABB are transferred to the CPU via Data Buffer A24.

4-39. RAM OPERATION

As described in paragraph 4-31, the Dual Port Control logic allows the on-board RAM facilities to be shared by the on-board CPU and by another bus master via the Multibus. The following paragraphs describe briefly the RAM Controller and the overall operation of how the RAM is addressed for read/write operation.

4-40. RAM CONTROLLER. All address and control inputs to the on-board RAM is supplied by RAM Controller A66 (3ZD5). The RAM Controller provides a 64-cycle RAS/CAS refresh timing cycle to dynamic RAM chips A77-A84. Default jumper 110-111 holds the REFRESH RQT signal false and the RAM Controller operates in the automatic refresh mode. In the automatic refresh mode, a read or write request can be delayed if a refresh cycle is in progress. Option jumper position 110-106 allows the REFRESH RQT signal to be controlled by the CPU READY line and thereby implement the invisible refresh mode. In the invisible refresh mode, the RAM Controller works around memory accesses by refreshing RAM during the CPU instruction decode clock cycle which follows each instruction fetch. Refresh occurs during T₄ but, if the access is greater than the refresh time, the RAM Controller will automatically refresh RAM. Thus, the RAM will not generate CPU wait states but it will consume more power.

The RAM Controller, when enabled with a low input to its PCS/ pin, multiplexes the address to the RAM chips. Low-order address bits A0-A6 are presented at the RAM input pins and RAS/ is driven low at the beginning of the first 8202 clock cycle after RAM RD/ or RAM WRT/ is active. High-order address bits A7-A13 are presented at the RAM input pins and CAS/ is driven low during the second clock cycle after the RAS/ address hold time is satisfied.

The RAM Controller then examines its RD/ and WRT/ inputs. If RD/ is low, the RAM Controller drives its WE/ output high to provide a Read signal to RAM; if WRT/ is low, the RAM Controller drives its WE/ output high to provide a write signal to RAM. If RAM RD/ is asserted, Memory Data Buffer A76 is enabled and RAM data is latched into A76 when RAM XACK/ is subsequently driven low.

When the memory cycle begins, the RAM Controller drives its SACK/ output low; SACK/ is used as the RAM AACK/ signal. When the cycle is complete (i.e., data is valid), drives its XACK/ output low. The SACK/ and XACK/ outputs go high when the RD/ or WRT/ input goes high.

4-41. ON BOARD READ/WRITE OPERA-TION. When MEM ADR goes true, Address Decoder A17 (3ZD7) decodes address bits ABD-ABF. The decoded output from A17 goes low and (1) asserts ON BD RAM RQT/ to the Dual Port Control logic and (2) drives the RAM Controller PCS/ input low. The RAM Controller then multiplexers the address to RAM and, depending on which input command (RAMRD/ or RAMWRT/) is true, drives the WE/ output pin high or low. (The WE/ pin is driven high for a read and driven low during a write.) The SACK/ signal, which is asserted at the beginning of the memory cycle, generates the RAM AACK/ signal. When RAM AACK/ goes true, the Dual Port Control logic generates the RAM QAACK/ signal, which is qualified by being in the master mode. The XACK/ signal, which is asserted when the data is valid, generates the RAM XACK/ signal.

When RAM QAACK/ goes true, the READY input is driven high to the CPU to prevent a wait state. By the time the CPU timing has progressed to T3 of the memory cycle, the RAM Controller has already asserted XACK/ and the data is valid; i.e., the data has been written into RAM from the data bus or has been read from RAM onto the data bus. The XACK/ signal, however, is not used by the CPU during on-board RAM access.

4-42. BUS READ/WRITE OPERATION. When another bus master has control of the multibus, that bus master can address the iSBC 80/30 as a slave RAM device. Assuming that the bus master has a 20-bit addressing capability, ADR0/-ADRF/ and ADR10/-ADR13/ are placed on the Multibus and then either MWTC/ or MRDC/ is asserted. Address bits ADR10/-ADR13/ are decoded by A69 (5ZC6) and ADRD/-ADRF/ are decoded by A68 (5ZB5). (A description of 16-bit and 20-bit address assignment for bus access of RAM is given in paragraphs 2-17 through 2-19.) The decoded output of A69 provides an enable input to A68. The decoded output of A68 drives the SLAVE RAM RQT/ signal low to the Dual Port Control logic.

Assuming no CPU access of RAM is in progress, the SLAVE MODE/ signal is driven low and address bits ADR0/-ADRF/ are transferred through Address Buffers

A72/A73 to the RAM Controller. When the Dual Port Control logic asserts the OFF BD CS/ signal, the RAM Controller PCS/ input is enabled and, depending on which command has been asserted (RAMRD/ or RAMWRT/), drives the WE/ input high to RAM.

The RAM Controller generates SACK/ and XACK/ as previously described. For a bus access, the SACK/ signal develops SLAVE AACK/ and XACK/ develops SLAVE XACK/. The controlling bus master then drives the

MRDC/ or MWTC/ command high and deactivates address drivers A72/A73.

4-43. INTERRUPT OPERATION

All interrupts except INTR are connected to the CPU by jumper connections. (Refer to paragraph 2-20.) Since interrupt handling is handled by the internal CPU timing described in paragraph 4-24, no further explanation is considered necessary.



Figure 4-1. iSBC 80/30 Input/Output and Interrupt Block Diagram



NOTE:

Figure 4-2. iSBC 80/30 PROM and Dual Port RAM Block Diagram



CHAPTER 5 SERVICE INFORMATION

5-1. INTRODUCTION

This chapter provides a list of replaceable parts, service diagrams, and service and repair assistance instructions for the iSBC 80/30 Single Board Computer.

5-2. REPLACEABLE PARTS

Table 5-1 provides a list of replaceable parts for the iSBC 80/30. Table 5-2 identifies and locates the manufacturers specified in the MFR CODE column in table 5-1. Intel parts that are available on the open market are listed in the MFR CODE column as "COML"; every effort should be made to procure these parts from a local (commercial) distributor.

5-3. SERVICE DIAGRAMS

The iSBC 80/30 parts location diagram and schematic diagram are provided in figures 5-1 and 5-2, respectively. On the schematic diagram, a signal mnemonic that ends with a slash (e.g., IOWC/) is active low. Conversely, a signal mnemonic without a slash (e.g., BTMO) is active high.

5-4. SERVICE AND REPAIR ASSISTANCE

United States customers can obtain service and repair assistance from Intel by contacting the MCD Technical Support Center in Santa Clara, California, at one of the following numbers: Telephone:

From Alaska or Hawaii call — (408) 987-8080 From locations within California call toll free — (800) 672-3507 From all other U.S. locations call toll free — (800) 538-8014 TWX: 910-338-0026 TELEX: 34-6372

Always contact the MCD Technical Support Center before returning a product to Intel for service or repair. You will be given a "Repair Authorization Number", shipping instructions, and other important information which will help Intel provide you with fast, efficient service. If the product is being returned because of damage sustained during shipment from Intel, or if the product is out of warranty, a purchase order is necessary in order for the MCD Technical Support Center to initiate the repair.

In preparing the product for shipment to the MCD Technical Support Center, use the original factory packaging material, if available. If the original packaging is not available, wrap the product in a cushioning material such as Air Cap TH-240 (or equivalent) manufactured by the Sealed Air Corporation, Hawthorne, N.J., and enclose in a heavy-duty corrugated shipping carton. Seal the carton securely, mark it "FRAGILE" to ensure careful handling, and ship it to the address specified by MCD Technical Support Center personnel.

NOTE

Customers outside of the United States should contact their sales source (Intel Sales Office or Authorized Intel Distributor) for directions on obtaining service or repair assistance.

Reference Designation	Description	Mfr. Part No.	Mfr. Code	Qty.
A1,2,74,75	IC, 8226, Bidirectional Data Buffer	8266	COML	4
A3-11	IC, 7400, Quad 2-Input Positive NAND	SN7400	I TI	9
A12	IC, 74163, Synchronous 4-Bit Counter	SN74163	TI	1
A13,57	IC, 8224, System Clock Generator	8224	COML	2
A14	IC, 1488, Quad Line Driver	LM1488	NAT	1
A15,16	IC, 1489, Quad Line Driver	LM1489	NAT	2
A17,50	IC, 74LS138, 3-to-8 Decoder//Multiplexer	SN74LS138	TI	2
A18	IC, 74LS11, Triple 3-Input Positive NAND	SN74LS11	TI	1
A19	IC, 8255A, Programmable Peripheral Interface	8255A	COML	1
A21	IC, 8253, Programmable Interval Timer	8253	COML	1
A22	IC, 8251A, Serial I/O Interface (USART)	8251A	COML	1
A23	IC, 74LS373, Octal D-Type Latches	SN74LS373	TI	1
A24,52,72,73	IC. DP8304. Bidirectional Data Buffer	DP8304	COML	4
A26	IC, 74S03, Quad 2-Input Positive NAND	SN74S03	TI TI	1
A27,32	IC, 74LS32, Quad 2-Input OR	SN74LS32	דו	2

Table 5-1. Replaceable Parts

Reference Designation	Description	Mfr. Part No.	Mfr. Code	Qty.
A28 A29 A30 A31 A33,40,51,63 A34,43 A35 A36 A38 A39,45 A41 A42,62 A44,55 A46,58,61,64 A47 A46,58,61,64 A47 A53,54 A56 A59 A60 A65 A59 A60 A65 A66 A67 A68,69 A70,71 A76 A77-84	IC, 74S20, Dual Positive NAND IC, 7438, Quad 2-Input Positive NAND IC, 8259A, Programmable Interrupt Controller IC, 74LS74, Dual D-Type Flip-Flops IC, 74S04, Hex Inverters IC, 74S175, Quad D-Type Flip-Flops IC, 74S175, Data Selector/Multiplexer IC, 8085A Central Processor Unit IC, 9602, One-Shot Multivibrator IC, 74LS08, Quad 2-Input Positive AND IC, 74S08, Quad 2-Input Positive AND IC, 74S08, Quad 2-Input Positive AND IC, 74S04, Quad 2-Input Positive NAND IC, 74S04, Quad 2-Input Positive NAND IC, 74S04, Quad 2-Input Positive NAND IC, 74S11, Triple 3-Input Positive NAND IC, 74LS27, Triple 3-Input Positive NAND IC, 74LS240, Octal Buffer/Line Driver IC, 74LS02, Quad 2-Input Positive NAND IC, 74LS240, Octal Buffer/Line Driver IC, 74LS00, Quad 2-Input Positive NAND IC, 74LS00, Quad 2-Input P	SN74S20 SN7438 8259A SN74LS74 SN74S04 SN74S175 SN74LS157 8085A 9602 SN74LS08 SN74S08 SN74S08 SN74S08 SN74S00 SN74S11 SN74LS27 SN74S10 SN74LS27 SN74S10 SN74LS27 SN74LS02 8097 SN74LS00 8202 OBD 3205 SN74LS04 SN74LS04 SN74S373 2117	TI FI COML TI FI FI COML NAT FI FI FI FI FI FI FI FI FI COML NAT FI FI FI FI FI FI FI FI COML COML COML COML	1 1 1 4 2 1 1 2 2 4 1 1 1 2 1 1 1 1 1 2 2 1 8
C1–6,19,21,29,30,31,32 36–44,46,48,50–53, 55,60–63	Capacitor, Ceramic, 0.01µF, 25V, +80%, -20%	OBD	COML	32
C7 C8,49 C9,10,22–28,33,35,45, 47,54,56,58,59,64–72, 79,81,83,85,87,89, 91,94	Capacitor, tant, 10μ F, 20V, 10% Capacitor, mono, 10pF, 500V, 5% Capacitor, mono, 0.1μ F, 50V, +80%, -20%	OBD OBD OBD	COML COML COML	1 1 35
C57 C73-77,93 C78,82,86,90 C80,84,88,92 C95 CR1	Capacitor, mono, 220pF, 500V, 5% Capacitor, tant, 22 μ F, 15V, 10% Capacitor, mono, 1.0 μ F, 50V, 10% Capacitor, mono, 0.01 μ F, 50V, +80%, -20% Capacitor, mono, 33 μ F, 50V Diode, 1N4002	OBD OBD OBD OBD OBD 1N4002	COML COML COML COML COML TI	1 6 4 1 1
Q1	Transistor, 2N3904	2N3904	ті	1
R1,3-5,10,11,14,17-20,	Resistor, comp, 1K ohm, 1/4W 5%	OBD	COML	19
R2 R6-8,15,16,30,34,35 R9,12,13 R13 R21 R22 R24 R31 RP1,2	Resistor, comp, 100K ohm, ¼W 5% Resistor, comp, 10K ohm, ¼W 5% Resistor, comp, 620K ohm, ¼W 5% Resistor, comp, 5.1K ohm, ¼W 5% Resistor, comp, 330K ohm, ¼W 5% Resistor, comp, 33K ohm, ¼W 5% Resistor, comp, 270K ohm, ¼W 5% Resistor, comp, 2.2K ohm, ¼W 5% Resistor, package, 1K ohm, 10-pin	OBD OBD OBD OBD OBD OBD OBD OBD OBD	COML COML COML COML COML COML COML COML	1 8 3 1 1 1 1 1 2
XA1,2 XA3-11 XA20,36,66 XA25,37	Socket, IC, 16-pin Socket, IC, 14-pin Socket, IC, 40-pin Socket, IC, 24-pin	C-93-16-02 C-93-14-02 540-A367D C-93-24-02	TI TI AUG TI	2 9 1 2
Y1 Y2	Crystal, 22.1184-MHz Crystal, 18.432-MHz Post, Wire Wrap Extractor, Card	HW3 MP184 85931-6 S-203	CTS CTS AMP SCA	1 1 192 2

Table 5-1	Renlaceable	Parts	(Continued)	١
Table 3-1	Replaceable	1 al 13	(Commucu)	,

MFR. CODE	MANUFACTURER	ADDRESS	MFR. CODE	MANUFACTURER	ADDRESS
AMP	AMP, Inc.	Harrisburg, PA	NAT	National Semiconductor	Santa Clara, CA
AUG	Augat, Inc.	Attleboro, MA	SCA	Scanbe, Inc.	El Monte, CA
CTS	CTS Corp.	Elkhart, IN	TI	Texas Instruments	Dallas, TX
			OBD	Order by Description; commercial source	available from any

Table 5-2. List of Manufacturers' Codes



Service Information

Figure 5-1. iSBC 80/30 Parts Location Diagram

5-5/5-6



- 1

Service Information

Figure 5-2. iSBC 80/30 Schematic Diagram (Sheet 1 of 9)

POWER



P1 14 P2 17



CPU

5-9/5-10





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Service Information



MEMORY

5-11/5-12



Figure 5-2. iSBC 80/30 Schematic Diagram (Sheet 4 of 9) テアエ



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	11] POET	۲ I	Э				
 []	3			PORT	۰۱	4				
	<u>ل</u> م			POR	т١	5				
D PIN	8		36	PORT	· I	مک				
KET	11			PORT	۰.	7				
6.] Ø	3	·	18	סד [
PIN	۵		20	וד [с
	100			- 41RS2	320	UT 6ZBB				
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d L			G	PORT	12	5				
PIN	<u> </u>		4	POR	ΓZ	۵				
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	.1				•					

Figure 5-2. iSBC 80/30 Schematic Diagram (Sheet 5 of 9)

CPU/UPI BUS & CONSTRUCT CLOCKS SLAVE RAY AUUK, DECOLES-15/5-16





VSART



Service Information



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1

Figure 5-2. iSBC 80/30 Schematic Diagram (Sheet 8 of 9)

BUS CONTROL/BUS DRIVERS

5-21/5-22





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5-25/5-26



Figure 5-4. iSBC 614 Schematic Diagram

EXPAPSION BACKPLANE 5-27/5-28



APPENDIX A 8085 INSTRUCTION SET

A computer, no matter how sophisticated, can only do what it is "told" to do. One "tells" the computer what to do via a series of coded instructions referred to as a **Pro**gram. The realm of the programmer is referred to as **Soft**ware, in contrast to the Hardware that comprises the actual computer equipment. A computer's software refers to all of the programs that have been written for that computer.

When a computer is designed, the engineers provide the Central Processing Unit (CPU) with the ability to perform a particular set of operations. The CPU is designed such that a specific operation is performed when the CPU control logic decodes a particular instruction. Consequently, the operations that can be performed by a CPU define the computer's Instruction Set.

Each computer instruction allows the programmer to initiate the performance of a specific operation. All computers implement certain arithmetic operations in their instruction set, such as an instruction to add the contents of two registers. Often logical operations (e.g., OR the contents of two registers) and register operate instructions (e.g., increment a register) are included in the instruction set. A computer's instruction set will also have instructions that move data between registers, between a register and memory, and between a register and an I/O device. Most instruction sets also provide Conditional Instructions. A conditional instruction specifies an operation to be performed only if certain conditions have been met; for example, jump to a particular instruction if the result of the last operation was zero. Conditional instructions provide a program with a decision-making capability.

By logically organizing a sequence of instructions into a coherent program, the programmer can "tell" the computer to perform a very specific and useful function.

The computer, however, can only execute programs whose instructions are in a binary coded form (i.e., a series of 1's and 0's), that is called Machine Code. Because it would be extremely cumbersome to program in machine code, programming languages have been developed. There

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are programs available which convert the programming language instructions into machine code that can be interpreted by the processor.

One type of programming language is Assembly Language. A unique assembly language mnemonic is assigned to each of the computer's instructions. The programmer can write a program (called the Source Program) using these mnemonics and, certain operands; the source program is then converted into machine instructions (called the Object Code). Each assembly language instruction is converted into one machine code instruction (1 or more bytes) by an Assembler program. Assembly languages are usually machine dependent (i.e., they are usually able to run on only one type of computer).

THE 8085 INSTRUCTION SET

The 8085 instruction set includes five different types of instructions:

- Data Transfer Group move data between registers or between memory and registers
- Arithmetic Group add, subtract, increment or decrement data in registers or in memory.
- Logical Group AND, OR, EXCLUSIVE-OR, compare, rotate or complement data in registers or in memory
- Branch Group conditional and unconditional jump instructions, subroutine call instructions and return instructions
- Stack, I/O and Machine Control Group includes I/O instructions, as well as instructions for maintaining the stack and internal control flags.

Instruction and Data Formats:

Memory for the 8085 is organized into 8-bit quantities, called Bytes. Each byte has a unique 16-bit binary address corresponding to its sequential position in memory. The 8085 can directly address up to 65,536 bytes of memory, which may consist of both read-only memory (ROM) elements and random-access memory (RAM) elements (read/ write memory).

Data in the 8085 is stored in the form of 8-bit binary integers:

$$\begin{bmatrix} D_7 & D_6 & D_5 & D_4 & D_3 & D_2 & D_1 & D_0 \end{bmatrix}$$

$$MSB \qquad \qquad LSB$$

When a register or data word contains a binary number, it is necessary to establish the order in which the bits of the number are written. In the Intel 8085, BIT 0 is referred to as the Least Significant Bit (LSB), and BIT 7 (of an 8 bit number) is referred to as the Most Significant Bit (MSB).

The 8085 program instructions may be one, two or three bytes in length. Multiple byte instructions must be stored in successive memory locations; the address of the first byte is always used as the address of the instructions. The exact instruction format will depend on the particular operation to be executed.



Addressing Modes:

Often the data that is to be operated on is stored in memory. When multi-byte numeric data is used, the data, like instructions, is stored in successive memory locations, with the least significant byte first, followed by increasingly significant bytes. The 8085 has four different modes for addressing data stored in memory or in registers:

- Direct Bytes 2 and 3 of the instruction contain the exact memory address of the data item (the low-order bits of the address are in byte 2, the high-order bits in byte 3).
- Register The instruction specifies the register or register-pair in which the data is located.
- Register Indirect The instruction specifies a register-pair which contains the memory

address where the data is located (the high-order bits of the address are in the first register of the pair, the low-order bits in the second).

 Immediate – The instruction contains the data itself. This is either an 8-bit quantity or a 16-bit quantity (least significant byte first, most significant byte second).

Unless directed by an interrupt or branch instruction, the execution of instructions proceeds through consecutively increasing memory locations. A branch instruction can specify the address of the next instruction to be executed in one of two ways:

- Direct The branch instruction contains the address of the next instruction to be executed. (Except for the 'RST' instruction, byte 2 contains the low-order address and byte 3 the high-order address.)
- Register indirect The branch instruction indicates a register-pair which contains the address of the next instruction to be executed. (The high-order bits of the address are in the first register of the pair, the low-order bits in the second.)

The RST instruction is a special one-byte call instruction (usually used during interrupt sequences), RST includes a three-bit field; program control is transferred to the instruction whose address is eight times the contents of this three-bit field.

Condition Flags:

There are five condition flags associated with the execution of instructions on the 8085. They are Zero, Sign, Parity, Carry, and Auxiliary Carry, and are each represented by a 1-bit register in the CPU. A flag is "set" by forcing the bit to 1; "reset" by forcing the bit to 0.

Unless indicated otherwise, when an instruction affects a flag, it affects it in the following manner:

- Zero: If the result of an instruction has the value 0, this flag is set; otherwise it is reset.
- Sign: If the most significant bit of the result of the operation has the value 1, this flag is set; otherwise it is reset.
- Parity: If the modulo 2 sum of the bits of the result of the operation is 0, (i.e., if the result has even parity), this flag is set; otherwise it is reset (i.e., if the result has odd parity).
- Carry: If the instruction resulted in a carry (from addition), or a borrow (from subtraction or a comparison) out of the highorder bit, this flag is set; otherwise it is reset.

The first (high-order) register of a designated

The second (low-order) register of a desig-

16-bit program counter register (PCH and

PCL are used to refer to the high-order and

16-bit stack pointer register (SPH and SPL

are used to refer to the high-order and low-

Bit m of the register r (bits are number 7

register pair.

Z,S,P,CY,AC The condition flags:

nated register pair.

low-order 8 bits respectively).

order 8 bits respectively).

through 0 from left to right).

Auxiliary Carry: If the instruction caused a carry out of bit 3 and into bit 4 of the resulting value, the auxiliary carry is set; otherwise it is reset. This flag is affected by single precision additions, subtractions, increments, decrements, comparisons, and logical operations, but is principally used with additions and increments preceding a DAA (Decimal Adjust Accumulator) instruction.

Symbols and Abbreviations:

The following symbols and abbreviations are used in the subsequent description of the 8085 instructions:

SYMBOLS	MEANING			Zero,
				Sign,
accumulator	Register A			Parity,
addr	16-bit address quantity			Carry,
data	8-bit data quantity	·		and Auxiliary Carry, respectively.
data 16	16-bit data quantity		()	The contents of the memory location or reg- isters enclosed in the parentheses.
byte 2	The second byte of the instructi	on	◄	"Is transferred to"
byte 3	The third byte of the instruction	n .	\wedge	Logical AND
port	8-bit address of an I/O device		\forall	Exclusive OR
r,r1,r2	One of the registers A,B,C,D,E,H	۱,L	\vee	Inclusive OR
DDD,SSS	The bit pattern designating one	of the regis-	+	Addition
	source):	nation, 333-	_	Two's complement subtraction
			*	Multiplication
	DDD or SSS REGISTER	RNAME	* *	"Is exchanged with"
	111 A 000 B			The one's complement (e.g., (\overline{A}))
	001 C		n	The restart number 0 through 7
	010 D		NNN	The binary representation 000 through 111
	011 E			for restart number 0 through 7 respectively.

rh

rl

PC

SP

rm

rp

RP

101 One of the register pairs:

100

B represents the B,C pair with B as the highorder register and C as the low-order register;

Н

L

D represents the D,E pair with D as the highorder register and E as the low-order register;

H represents the H,L pair with H as the highorder register and L as the low-order register;

SP represents the 16-bit stack pointer register.

The bit pattern designating one of the register pairs B,D,H,SP:

RP	REGISTER PAIR
00	B-C
01	D-E
10	H-L
11	SP

Description Format:

The following pages provide a detailed description of the instruction set of the 8085. Each instruction is described in the following manner:

- 1. The MCS 85[™]macro assembler format, consisting of the instruction mnemonic and operand fields, is printed in BOLDFACE on the left side of the first line.
- 2. The name of the instruction is enclosed in parenthesis on the right side of the first line.
- 3. The next line(s) contain a symbolic description of the operation of the instruction.
- 4. This is followed by a narative description of the operation of the instruction.
- 5. The following line(s) contain the binary fields and patterns that comprise the machine instruction.

6. The last four lines contain incidental information about the execution of the instruction. The number of machine cycles and states required to execute the instruction are listed first. If the instruction has two possible execution times, as in a Conditional Jump, both times will be listed, separated by a slash. Next, any significant data addressing modes (see Page A-2) are listed. The last line lists any of the five Flags that are affected by the execution of the instruction.

Data Transfer Group:

This group of instructions transfers data to and from registers and memory. Condition flags are not affected by any instruction in this group.

MOV r1, r2 (Move Register)

(r1) ← (r2)

The content of register r2 is moved to register r1.

0	1	D	D	D	S	S	S
	-	(Addr	Cycles: States: essing: Flags:	1 4 regis non	ster		

MOV r, M (Move from memory)

(r) 🔶 ((H) (L))

The content of the memory location, whose address is in registers H and L, is moved to register r.

0	1		D	1	1	0
		Cycles:	2			
		Addressing:	, req.	indire	ct	

Flags: none

MOV M, r (Move to memory)

((H) (L)) ← (r)

The content of register r is moved to the memory location whose address is in registers H and L.

	0	1	1	1	0	S	s	S			
			(Cvcles:	2	-					
	States: 7										
			Addr	essing:	reg.	reg. indirect					
				Flags:	non	none					
	Inemo	nics Co	pyrighte	l © Inte	el Corpor	ation 1	976,197	7			
A-4							•				

MVI r, data (Move Immediate) (r) ← (byte 2) The content of byte 2 of the instruction is moved to register r.

0	Т	0	D	D	D	1	1	0
				da	ta			
			(Cycles:	2			
				States:	7			
			Addr	essina:	imn	nediate		

Flags:

none

MVI M, data (Move to memory immediate) ((H) (L)) ← (byte 2)

> The content of byte 2 of the instruction is moved to the memory location whose address is in registers H and L.

0	0	1	1	1	0	I	1	1	1	l	0
			d	ata							
Cycles: 3											

States: 10 Addressing: immed./reg. indirect Flags: none

LXI rp, data 16 (Load register pair immediate) (rh) - (byte 3),

(rl) - (byte 2)

Byte 3 of the instruction is moved into the high-order register (rh) of the register pair rp. Byte 2 of the instruction is moved into the low-order register (rl) of the register pair rp.

0	0	R	Ρ	0	0	0	1
low-order data							
high-order data							
			Cycles	3			

Cycles:	3
States:	10
Addressing:	immediate
Flags:	none

LDA addr (Load Accumulator direct)

(A) ← ((byte 3)(byte 2))

The content of the memory location, whose address is specified in byte 2 and byte 3 of the instruction, is moved to register A.

0	0	1	1	1	0	1	0
high-order addr							
	Cycles:						
		:	States:	13			
		Addr	essing:	direc	t		
			Flags:	none			

STA addr (Store Accumulator direct)

((byte 3)(byte 2)) - (A)

The content of the accumulator is moved to the memory location whose address is specified in byte 2 and byte 3 of the instruction.

0	0	1	1	0	0	1	0
		lo	ow-ord	er addr			
high-order addr							

Cycles: 4 States: 13 Addressing: direct Flags: none

- (L) ((byte 3)(byte 2))
- (H) ((byte 3)(byte 2) + 1)

The content of the memory location, whose address is specified in byte 2 and byte 3 of the instruction, is moved to register L. The content of the memory location at the succeeding address is moved to register H.

0	0	1	0	1	0	1	0	
low-order addr								
high-order addr								
		(Cycles:	5				



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SHLD addr (Store H and L direct) ((byte 3)(byte 2)) ← (L) ((byte 3)(byte 2) + 1) ← (H)

> The content of register L is moved to the memory location whose address is specified in byte 2 and byte 3. The content of register H is moved to the succeeding memory location.

0	0	Ι	1		0	1	0	Т	0	1	0
				lo	w-or	de	r ade	dr			
high-order addr											

Cycles: 5 States: 16 Addressing: direct Flags: none

LDAX rp (Load accumulator indirect)

(A) 🛶 ((rp))

The content of the memory location, whose address is in the register pair rp, is moved to register A. Note: only register pairs rp=B (registers B and C) or rp=D (registers D and E) may be specified.

0	0	R	Р	1	0	1	0
				0			

2
7
reg. indirect
none

STAX rp (Store accumulator indirect)

((rp)) < (A)

The content of register A is moved to the memory location whose address is in the register pair rp. Note: only register pairs rp=B (registers B and C) or rp=D(registers D and E) may be specified.

0	0	R	Р	0	0	1	0
		C	Cycles:	2			
			•	-			

States: 7 Addressing: reg. indirect Flags: none

XCHG (Exchange H and L with D and E)

(H) 🛶 (D)

(L) → (E)

The contents of registers H and L are exchanged with the contents of registers D and E.



Arithmetic Group:

This group of instructions performs arithmetic operations on data in registers and memory.

Unless indicated otherwise, all instructions in this group affect the Zero, Sign, Parity, Carry, and Auxiliary Carry flags according to the standard rules.

All subtraction operations are performed via two's complement arithmetic and set the carry flag to one to indicate a borrow and clear it to indicate no borrow.

ADD r (Add Register)

(A) ← (A) + (r)

The content of register r is added to the content of the accumulator. The result is placed in the accumulator.

1	l o	0	0	0	S	s	s
`		Addı	Cycles: States: ressing: Flags:	1 4 regis Z,S,	ster P,CY,A	٩C	

ADD M (Add memory)

(A) ← (A) + ((H) (L))

The content of the memory location whose address is contained in the H and L registers is added to the content of the accumulator. The result is placed in the accumulator.

1) I	0	0	0	1	I	1	0
		(Cycles:	2				
			States:	7				
		Addr	essing:	reg	. indir	ect		
			Flags:	Z,S	S,P,CY	,AC		

ADI data

(A) ← (A) + (byte 2)

(Add immediate)

The content of the second byte of the instruction is added to the content of the accumulator. The result is placed in the accumulator.

1	Τ	1	0	0	0	Τ	1	Γ	1	Γ	0
				da	ta						_
				Cycles:	2						
				States:	- 7						
			Addı	ressing:	im	med	diate	9			
				Flags:	Ζ,	S,P,	CY,	AC			

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- ADC r (Add Register with carry) (A) \leftarrow (A) + (r) + (CY)
 - The content of register r and the content of the carry bit are added to the content of the accumulator. The result is placed in the accumulator.

1	[0	1	0	0	1	S	s	s
					Cycles:	1			
				• • •	States:	4			
				Add	Flags:	regi: Z.S.	ster .P.CY.A	AC	

ADC M (Add memory with carry)

(A) ← (A) + ((H) (L)) + (CY)

The content of the memory location whose address is contained in the H and L registers and the content of the CY flag are added to the accumulator. The result is placed in the accumulator.

1	Τ	0	Τ	0	0	Τ	1	1	Ţ	1	Т	0
					Cycles	::	2					
					States	::	7					
				Add	ressing	:	reg.	indir	ect	t		
					Flags	::	Z,S	,P,CY	΄,Α	С		

ACI data (Add immediate with carry)

(A) ← (A) + (byte 2) + (CY)

The content of the second byte of the instruction and the content of the CY flag are added to the contents of the accumulator. The result is placed in the accumulator.



States: 7 Addressing: immediate Flags: Z,S,P,CY,AC

SUB r (Subtract Register)

(A) ← (A) – (r)

The content of register r is subtracted from the content of the accumulator. The result is placed in the accumulator.



SUB M (Subtract memory)

(A) ← (A) – ((H) (L))

The content of the memory location whose address is contained in the H and L registers is subtracted from the content of the accumulator. The result is placed in the accumulator.

0 1	0 1 1 0
Cycles: States: Addressing: Elags:	2 7 reg. indirect Z S P C Y AC

SUI data (Subtract immediate)

(A) ← (A) – (byte 2)

The content of the second byte of the instruction is subtracted from the content of the accumulator. The result is placed in the accumulator.

1	Τ	1		0	Т	1	0	1		1	0	
data												
					C	cles:	2					

•	
States:	7
Addressing:	immediate
Flags:	Z,S,P,CY,AC

SBB r (Subtract Register with borrow)

(A) \leftarrow (A) - (r) - (CY)

L

The content of register r and the content of the CY flag are both subtracted from the accumulator. The result is placed in the accumulator.

1	0	0	1	1	S	S	S
		C	ycles:	1			
		S	States:	4			
		Addre	essing:	regi	ster		
			Flags:	Z,S,	P,CY,	٩C	

SBB M (Subtract memory with borrow) (A) \leftarrow (A) – ((H) (L)) – (CY)

The content of the memory location whose address is contained in the H and L registers and the content of the CY flag are both subtracted from the accumulator. The result is placed in the accumulator.

	1	0	0	1	1	1	1	1	1	0
			C	Cycles:	2					
			:	States:	7					
			Addr	essing:	reg.	. indir	ect			
				Flags:	Z,S	,P,CY	,AC	2		
AII	Mnemo	nics Cop	yrighted	l Corpo	ration	197	6,19	77		

SBI data (Subtract immediate with borrow)

(A) \leftarrow (A) - (byte 2) - (CY)

The contents of the second byte of the instruction and the contents of the CY flag are both subtracted from the accumulator. The result is placed in the accumulator.

1	Ι	1	0	Т	1	Т	1	Т	1	1	1	Т	0	
			 		C	lata	1							
			Ado	Cy S dres F	vole: tate ssing lag:	s: s: g: s:	2 7 im Z,S	me S,P	diat ,CY	:e ,AC	2		·	

INR r (Increment Register)

(r) 🖛 (r) + 1

The content of register r is incremented by one. Note: All condition flags **except CY** are affected.

0	0	D	D	D	1	0	1 0
		Addr	Cycles: States: essing: Flags:	1 4 regi Z,S,	ster P,AC		

INR M (Increment memory)

((H) (L)) → ((H) (L)) + 1

The content of the memory location whose address is contained in the H and L registers is incremented by one. Note: All condition flags **except CY** are affected.

0	0	1	1	0	1		0
			Cycles:	3			
			States:	10			
		Add	ressing:	reg.	indire	ect	
			Flags:	Z,S	,P,AC		

DCR r (Decrement Register)

(r) -1

The content of register r is decremented by one. Note: All condition flags **except CY** are affected.

0 0	DD	D 1	0	1
	Cycles: States: Addressing:	1 4 register		

DCR M (Decrement memory)

((H) (L)) ← ((H) (L)) – 1

The content of the memory location whose address is contained in the H and L registers is decremented by one. Note: All condition flags **except CY** are affected.

0	Τ	0	1	1	Ţ	1	Τ	0	Τ	1	I	0	Τ	1	
				Cycles:				3							
				Addressing:			reg	. ir	ndir	ec	t				

Flags: Z,S,P,AC

INX rp (Increment register pair)

(rh) (rl) ← (rh) (rl) + 1

The content of the register pair rp is incremented by one. Note: No condition flags are affected.

0	0	R	Ρ	0	0	1	1	1	
		С	ycles:	1					
		S	States:	6					
		Addre	essing:	regi	ster				
			Flags:	non	е				

DCX rp (Decrement register pair)

(rh) (rl) ← (rh) (rl) – 1

The content of the register pair rp is decremented by one. Note: No condition flags are affected.

0	0	RP	1 0	1	1
		Cycles:	1		
		States:	6		
		Addressing:	register		
		Flags:	none		

DAD rp (Add register pair to H and L) (H) (L) ← (H) (L) + (rh) (rl)

The content of the register pair rp is added to the content of the register pair H and L. The result is placed in the register pair H and L. Note: Only the CY flag is affected. It is set if there is a carry out of the double precision add; otherwise it is reset.

0	0	R	Р	1	0	0	1
		· (Cycles: States:	3 10			
		Addr	essing: Flags:	regis CY	ster		
			~				

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DAA (Decimal Adjust Accumulator)

The eight-bit number in the accumulator is adjusted to form two four-bit Binary-Coded-Decimal digits by the following process:

- 1. If the value of the least significant 4 bits of the accumulator is greater than 9 or if the AC flag is set, 6 is added to the accumulator.
- 2. If the value of the most significant 4 bits of the accumulator is now greater than 9, or if the CY flag is set, 6 is added to the most significant 4 bits of the accumulator.

NOTE: All flags are affected.

0	0	1	10		0	1	1	1	1	1	
			Cycles	::	1						

States: 4 Flags: Z,S,P,CY,AC

Logical Group:

This group of instructions performs logical (Boolean) operations on data in registers and memory and on condition flags.

Unless indicated otherwise, all instructions in this group affect the Zero, Sign, Parity, Auxiliary Carry, and Carry flags according to the standard rules.

ANA r (AND Register)

(A) \leftarrow (A) \wedge (r)

The content of register r is logically anded with the content of the accumulator. The result is placed in the accumulator. The CY flag is cleared and AC is set.

1	Т	0	1	0	0	S	l s	s
			(Sycles:	1			
			:	States:	4			
			Addr	essing:	regis	ter		
				Flags:	Z,S,F	[,] ,CY,/	۹C	

ANA M (AND memory)

(A) ← (A) ∧ ((H) (L))

The contents of the memory location whose address is contained in the H and L registers is logically anded with the content of the accumulator. The result is placed in the accumulator. The CY flag is cleared and AC is set.



ANI data (AND immediate)

(A) ← (A) ∧ (byte 2)

The content of the second byte of the instruction is logically anded with the contents of the accumulator. The result is placed in the accumulator. The CY flag is cleared and AC is set.

1	1	1	0	0	1	1	0
			dat	ta			

Cycles: 2 States: 7 Addressing: immediate Flags: Z,S,P,CY,AC

XRA r (Exclusive OR Register)

 $(A) \longleftarrow (A) \forall (r)$

The content of register r is exclusive-or'd with the content of the accumulator. The result is placed in the accumulator. The CY and AC flags are cleared.

1	0	1	0	1	S	S	s
-		Add	Cycles: States: dressing: Flags:	1 4 regis Z,S,	ster P,CY,A	AC	

XRA M (Exclusive OR Memory)

 $(A) \longleftarrow (A) \forall ((H) (L))$

The content of the memory location whose address is contained in the H and L registers is exclusive-OR'd with the content of the accumulator. The result is placed in the accumulator. The CY and AC flags are cleared.

1	T	0	I	1	T	0	1	1	Ι	1	Ι	1	1	0
					C,	/cles	5:	2				_		
					St	tates	s:	7						
				Add	lres	sing	J:	reg	g. ir	ndire	ect			
					F	lage	5:	Ζ,	S,P	,CY,	,AC	;		

XRI data (Exclusive OR immediate)

(A) \leftarrow (A) \forall (byte 2)

The content of the second byte of the instruction is exclusive-OR'd with the content of the accumulator. The result is placed in the accumulator. The CY and AC flags are cleared.

1		1	Ţ	1		0	Τ	1	Ţ	1	Ţ	1	Τ	0	
						d	ata	3							
					C۷	cles	:	2							



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ORA r (OR Register)

(A) ← (A) V (r)

The content of register r is inclusive-OR'd with the content of the accumulator. The result is placed in the accumulator. The CY and AC flags are cleared.

1	0	1		0	S	S	s
		(Addr	Cycles: States: essing: Flags:	1 4 regis Z,S,F	ter P,CY,A	AC	

ORA M (OR memory)

(A) ← (A) V ((H) (L))

The content of the memory location whose address is contained in the H and L registers is inclusive-OR'd with the content of the accumulator. The result is placed in the accumulator. The CY and AC flags are cleared.

1	0	1	1	0	1	1	0
			Cycles:	2			
			States:	7			
		Add	dressing:	reg.	indire	ct	
			Flags:	Z,S,	P,CY,	AC	

ORI data (OR Immediate)

(A) 🛶 (A) V (byte 2)

The content of the second byte of the instruction is inclusive-OR'd with the content of the accumulator. The result is placed in the accumulator. The CY and AC flags are cleared.

1	1.		1	Τ	1	0	Т	1	Τ	1	0
data											

Cycles: 2 States: 7 Addressing: immediate Flags: Z,S,P,CY,AC

CMP r (Compare Register)

(A) – (r)

The content of register r is subtracted from the accumulator. The accumulator remains unchanged. The condition flags are set as a result of the subtraction. The Z flag is set to 1 if (A) = (r). The CY flag is set to 1 if (A) < (r).

1	1 ₀	1	1	1	S	S	s
			Cycles: States:	1 4			
		Add	lressing: Flags:	regis Z,S,	ster P,CY,A	٩C	

CMP_M (Compare memory)

(A) – ((H) (L))

The content of the memory location whose address is contained in the H and L registers is subtracted from the accumulator. The accumulator remains unchanged. The condition flags are set as a result of the subtraction. The Z flag is set to 1 if (A) = ((H) (L)). The CY flag is set to 1 if (A) < ((H) (L)).

1	0	1	1	T	.1	T	1	T	1	Т	1	0
			Ado	C S dre	ycle tate ssing = lag	s: s: g:	2 7 reg 7	g. ir S P	ndir CY	ect	-	

CPI data (Compare immediate)

(A) - (byte 2)

The content of the second byte of the instruction is subtracted from the accumulator. The condition flags are set by the result of the subtraction. The Z flag is set to 1 if (A) = (byte 2). The CY flag is set to 1 if (A) < (byte 2).

RLC (Rotate left) $(A_{n+1}) \leftarrow (A_n); (A_0) \leftarrow (A_7)$ $(CY) \leftarrow (A_7)$

The content of the accumulator is rotated left one position. The low order bit and the CY flag are both set to the value shifted out of the high order bit position. Only the CY flag is affected.

0	Τ	0	I	0		0		0	Τ	1	Τ	1	ļ	1	
					C	ycles	::	1.							
					S	tates	::	4							
					F	Flags	::	C١	/						

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RRC (Rotate right)

$$(A_n) \leftarrow (A_{n-1}); (A_7) \leftarrow (A_0)$$

(CY) \leftarrow (A_0)

The content of the accumulator is rotated right one position. The high order bit and the CY flag are both set to the value shifted out of the low order bit position. Only the CY flag is affected.



RAL (Rotate left through carry) (A_{n+1}) \leftarrow (A_n); (CY) \leftarrow (A₇)

(A₀) ← (CY)

The content of the accumulator is rotated left one position through the CY flag. The low order bit is set equal to the CY flag and the CY flag is set to the value shifted out of the high order bit. Only the CY flag is affected.

0	0	0	1	0	1	1	1
		C	ycles:	1			
		5	States:	4			
			Flags:	CY			

RAR (Rotate right through carry)

 $\begin{array}{cccc} (A_n) & \longleftarrow & (A_{n+1}) ; & (CY) & \longleftarrow & (A_0) \\ (A_7) & \longleftarrow & (CY) \end{array}$

The content of the accumulator is rotated right one position through the CY flag. The high order bit is set to the CY flag and the CY flag is set to the value shifted out of the low order bit. Only the CY flag is affected.

0	I	0	1	0	Т	1	1	1	Į	1	Ι	1	Τ	1	
					Су	cles	:	1							
					St	ates	:	4							
					F	lags	:	CY							

CMA (Complement accumulator)

(A) ← (Ā)

The contents of the accumulator are complemented (zero bits become 1, one bits become 0). No flags are affected.


CMC (Complement carry)

(CY) ← (<u>CY</u>)

The CY flag is complemented. No other flags are affected.

0	0	1	1	1	1	۱ _{1.}	1
			Cycles: States:	1 4			
			Flags:	CY			

STC (Set carry)

(CY) -- 1

The CY flag is set to 1. No other flags are affected.

0]	0	1	Τ	1	T	0	Τ	1	Τ	1	1
				C١	ycle	s:	1					
				S	tate	s:	4					
				F	=lag	s:	CY	,				

Branch Group:

This group of instructions alter normal sequential program flow.

Condition flags are not affected by any instruction in this group.

The two types of branch instructions are unconditional and conditional. Unconditional transfers simply perform the specified operation on register PC (the program counter). Conditional transfers examine the status of one of the four processor flags to determine if the specified branch is to be executed. The conditions that may be specified are as follows:

CON	IDITION	CCC
NZ	– not zero (Z = 0)	000
Z	— zero (Z = 1)	001
NC	— no carry (CY = 0)	010
С	carry (CY = 1)	011
PO	 parity odd (P = 0) 	100
ΡE	— parity even (P = 1)	101
Р	— plus (S = 0)	110
М	— minus (S = 1)	111

(Jump) (PC) - (byte 3) (byte 2)

Control is transferred to the instruction whose ad-*All Mnemonics Copyrighted C Intel Corporation 1976,1977

dress is specified in byte 3 and byte 2 of the current instruction.

1	1		0 0	1	1
		low-orde	er addr		
		high-orde	er addr		
		Cycles:	3		
		States:	10		
		Addressing:	immediat	е	

Flags: none

Jcondition addr

If (CCC),

(PC) - (byte 3) (byte 2)

If the specified condition is true, control is transferred to the instruction whose address is specified in byte 3 and byte 2 of the current instruction; otherwise, control continues sequentially.

(Conditional jump)

1 1	ССС	С	0	1	0		
low-order⊨addr							
high-order addr							

Cycles:	2/3
States:	7/10
Addressing:	immediate
Flags:	none

CALL addr (Call)

USP) - 1'	 (PCH)
1101	, ,	(1 0117

((SP) - 2) ← (PCL)

(PC) - (byte 3) (byte 2)

The high-order eight bits of the next instruction address are moved to the memory location whose address is one less than the content of register SP. The low-order eight bits of the next instruction address are moved to the memory location whose address is two less than the content of register SP. The content of register SP is decremented by 2. Control is transferred to the instruction whose address is specified in byte 3 and byte 2 of the current instruction.

1	I	Τ	1	I	0	0	1	Τ	1	1	0	1
low-order addr												
high-order addr												

5
18
immediate/reg. indirect
none

Ccondition addr	(Condition call)
If (CCC),	
((SP) – 1)	← (PCH)
((SP) - 2)	(PCL)
(SP) 🖛	(SP) – 2

(PC) ← (byte 3) (byte 2)

If the specified condition is true, the actions specified in the CALL instruction (see above) are performed; otherwise, control continues sequentially.

1	1	С	С	С	1	0	0
low-order addr							
high-order addr							

Cycles:	2/5
States:	9/18
Addressing:	immediate/reg. indirect
Flags:	none

RST n (Restart)

((SP) – 1) ← (PCH) ((SP) - 2) ← (PCL) (SP) ← (SP) - 2

(PC) - 8 * (NNN)

The high-order eight bits of the next instruction address are moved to the memory location whose address is one less than the content of register SP. The low-order eight bits of the next instruction address are moved to the memory location whose address is two less than the content of register SP. The content of register SP is decremented by two. Control is transferred to the instruction whose address is eight times the content of NNN.

1	1	N	N	N	1	1	1
		(Cycles:	3			
		:	States:	12			
		Addr	essing:	reg.	indire	ct	
			Flags:	nor	e		

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	0	0	0	Ν	Ν	Ν	0	0	0

Program Counter After Restart



(PCL) ← (L)

The content of register H is moved to the high-order eight bits of register PC. The content of register L is moved to the low-order eight bits of register PC.

1	1	1	0	1	1 0	0	1	
		Ad	Cycles: States: dressing: Flags:	1 6 regi non	ster Ie			



(Return) (PCL) ← ((SP)); (PCH) ← ((SP) + 1); (SP) ← (SP) + 2;

The content of the memory location whose address is specified in register SP is moved to the low-order eight bits of register PC. The content of the memory location whose address is one more than the content of register SP is moved to the high-order eight bits of register PC. The content of register SP is incremented by 2.



(Conditional return)

If (CCC). (PCL) ← ((SP))

Rcondition

If the specified condition is true, the actions specified in the RET instruction (see above) are performed; otherwise, control continues sequentially.

	1	1	с	С	с с	0	0	0
			C	ycles:	1/3			
			Addre	essing: Flags:	reg.	z indire •	ct	
۰۸	II Mnen	nonics Co	pyrighte	d © Inte	el Corpo	ration 1	1976, 19	77

Stack, I/O, and Machine Control Group:

This group of instructions performs I/O, manipulates the Stack, and alters internal control flags.

Unless otherwise specified, condition flags are not affected by any instructions in this group.

PUSH rp (Push)

- ((SP) 1) ← (rh)
- ((SP) 2) ← (rl)
- (SP) 🖛 (SP) 2

The content of the high-order register of register pair rp is moved to the memory location whose address is one less than the content of register SP. The content of the low-order register of register pair rp is moved to the memory location whose address is two less than the content of register SP. The content of register SP is decremented by 2. Note: Register pair rp = SP may not be specified.

1	1	1	R	Р	0	1	0	1

Cycles:	- 3
States:	12
Addressing:	reg. indirect
Flags:	none

PUSH PSW	(Push processor status word)	
((SP) — 1	I)	

 $((SP) - 2)_0 \leftarrow (CY)$, $((SP) - 2)_1 \leftarrow X$ $((SP) - 2)_2 \leftarrow (P)$, $((SP) - 2)_3 \leftarrow X$ $((SP) - 2)_4 \leftarrow (AC)$, $((SP) - 2)_5 \leftarrow X$ $((SP) - 2)_6 \leftarrow (Z)$, $((SP) - 2)_7 \leftarrow (S)$ $(SP) \leftarrow (SP) - 2$ X: Undefined. The content of register A is moved to the memory location whose address is one less than register SP. The contents of the condition flags are assembled into a processor status word and the word is moved to the memory location whose address is two less

than the content of register SP. The content of register SP is decremented by two. 0 1 1 1 1 1 0 1 Cycles: 3 States: 12 Addressing: reg. indirect Flags: none

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F	LAG	WO	RC



X: undefined

POP rp (Pop)

(rl) 🔶 ((SP))

(rh) ← ((SP) + 1)

(SP) ← (SP) + 2

The content of the memory location, whose address is specified by the content of register SP, is moved to the low-order register of register pair rp. The content of the memory location, whose address is one more than the content of register SP, is moved to the highorder register of register pair rp. The content of register SP is incremented by 2. Note: Register pair rp = SP may not be specified.

1	1	RP	0	0	0	1
		Cycles: States:	3 10			
		Addressing:	rea.	indirec	t	

Flags: none

POP PSW (Pop processor status word)

 $(CY) \leftarrow ((SP))_{0}$ $(P) \leftarrow ((SP))_{2}$ $(AC) \leftarrow ((SP))_{4}$ $(Z) \leftarrow ((SP))_{6}$ $(S) \leftarrow ((SP))_{7}$ $(A) \leftarrow ((SP) + 1)$ $(SP) \leftarrow (SP) + 2$

The content of the memory location whose address is specified by the content of register SP is used to restore the condition flags. The content of the memory location whose address is one more than the content of register SP is moved to register A. The content of register SP is incremented by 2.





(L) ↔ ((SP))

(H) → ((SP) + 1)

The content of the L register is exchanged with the content of the memory location whose address is specified by the content of register SP. The content of the H register is exchanged with the content of the memory location whose address is one more than the content of register SP.

1	1	1 0	1 0	0	1	1
		Cycle State Addressir Flag	es: 5 es: 1 ig: re gs: no	6 g. indir one	ect	

SPHL (Move HL to SP) (SP) ← (H) (L) The contents of registers H and L (16 bits) are moved to register SP.

1	1	Ι	1	Τ	1	Т	1	0	Τ	0	T	1
				C	vcle	s: -	1					-

States: 6 Addressing: register Flags: none

IN port (Input)

(A) 🔶 (data)

The data placed on the eight bit bi-directional data bus by the specified port is moved to register A.

novt	1	Т	1	0	1	1	0	1	1
μοι									

Cycles: 3 States: 10 Addressing: direct Flags: none

OUT port (Output)

(data) 🖛 (A)

The content of register A is placed on the eight bit bi-directional data bus for transmission to the specified port.

1	Τ	1	Ι	0		1	0		0	T	1	I	1
						ро	t						
					Сус	les:	3						
					Sta	tes:	10						
				Add	dress	ing:	dire	ect					
					F١	ags:	nor	ne					

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EI (Enable interrupts)

The interrupt system is enabled following the execution of the next instruction.



DI (Disable interrupts)

The interrupt system is disabled immediately following the execution of the DI instruction.



HLT (Halt)

The processor is stopped. The registers and flags are unaffected.



States: 5 Flags: none

NOP (No op)

No operation is performed. The registers and flags are unaffected.

0	1 0	0	0	1	0		0	0	10
			Cycles	5:	1				
			States	5:	4				
			Flags	s:	no	ne			

RIM (Read Interrupt Mask)

After the execution of the RIM instruction, the accumulator is loaded with the restart interrupt masks, any pending interrupts, and the contents of the serial input data line (SID).



SIM (Set Interrupt Masks)

During execution of the SIM instruction, the contents of the accumulator will be used in programming the restart interrupt masks. Bits 0-2 will set/reset the mask bit for RST 5.5, 6.5, 7.5 of the interrupt mask register, if bit 3 is 1 ("set"). Bit 3 is a "Mask Set Enable" control.

Setting the mask (i.e. masked bit = 1) disables the corresponding interrupt.

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-	Set	Reset
RST 5.5 MASK	if bit 0 = 1	if bit 0 = 0
RST 6.5 MASK	bit 1 = 1	bit 1 = 0
RST 7.5 MASK	bit 2 = 1	bit 2 = 0

RST 7.5 (edge trigger enable) internal request flip flop will be reset if bit 4 of the accumulator = 1; regardless of whether RST 7.5 is masked or not.

RESET IN input (pin 36) will set all RST MASKs, and reset/disable all interrupts.

SIM can, also, load the SOD output latch. Accumulator bit 7 is loaded into the SOD latch if bit 6 is set. The latch is unaffected if bit 6 is a zero. RESET IN input sets the SOD latch to zero.



INSTRUCTION SET

Summary of Processor Instructions

				Inst	ructi	on C	ode (1]		Clock ^[2]	1				Inst	ructi	on C	odel	1]		Clock (2
Mnemonic	Description	D7	D ₆	D5	D4	D:	3 D ₂	D	D ₀	Cycles	Mnemonic	Description	· D7	D ₆	D ₅	04	Dg	D ₂	D	1 ^D 0	Cycles
MOV	Move register to register	0	1	D	D	D	S	s	s	5	RZ	Return on zero	1	1	0	0	-1	0	0	0	5/11
MOV M, r	Move register to memory	Ō	1	1	1	0	S	S	S	7	RNZ	Return on no zero	1	1	0	0	0	. 0	C	0	5/11
MOV r, M	Move memory to register	0	1	D	D	D	1	1	0	7	RP	Return on positive	1	1	1	1	0	0	0	0	5/11
HLT	Halt	Û	1	1	1	0	1	1	0	7	RM	Return on minus	1	1	1	1	1	0	0	0	5/11
MVIr	Move immediate register	0	0	D	D	D	1	1	0	1	RPE	Return on parity even	1	1	1	0	1	0	0	0	5/11
MVIM	Move immediate memory	0	0	1	1	0	1	1	0	10	RPO	Return on parity odd	1	1	1	0	0	0	0	0	5/11
	Increment register	0	0	U D	D.	D D	÷	0	1	5	HSI	Restart	1		A	A.	A	0	;	-	10
INRM	Increment memory	n	ñ	1	1	n	1	0	n	10		Outout	1	1	0	-	1	0	1	-	10
DCR M	Decrement memory	õ	õ	i	i	õ	i	Ö	ĩ	10		L nad ummediate register	'n	'n	n	0	n	n	'n	i	10
ADDr	Add register to A	1	Ō	ò	Ó	0	S	S	S	4		Pair B & C	v	Ū	Ū	Ū	v	Ŭ	v	•	
ADC r	Add register to A with carry	1	0	0	0	1	S	S	S	4	LXID	Load immediate register	0	0	0	1	0	0	0	1	10
SUB r	Subtract register from A	1	0	0	1	0	S	S	S	4		Pair D & E									
SBBr	Subtract register from A with borrow	1	0	0	1	1	S	S	S	4	LXIH	Load immediate register Pair H & L	0	0	1	0	0	0	0	1	10
ANA r	And register with A	1	0	1	0	0	S	S	S	4	LXI SP	Load immediate stack pointer	0	0	1	1	0	0	0	1	10
XRA r	Exclusive Or register with A	1	0	1	0	1	S	S	S	4	PUSH B	Push register Pair B & C on	1	1	0	.0	0	1	0	1	11.
ORAr	Or register with A	1	0	1	1	0	S	S	S	4	[stack									
CMPr	Compare register with A	1	0	1	1	1	S	S	S	4	PUSH D	Push register Pair D & E on	1	1	0	1	0	1	0	1	11
	Add memory to A		U	0	U	0			U	1		stack					_				
	Add memory to A with carry		U	0	U 1	0	-	1	0	7	PUSH H	Push register Pair H & L on	1	1	1	0	0	1	0	1	11
SBB M	Subtract memory from A Subtract memory from A	1	0	0	1	1	1	1	0	7	PUSH PSW	stack Push A and Flags	1	1	1	1	0	1	0	1	11
ΔΝΔ Μ	And memory with A	,	n	1	n	٥	1	1	n	7	000 0	on stack			•	^	•	•	•		10
XRAM	Exclusive Or memory with A	i	ñ	i	n	1	i	i	n	7	FUFB	stack	1	1	U	υ	U	U	U	I	10
ORAM	Or memory with A	1	Õ	i	1	0	1	1	0	7	POPD	Pop register pair D & F off	i	1	n	1	n	n	n	1	10
CMP M	Compare memory with A	1	0	1	1	1	1	1	0	7		stack		•	U	•	Ŭ	Ũ	v	•	
ADI	Add immediate to A	1	1	0	0	0	1	1	0	7	РОР Н	Pop register pair H & L off	1	1	1	0	0	0	0	1	10
ACI	Add immediate to A with carry	1	1	0	0	1	1	1	0	7	POP PSW	stack Pop A and Flags	1	1	1	1	0	0	0	1	10
SUI	Subtract immediate from A	1	1	0	1	0	1	1	0	7		off stack									
SBI	Subtract immediate from A	1	1	0	1	1	1	1	0	7	STA	Store A direct	0	0	1	1	0	0	1	0	13
	with borrow									-	LDA	Load A direct	0	0	1	1	1	0	1	0	13
ANI XRI	And immediate with A Exclusive Or immediate with	1	1	1	0	0 1	1	1	0	7 7	XCHG	Exchange D & E, H & L Registers	1	1	1	0	1	0	1	1	4
0.01	A Or immediate with A		,	,	,	0		1	0	7	XTHL	Exchange top of stack, H & L	1	1	1	0	0	0	1	1	18
	Compare immediate with A	1		+	1	1		1	0	7	SPHL	H & L to stack pointer	1	1	1	1	1	U	0	1	5
RIC	Rotate A left	0	'n	'n	'n	0	i	1	1	4		H & L to program counter	1	1	1	0	1	0	0	-	5
RRC	Botate A right	0	ñ	n	n	1	i	1	i	4			0	n	0	1	1	0	0	1	10
RAL	Rotate A left through carry	ŏ	õ	õ	ĩ	ò	1	i	i	4	ПАПН		ñ	ñ	1		1	ñ	n	1	10
RAR	Rotate A right through	0	0	0	1	1	1	1	1	4	DAD SP	Add stack pointer to H & L	ñ	õ	1	ĩ	i	ŏ	Ő	1	10
	carry										STAX B	Store A indirect	· Ö	Ō	Ó	Ó	0	Ō	1	Ó	7
JMP	Jump unconditional	1	1	G	0	0	0	1	1	10	STAX D	Store A indirect	0	0	0	1	0	0	1	0	7
JC	Jump on carry	1	1	0	1	1	0	1	0	10	LDAX B	Load A indirect	0	0	0	0	1	0	1	0	1
JNC	Jump on no carry	1	1	0	1	0	0	1	0	10	LDAXD	Load A indirect	0	0	0	1	1	0	1	0	7
JZ	Jump on zero	1	1.	0	0	1	0	1	0	10	INX B	Increment B & C registers	0	0	0.	0	0	0	1	1	5
10	Jump on no zero	1	1	1	1	0	0	-	0	10	INXD	Increment D & E registers	0	0	0	1	0	0	1	1	5
JF IM	lump on positive	1		1	1	1	n	÷	0	10		Increment H & L registers	0	0	1	0	0	0	1	1	5
JPF	Jump on parity even	i	i	i	'n	i	Ő	i	Ő	10		Decrement Stack pointer	0	0	0	0	1	0		1	5
JPO	Jump on parity odd	i	i	1	Õ	0	Ō	1	Ō	10		Decrement D & E	0	0	0	1	-	n	+	1	5
CALL	Call unconditional	1	1	Ó	0	r	1	0	1	17	псхн	Decrement H & L	ñ	ñ	1	ó	i	n	÷	1	5
CC	Call on carry	1	1	0	1	1	1	0	0	11/17	DCX SP	Decrement stack pointer	Ő	õ	i	1	i	õ	i	i	5
CNC	Call on no carry	<u>1</u>	1	0	1	0	1	0	0	11/17	CMA	Complement A	ō	Ō	1	Ó	1	1	1	1	4
CZ	Call on zero	1	1	0	0	1	1	0	0	11/17	STC	Set carry	0	0	1	1	0	1	1	1	4
CNZ	Call on no zero	1	1	0	0	0	1	0	0	11/17	CMC	Complement carry	0	0	1	1	1	1	1	1	4
CP	Call on positive	1	1	1	1	0	1	0	0	11/17	DAA	Decimal adjust A	0	0	1	0	0	1	1	1	4
UM	Call on minus	1	1	1	1	1	1	0	0	11/17	SHLD	Store H & L direct	0	0	1	0	0	0	1	0	16
CPO	Call on parity even	1	1	1	U	1	1	U	U	11/17		Load H & L direct	0	0.	1	0	1	U	1	0	16
RFT	Can on parity 000 Return		1	0	U n	0	ו ח	U n	U 1	10		Enable Interrupts	1	1	1	1	1	U	1	1	4
RC	Return on carry	1	i	n	1	1	n	n	ò	5/11	NOP	No.operation	'n	0	'n	n n	U N	U N	n n	'n	4
RNC	Return on no carry	i	1	Ő	i	Ö	õ	õ	õ	5/11	BIM	Read Interrunt Mask	n	0	1	n	ñ	0	n	n	4
				-		-					SIM	Set Interrunt Mask	n	ñ	i	1	ņ	ñ	n	ñ	4
												eet monthly man	•	•				•			-

NOTES: 1. DDD or SSS - 000 B - 001 C - 010 D - 011 E - 100 H - 101 L - 110 Memory - 111 A. 2. Two possible cycle times, (5/11) indicate instruction cycles dependent on condition flags.



B-1. INTRODUCTION

This appendix provides information required to modify a Model ASR-33 Teletypewriter for use with certain Intel SBC 80 computer systems.

B-2. INTERNAL MODIFICATIONS

/WARNING/

Hazardous voltages are exposed when the top cover of the teletypewriter is removed. To prevent accidental shock, disconnect the teleprinter power cord before proceeding beyond this point.

Remove the top cover and modify the teletypewriter as follows:

- a. Remove blue lead from 750-ohm tap on current source register; reconnect this lead to 1450-ohm tap. (Refer to figures B-1 and B-2.)
- b. On terminal block, change two wires as follows to create an internal full-duplex loop (refer to figures B-1 and B-3):
 - 1. Remove brown/yellow lead from terminal 3; reconnect this lead to terminal 5.
 - 2. Remove white/blue lead from terminal 4; reconnect this lead to terminal 5.
- c. On terminal block, remove violet lead from terminal 8; reconnect this lead to terminal 9. This changes the receiver current level from 60 mA to 20 mA.

A relay circuit card must be fabricated and connected to the paper tape reader drive circuit. The relay circuit card to be fabricated requires a relay, a diode, a thyractor, a small 'vector' board for mounting the components, and suitable hardware for mounting the assembled relay card.

A circuit diagram of the relay circuit card is included in figure B-4; this diagram also includes the part numbers of the relay, diode, and thyractor. (Note that a 470-ohm resistor and a 0.1 μ F capacitor may be substituted for the thyractor.) After the relay circuit card has been

assembled, mount it in position as shown in figure B-5. Secure the card to the base plate using two self-tapping screws. Connect the relay circuit to the distributor trip magnet and mode switch as follows:

- a. Refer to figure B-4 and connect a wire (Wire 'A') from relay circuit card to terminal L2 on mode switch. (See figure B-6.)
- b. Disconnect brown wire shown in figure B-7 from plastic connector. Connect this brown wire to terminal L2 on mode switch. (Brown wire will have to be extended.)
- c. Refer to figure B-4 and connect a wire (Wire 'B') from relay circuit board to terminal L1 on mode switch.

B-3. EXTERNAL CONNECTIONS

Connect a two-wire receive loop, a two-wire send loop, and a two-wire tape reader control loop to the external device as shown in figure B-4. The external connector pin numbers shown in figure B-4 are for interface with an RS232C device.

B-4. SBC 530 TTY ADAPTER

The SBC 530, which converts RS232C signal levels to an optically isolated 20 mA current loop interface, provides signal translation for transmitted data, received data, and a paper tape reader relay. The SBC 530 interfaces an Intel SBC 80 computer system to a teletype-writer as shown in figure B-8.

The SBC 530 requires +12V at 98 mA and -12V at 98 mA. An auxiliary supply must be used if the SBC 80 system does not supply this power. A schematic diagram of the SBC 530 is supplied with the unit. The following auxiliary power connector (or equivalent) must be procured by the user:

Connector, Molex 09-50-7071 Pins, Molex 08-50-0106 Polarizing Key, Molex 15-04-0219



Figure B-1. Teletype Component Layout



Figure B-2. Current Source Resistor



Figure B-3. Terminal Block



Figure B-4. Teletypewriter Modifications



Figure B-5. Relay Circuit



Figure B-6. Mode Switch



Figure B-7. Distributor Trip Magnet



Figure B-8. TTY Adapter Cabling

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